

# **CLRC663**

High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Rev. 4.6 — 16 May 2018 171146

Product data sheet COMPANY PUBLIC

## **1** General description

CLRC663, the high performance multi-protocol NFC frontend.

The CLRC663 multi-protocol NFC frontend IC supports the following operating modes

- Read/write mode supporting ISO/IEC 14443 type A and MIFARE Classic
- Read/write mode supporting ISO/IEC 14443B
- Read/write mode supporting JIS X 6319-4 (comparable with FeliCa<sup>1</sup> (see <u>Section 20.5</u>) scheme)
- Passive initiator mode according to ISO/IEC 18092
- Read/write mode supporting ISO/IEC 15693
- Read/write mode supporting ICODE EPC UID/ EPC OTP
- Read/write mode supporting ISO/IEC 18000-3 mode 3/ EPC Class-1 HF

The CLRC663's internal transmitter is able to drive a reader/writer antenna designed to communicate with ISO/IEC 14443A and MIFARE Cassic IC-based cards and transponders without additional active circuitry. The digital module manages the complete ISO/IEC 14443A framing and error detection functionality (parity and CRC).

The CLRC663 supports MIFARE Classic with 1 kB memory, MIFARE Classic with 4 kB memory, MIFARE Ultralight, MIFARE Ultralight C, MIFARE Plus and MIFARE DESFire products. The CLRC663 supports higher transfer speeds of the MIFARE product family of up to 848 kbit/s in both directions.

The CLRC663 supports layer 2 and 3 of the ISO/IEC 14443B reader/writer communication scheme except anticollision. The anticollision needs to be implemented in the firmware of the host controller as well as in the upper layers.

The CLRC663 is able to demodulate and decode FeliCa coded signals. The FeliCa receiver part provides the demodulation and decoding circuitry for FeliCa coded signals. The CLRC663 handles the FeliCa framing and error detection such as CRC. The CLRC663 supports FeliCa higher transfer speeds of up to 424 kbit/s in both directions.

The CLRC663 is supporting the P2P passive initiator mode in accordance with ISO/IEC 18092.

The CLRC663 supports the vicinity protocol according to ISO/IEC15693, EPC UID and ISO/IEC 18000-3 mode 3/ EPC Class-1 HF.

The following host interfaces are supported:

- Serial Peripheral Interface (SPI)
- Serial UART (similar to RS232 with voltage levels dependent on pin voltage supply)
- I<sup>2</sup>C-bus interface (two versions are implemented: I2C and I2CL)

The CLRC663 supports the connection of a secure access module (SAM). A dedicated separate I2C interface is implemented for a connection of the SAM. The SAM can be



<sup>1</sup> In the following the word FeliCa is used for JIS X 6319-4

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used for high secure key storage and acts as a very performant crypto-coprocessor. A dedicated SAM is available for connection to the CLRC663.

In this document the term "MIFARE Classic card" refers to a MIFARE Classic IC-based contactless card.

## 2 Features and benefits

- Includes NXP ISO/IEC14443-A and Innovatron ISO/IEC14443-B intellectual property licensing rights
- High performance multi-protocol NFC frontend for transfer speed up to 848 kbit/s
- Supports ISO/IEC 14443 type A, MIFARE Classic, ISO/IEC 14443 B and FeliCa reader modes
- P2P passive initiator mode in accordance with ISO/IEC 18092
- Supports ISO/IEC15693, ICODE EPC UID and ISO/IEC 18000-3 mode 3/ EPC Class-1 HF
- Supports MIFARE Classic product encryption by hardware in read/write mode Allows reading cards based on MIFARE Ultralight, MIFARE Classic with 1 kB memory, MIFARE Classic with 4 kB memory, MIFARE DESFire EV1, MIFARE DESFire EV2 and MIFARE Plus ICs
- Low-Power Card Detection
- Compliance to EMV contactless protocol specification on RF level can be achieved
- Supported host interfaces:
  - SPI up to 10 Mbit/s
  - I<sup>2</sup>C-bus interfaces up to 400 kBd in Fast mode, up to 1000 kBd in Fast mode plus
  - RS232 Serial UART up to 1228.8 kBd, with voltage levels dependent on pin voltage supply
- Separate I<sup>2</sup>C-bus interface for connection of a secure access module (SAM)
- · FIFO buffer with size of 512 bytes for highest transaction performance
- Flexible and efficient power-saving modes including hard power down, standby and low-power card detection
- · Cost saving by integrated PLL to derive system clock from 27.12 MHz RF quartz crystal
- 3.0 V to 5.5 V power supply (CLRC66301, CLRC66302)
  - 2.5 V to 5.5 V power supply (CLRC66303)
- Up to 8 free programmable input/output pins
- Typical operating distance in read/write mode for communication to a ISO/IEC 14443 type A and MIFARE Classic card up to 12 cm, depending on the antenna size and tuning

The version CLRC66303 offers a more flexible configuration for Low-Power Card detection compared to the CLRC66301 and CLRC66302 with the new register LPCD\_OPTIONS. In addition, the CLRC66303 offers new additional settings for the Load Protocol which fit very well to smaller antennas. The CLRC66303 is therefore the recommended version for new designs.

#### Quick reference data 3

#### Table 1. Quick reference data CLRC66301HN and CLRC66302HN

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>DD</sub>	supply voltage			3.0	5.0	5.5	V
V <sub>DD(PVDD)</sub>	PVDD supply voltage		[1]	3.0	5.0	V <sub>DD</sub>	V
V <sub>DD(TVDD)</sub>	TVDD supply voltage			3.0	5.0	5.5	V
I <sub>pd</sub>	power-down current	PDOWN pin pulled HIGH	[2]	-	8	40	nA
I <sub>DD</sub>	supply current			-	17	20	mA
I <sub>DD(TVDD)</sub>	TVDD supply current			-	100	250	mA
T <sub>amb</sub>	operating ambient temperature			-25	+25	+85	°C
T <sub>stg</sub>	storage temperature	no supply voltage applied		-55	+25	+125	°C

VDD(PVDD) must always be the same or lower voltage than VDD.  $I_{\text{pd}}$  is the sum of all supply currents [1]

[2]

#### Table 2. Quick reference data CLRC66303HN

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>DD</sub>	supply voltage			2.5	5.0	5.5	V
V <sub>DD(PVDD)</sub>	PVDD supply voltage		[1]	2.5	5.0	$V_{DD}$	V
V <sub>DD(TVDD)</sub>	TVDD supply voltage			2.5	5.0	5.5	V
I <sub>pd</sub>	power-down current	PDOWN pin pulled HIGH	[2]	-	8	40	nA
I <sub>DD</sub>	supply current			-	17	20	mA
I <sub>DD(TVDD)</sub>	TVDD supply current			-	180	350	mA
		absolute limiting value		-	-	500	mA
T <sub>amb</sub>	operating ambient temperature	device mounted on PCB which allows sufficient heat dissipation		-40	+25	+105	°C
T <sub>stg</sub>	storage temperature	no supply voltage applied		-55	+25	+125	°C

VDD(PVDD) must always be the same or lower voltage than VDD. [1] [2]

I<sub>pd</sub> is the sum of all supply currents

#### **Ordering information** 4

#### Table 3. Ordering information

Type number	Package					
	Name	Description	Version			
CLRC66301HN/TRAYB <sup>[1]</sup>	HVQFN32	plastic thermal enhanced very thin quad flat package; no	SOT617-1			
CLRC66301HN/TRAYM <sup>[2]</sup>		leads; MSL2, 32 terminals + 1 central ground; body 5 × 5 × 0.85 mm				
CLRC66302HN/TRAYB <sup>[1]</sup>		plastic thermal enhanced very thin quad flat package; no	SOT617-1			
CLRC66302HN/TRAYBM <sup>[2]</sup>		leads; MSL1, 32 terminals + 1 central ground; body 5 × 5 × 0.85 mm				
CLRC66302HN/T/R <sup>[3]</sup>						
CLRC66303HN/TRAYB <sup>[1]</sup>		plastic thermal enhanced very thin quad flat package; no	SOT617-1			
CLRC66303HN/T/R <sup>[3]</sup>		leads; MSL2, 32 terminals + 1 central ground; body 5 x 5 x 0.85 mm, wettable flanks				

Delivered in one tray, MOQ (Minimum order quantity) : 490 pcs Delivered in five trays; MOQ: 5x 490 pcs Delivered on reel with 6000 pieces; MOQ: 6000 pcs [1]

[2] [3]

## 5 Block diagram

The analog interface handles the modulation and demodulation of the antenna signals for the contactless interface.

The contactless UART manages the protocol dependency of the contactless interface settings managed by the host.

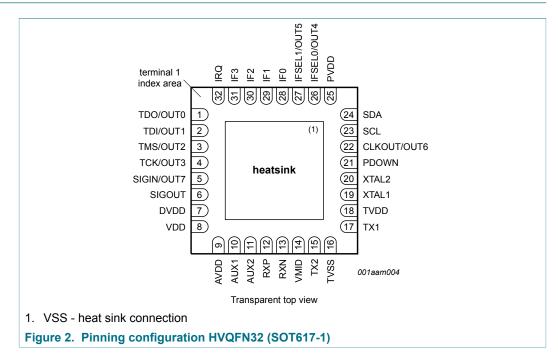
The FIFO buffer ensures fast and convenient data transfer between host and the contactless UART.

The register bank contains the settings for the analog and digital functionality.

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## 6 Pinning information



## 6.1 Pin description

#### Table 4. Pin description

Pin	Symbol	Туре	Description
1	TDO / OUT0	0	test data output for boundary scan interface / general purpose output 0
2	TDI / OUT1	I	test data input boundary scan interface / general purpose output 1
3	TMS / OUT2	I	test mode select boundary scan interface / general purpose output 2
4	TCK / OUT3	I	test clock boundary scan interface / general purpose output 3
5	SIGIN /OUT7	I	Contactless communication interface output. / general purpose output 7
6	SIGOUT	0	Contactless communication interface input.
7	DVDD	PWR	digital power supply buffer <sup>[1]</sup>
8	VDD	PWR	power supply
9	AVDD	PWR	analog power supply buffer <sup>[1]</sup>
10	AUX1	0	auxiliary outputs: Pin is used for analog test signal
11	AUX2	0	auxiliary outputs: Pin is used for analog test signal
12	RXP	I	receiver input pin for the received RF signal.
13	RXN	I	receiver input pin for the received RF signal.
14	VMID	PWR	internal receiver reference voltage [1]
15	TX2	0	transmitter 2: delivers the modulated 13.56 MHz carrier
16	TVSS	PWR	transmitter ground, supplies the output stage of TX1, TX2
17	TX1	0	transmitter 1: delivers the modulated 13.56 MHz carrier

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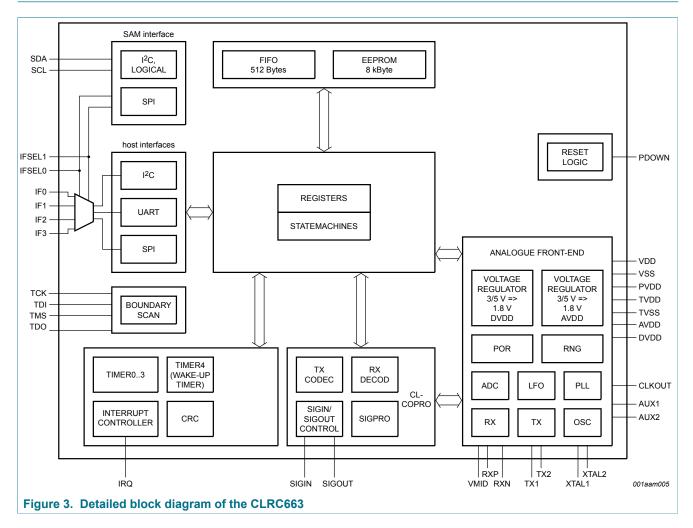
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Pin	Symbol	Туре	Description
18	TVDD	PWR	transmitter voltage supply
19	XTAL1	1	crystal oscillator input: Input to the inverting amplifier of the oscillator. This pin is also the input for an externally generated clock (fosc = 27.12 MHz)
20	XTAL2	0	crystal oscillator output: output of the inverting amplifier of the oscillator
21	PDOWN	I	Power Down (RESET)
22	CLKOUT / OUT6	0	clock output / general purpose output 6
23	SCL	0	Serial Clock line
24	SDA	I/O	Serial Data Line
25	PVDD	PWR	pad power supply
26	IFSEL0 / OUT4	I	host interface selection 0 / general purpose output 4
27	IFSEL1 / OUT5	I	host interface selection 1 / general purpose output 5
28	IF0	I/O	interface pin, multifunction pin: Can be assigned to host interface RS232, SPI, $\rm I^2C,  I^2C\text{-}L$
29	IF1	I/O	interface pin, multifunction pin: Can be assigned to host interface SPI, I <sup>2</sup> C, I <sup>2</sup> C-L
30	IF2	I/O	interface pin, multifunction pin: Can be assigned to host interface RS232, SPI, $\rm I^2C,  I^2C\text{-}L$
31	IF3	I/O	interface pin, multifunction pin: Can be assigned to host interface RS232, SPI, $\rm I^2C,  I^2C\text{-}L$
32	IRQ	0	interrupt request: output to signal an interrupt event
33	VSS	PWR	ground and heat sink connection

[1] This pin is used for connection of a buffer capacitor. Connection of a supply voltage might damage the device.

## 7 Functional description



## 7.1 Interrupt controller

The interrupt controller handles the enabling/disabling of interrupt requests. All of the interrupts can be configured by firmware. Additionally, the firmware has possibilities to trigger interrupts or clear pending interrupt requests. Two 8-bit interrupt registers IRQ0 and IRQ1 are implemented, accompanied by two 8-bit interrupt enable registers IRQ0En and IRQ1En. A dedicated functionality of bit 7 to set and clear bits 0 to 6 in this interrupt controller register is implemented.

The CLRC663 indicates certain events by setting bit IRQ in the register Status1Reg and additionally, if activated, by pin IRQ. The signal on pin IRQ may be used to interrupt the host using its interrupt handling capabilities. This allows the implementation of efficient host software.

Table 4. shows the available interrupt bits, the corresponding source and the condition for its activation. The interrupt bits Timer0IRQ, Timer1IRQ, Timer2IRQ, Timer3OIRQ, in register IRQ1 indicate an interrupt set by the timer unit. The setting is done if the timer underflows.

The TxIRQ bit in register IRQ0 indicates that the transmission is finished. If the state changes from sending data to transmitting the end of the frame pattern, the transmitter unit sets the interrupt bit automatically.

The bit RxIRQ in register IRQ0 indicates an interrupt when the end of the received data is detected.

The bit IdleIRQ in register IRQ0 is set if a command finishes and the content of the command register changes to idle.

The register WaterLevel defines both - minimum and maximum warning levels - counting from top and from bottom of the FIFO by a single value.

The bit HiAlertIRQ in register IRQ0 is set to logic 1 if the HiAlert bit is set to logic 1, that means the FIFO data number has reached the top level as configured by the register WaterLevel and bit WaterLevelExtBit.

The bit LoAlertIRQ in register IRQ0 is set to logic 1 if the LoAlert bit is set to logic 1, that means the FIFO data number has reached the bottom level as configured by the register WaterLevel.

The bit ErrIRQ in register IRQ0 indicates an error detected by the contactless UART during receive. This is indicated by any bit set to logic 1 in register Error.

The bit LPCDIRQ in register IRQ0 indicates a card detected.

The bit RxSOFIRQ in register IRQ0 indicates a detection of a SOF or a subcarrier by the contactless UART during receiving.

The bit GlobalIRQ in register IRQ1 indicates an interrupt occurring at any other interrupt source when enabled.

Interrupt bit	Interrupt source	Is set automatically, when
Timer0IRQ	Timer Unit	the timer register T0 CounterVal underflows
Timer1IRQ	Timer Unit	the timer register T1 CounterVal underflows
Timer2IRQ	Timer Unit	the timer register T2 CounterVal underflows
Timer3IRQ	Timer Unit	the timer register T3 CounterVal underflows
TxIRQ	Transmitter	a transmitted data stream ends
RxIRQ	Receiver	a received data stream ends
IdleIRQ	Command Register	a command execution finishes
HiAlertIRQ	FIFO-buffer pointer	the FIFO data number has reached the top level as configured by the register WaterLevel
LoAlertIRQ	FIFO-buffer pointer	the FIFO data number has reached the bottom level as configured by the register WaterLevel
ErrIRQ	contactless UART	a communication error had been detected
LPCDIRQ	LPCD	a card was detected when in low-power card detection mode
RxSOFIRQ	Receiver	detection of a SOF or a subcarrier
GlobalIRQ	all interrupt sources	will be set if another interrupt request source is set

#### Table 5. Interrupt sources

## 7.2 Timer module

#### Timer module overview

The CLRC663 implements five timers. Four timers -Timer0 to Timer3 - have an input clock that can be configured by register T(x)Control to be 13.56 MHz, 212 kHz, (derived from the 27.12 MHz quartz) or to be the underflow event of the fifth Timer (Timer4). Each timer implements a counter register which is 16 bit wide. A reload value for the counter is defined in a range of 0000h to FFFFh in the registers TxReloadHi and TxReloadLo. The fifth timer Timer4 is intended to be used as a wake-up timer and is connected to the internal LFO (Low Frequency Oscillator) as input clock source.

The TControl register allows the global start and stop of each of the four timers Timer0 to Timer3. Additionally, this register indicates if one of the timers is running or stopped. Each of the five timers implements an individual configuration register set defining timer reload value (e.g. T0ReloadHi,T0ReloadLo), the timer value (e.g. T0CounterValHi, T0CounterValLo) and the conditions which define start, stop and clockfrequency (e.g. T0Control).

The external host may use these timers to manage timing relevant tasks. The timer unit may be used in one of the following configurations:

- Time-out counter
- Watch-dog counter
- · Stop watch
- Programmable one-shot timer
- · Periodical trigger

The timer unit can be used to measure the time interval between two events or to indicate that a specific event has occurred after an elapsed time. The timer register content is modified by the timer unit, which can be used to generate an interrupt to allow a host to react on this event.

The counter value of the timer is available in the registers T(x)CounterValHi, T(x)CounterValLo. The content of these registers is decremented at each timer clock.

If the counter value has reached a value of 0000h and the interrupts are enabled for this specific timer, an interrupt will be generated as soon as the next clock is received.

If enabled, the timer event can be indicated on the pin IRQ (interrupt request). The bit Timer(x)IRQ can be set and reset by the host controller. Depending on the configuration, the timer will stop counting at 0000h or restart with the value loaded from registers T(x)ReloadHi, T(x)ReloadLo.

The counting of the timer is indicated by bit TControl.T(x)Running.

The timer can be started by setting bits TControl.T(x)Running and TControl.T(x)StartStopNow or stopped by setting the bits TControl.T(x)StartStopNow and clearing TControl.T(x)Running.

Another possibility to start the timer is to set the bit T(x)Mode.T(x)Start. This can be useful if dedicated protocol requirements need to be fulfilled.

## 7.2.1 Timer modes

#### 7.2.1.1 Time-Out- and Watch-Dog-Counter

Having configured the timer by setting *register* T(x)ReloadValue and starting the counting of Timer(x) by setting bit TControl.T(x)StartStop and TControl.T(x)Running, the timer unit decrements the T(x)CounterValue Register beginning with the configured start event. If the configured stop event occurs before the Timer(x) underflows (e.g. a bit is received from the card), the timer unit stops (no interrupt is generated).

If no stop event occurs, the timer unit continues to decrement the counter registers until the content is zero and generates a timer interrupt request at the next clock cycle. This allows indicating to a host that the event did not occur during the configured time interval.

#### 7.2.1.2 Wake-up timer

The wake-up Timer4 allows to wake-up the system from standby after a predefined time. The system can be configured in such a way that it is entering the standby mode again in case no card had been detected.

This functionality can be used to implement a low-power card detection (LPCD). For the low-power card detection, it is recommended to set T4Control.T4AutoWakeUp and T4Control.T4AutoRestart, to activate the Timer4 and automatically set the system in standby. The internal low frequency oscillator (LFO) is then used as input clock for this Timer4. If a card is detected, the host-communication can be started. If bit T4Control.T4AutoWakeUp is not set, the CLRC663 will not enter the standby mode again in case no card is detected but stays fully powered.

#### 7.2.1.3 Stop watch

The elapsed time between a configured start- and stop event may be measured by the CLRC663 timer unit. By setting the registers T(x)ReloadValueHi, T(x)reloadValueLo the timer starts to decrement as soon as activated. If the configured stop event occurs, the timer stops decrementing. The elapsed time between start and stop event can then be calculated by the host dependent on the timer interval TTimer:

$$\Delta T = \left( Treload_{value} - Timer_{value} \right)^* T_{Timer}$$
(1)

If an underflow occurred which can be identified by evaluating the corresponding IRQ bit, the performed time measurement according to the formula above is not correct.

#### 7.2.1.4 Programmable one-shot timer

The host configures the interrupt and the timer, starts the timer and waits for the interrupt event on pin IRQ. After the configured time, the interrupt request will be raised.

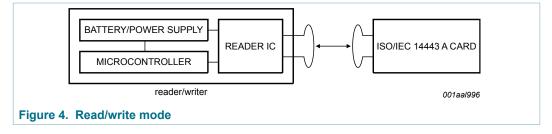
#### 7.2.1.5 Periodical trigger

If the bit T(x)Control. T(x)AutoRestart is set and the interrupt is activated, an interrupt request will be indicated periodically after every elapsed timer period.

## 7.3 Contactless interface unit

The contactless interface unit of the CLRC663 supports the following read/write operating modes:

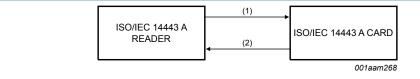
- ISO/IEC14443A/MIFARE
- ISO/IEC14443B
- FeliCa
- ISO/IEC15693/ICODE
- ICODE EPC UID
- · ISO/IEC 18000-3 mode 3/ EPC Class-1 HF



A typical system using the CLRC663 is using a microcontroller to implement the higher levels of the contactless communication protocol and a power supply (battery or external supply).

## 7.3.1 Communication mode for ISO/IEC 14443 type A and for MIFARE Classic

The physical level of the communication is shown in the following figure:



1. Reader to Card 100 % ASK, Miller Coded, Transfer speed 106 kbit/s to 848 kbit/s

 Card to Reader, Subcarrier Load Modulation Manchester Coded or BPSK, transfer speed 106 kbit/s to 848 kbit/s

Figure 5. Read/write mode for ISO/IEC 14443 type A and read/write mode for MIFARE Classic

The physical parameters are described in the following table:

#### Table 6. Communication overview for ISO/IEC 14443 type A and read/write mode for MIFARE Classic

Communication direction	Signal type	Transfer speed							
		106 kbit/s	212 kbit/s	424 kbit/s	848 kbit/s				
Reader to card (send data from the CLRC663 to a card) fc = 13.56 MHz	reader side modulation	100 % ASK	100% ASK	100% ASK	100% ASK				
	bit encoding	modified Miller encoding	modified Miller encoding	modified Miller encoding	modified Miller encoding				
	bit rate [kbit/s]	fc / 128	fc / 64	fc / 32	fc / 16				

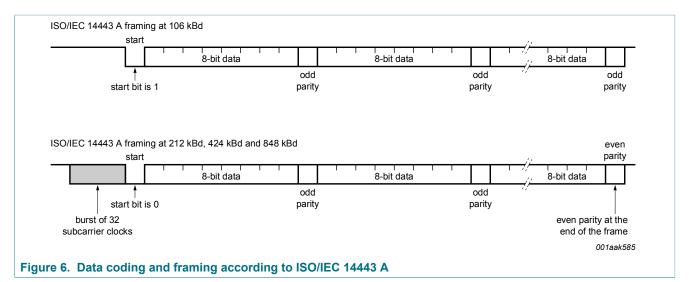
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Communication direction	Signal type	Transfer speed	Transfer speed							
		106 kbit/s	212 kbit/s	424 kbit/s	848 kbit/s					
Card to reader (CLRC663 receives data from a card)	card side modulation	subcarrier load modulation	subcarrier load modulation	subcarrier load modulation	subcarrier load modulation					
	subcarrier frequency	fc / 16	fc / 16	fc / 16	fc / 16					
	bit encoding	Manchester encoding	BPSK	BPSK	BPSK					

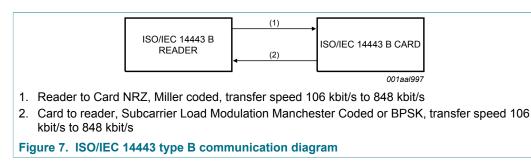
The CLRC663 connection to a host is required to manage the complete ISO/IEC 14443 type A and MIFARE Classic communication protocol. The following figure shows the data coding and framing according to ISO/IEC 14443 type A and MIFARE Classic..



The internal CRC coprocessor calculates the CRC value based on ISO/IEC 14443 A part 3 and handles parity generation internally according to the transfer speed.

## 7.3.2 ISO/IEC14443 type B functionality

The physical level of the communication is shown in the following figure:



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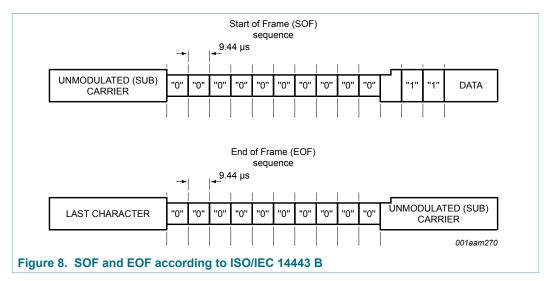
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Communication	Signal type	Transfer speed						
direction		106 kbit/s	212 kbit/s	424 kbit/s	848 kbit/s			
Reader to card (send data from the CLRC663 to a card) fc = 13.56 MHz	reader side modulation	10 % ASK	10 % ASK	10 % ASK	10 % ASK			
	bit encoding	NRZ	NRZ	NRZ	NRZ			
	bit rate [kbit/s]	128 / fc	64 / fc	32 / fc	16 / fc			
Card to reader (CLRC663 receives	card side modulation	subcarrier load modulation	subcarrier load modulation	subcarrier load modulation	subcarrier load modulation			
data from a card)	subcarrier frequency	fc / 16	fc / 16	fc / 16	fc / 16			
	bit encoding	BPSK	BPSK	BPSK	BPSK			

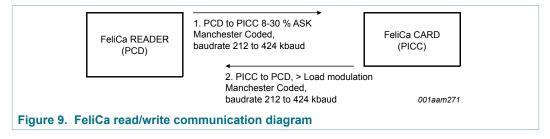
#### Table 7. Communication overview for ISO/IEC 14443 B reader/writer

The CLRC663 connected to a host is required to manage the complete ISO/IEC 14443 B protocol. The following figure shows the ISO/IEC 14443B SOF and EOF.



#### 7.3.3 FeliCa functionality

The FeliCa mode is the general reader/writer to card communication scheme according to the FeliCa specification. The communication on a physical level is shown in the following figure:



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Communication direction	Signal type	Transfer speed FeliCa	FeliCa higher transfe speeds		
		212 kbit/s	424 kbit/s		
Reader to card (send data from the CLRC663	reader side modulation	8 % to 30 % ASK	8 % to 30 % ASK		
to a card) fc = 13.56 MHz	bit encoding	Manchester encoding	Manchester encoding		
10 - 10.00 Miliz	bit rate	fc/64	fc/32		
Card to reader (CLRC663 receives data from a	card side load modulation	Load modulation	Load modulation		
card)	bit encoding	Manchester encoding	Manchester encoding		

#### Table 8. Communication overview for FeliCa reader/writer

The CLRC663 needs to be connected to a dedicated host to be able to support the complete FeliCa protocol.

#### 7.3.3.1 FeliCa framing and coding

#### Table 9. FeliCa framing and coding

Preamble (Hex.)			Sync (Hex.)	Sync Le (Hex.)		n-Data			CRC					
00	00	00	00	00	00	B2	4D							

To enable the FeliCa communication a 6 byte preamble (00h, 00h, 00h, 00h, 00h, 00h) and 2 bytes sync bytes (B2h, 4Dh) are sent to synchronize the receiver.

The following Len byte indicates the length of the sent data bytes plus the LEN byte itself.

The CRC calculation is done according to the FeliCa definitions with the MSB first.

To transmit data on the RF interface, the host controller has to send the Len- and databytes to the CLRC663's FIFO-buffer. The preamble and the sync bytes are generated by the CLRC663 automatically and must not be written to the FIFO by the host controller. The CLRC663 performs internally the CRC calculation and adds the result to the data frame.

#### 7.3.4 ISO/IEC15693 functionality

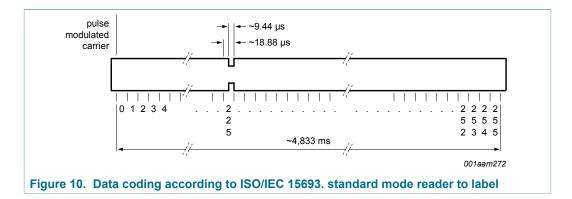
Table 10. Communication overview for ISO/IEC 15693 reader/writer reader to labe	el
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Communication	Signal type	Transfer speed			
direction		fc / 8192 kbit/s	fc / 512 kbit/s		
Reader to label (send data from the CLRC663 to a card)	reader side modulation	10 % to 30 % ASK or 100 % ASK	10 % to 30 % ASK 90 % to 100 % ASK		
	bit encoding	1/256	1/4		
	data rate	1.66 kbit/s	26.48 kbit/s		

Communication	olginal type						
direction		6.62 (6.67) kbit/s	13.24 kbit/s <sup>[1]</sup>	26.48 (26.69) kbit/s	52.96 kbit/s		
Label to reader (CLRC663 receives data from a card) fc = 13.56 MHz	card side modulation	not supported	not supported	single (dual) subcarrier load modulation ASK	single subcarrier load modulation ASK		
	bit length (µs)	-	-	37.76 (37.46)	18.88		
	bit encoding	-	-	Manchester coding	Manchester coding		
	subcarrier frequency [MHz]	-	-	fc / 32 (fc / 28)	fc / 32		

Table 11. Communication overview for ISO/IEC 15693 reader/writer label to readerCommunicationSignal typeTransfer speed

[1] Fast inventory (page) read command only (ICODE proprietary command).



#### 7.3.5 EPC-UID/UID-OTP functionality

Table 12.	Communication	overview for	· EPC/UID
-----------	---------------	--------------	-----------

Communication	Signal type	Transfer speed		
direction		26.48 kbit/s	52.96 kbit/s	
Reader to card	reader side modulation	10 % to 30 % ASK		
(send data from the CLRC663 to a card)	bit encoding	RTZ		
,	bit length	37.76 μs		
Card to reader (CLRC663 receives	card side modulation		single subcarrier load modulation	
data from a card)	bit length		18.88 µs	
	bit encoding		Manchester coding	

Data coding and framing according to EPC global 13.56 MHz ISM (industrial, scientific and medical) Band Class 1 Radio Frequency Identification Tag Interface Specification (Candidate Recommendation, Version 1.0.0).

## 7.3.6 ISO/IEC 18000-3 mode 3/ EPC Class-1 HF functionality

The ISO/IEC 18000-3 mode 3/ EPC Class-1 HF is not described in this document. For a detailed explanation of the protocol, refer to the ISO/IEC 18000-3 mode 3/ EPC Class-1 HF standard.

#### 7.3.6.1 Data encoding ICODE

The ICODE protocols have mainly three different methods of data encoding:

- "1" out of "4" coding scheme
- "1" out of "256" coding scheme
- "Return to Zero" (RZ) coding scheme

Data encoding for all three coding schemes is done by the ICODE generator.

The supported EPC Class-1 HF modes are:

- 2 pulse for 424 kbit subcarrier
- 4 pulse for 424 kbit subcarrier
- 2 pulse for 848 kbit subcarrier
- 4 pulse for 848 kbit subcarrier

## 7.3.7 ISO/IEC 18092 mode

The CLRC663 supports Passive Initiator Communication mode at the transfer speeds 106 kbit/s, 212 kbit/s and 424 kbit/s as defined in the ISO/IEC 18092 standard.

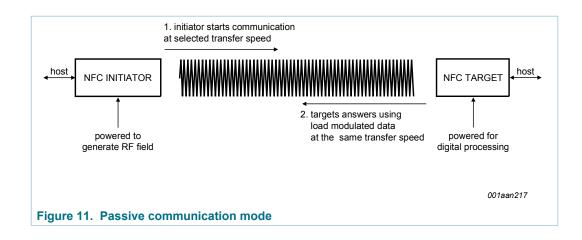
- Passive communication mode means that the target answers to an initiator command in a load modulation scheme. The initiator is active in terms of generating the RF field.
- Initiator: generates RF field at 13.56 MHz and starts the ISO/IEC 18092 communication.
- Target: responds to initiator command either in a load modulation scheme in Passive communication mode or using a self-generated and self-modulated RF field for Active Communication mode.

#### 7.3.7.1 Passive communication mode

Passive communication mode means that the target answers to an initiator command in a load modulation scheme. The initiator is active meaning generating the RF field.

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#### Table 13. Communication overview for Passive communication mode

Communication direction	106 kbit/s	212 kbit/s	424 kbit/s
Initiator $\rightarrow$ target	According to ISO/IEC 14443A 100 % ASK, Modified Miller Coded	According to FeliCa Manchester Coded	, 8 % to 30 % ASK
Target $\rightarrow$ initiator	According to ISO/IEC 14443A subcarrier load modulation, Manchester Coded	According to FeliCa Manchester Coded	, > 12 % ASK

The contactless UART of CLRC663 and a dedicated host controller are required to handle the ISO/IEC 18092 passive initiator protocol.

#### 7.3.7.2 ISO/IEC 18092 framing and coding

The ISO/IEC 18092 framing and coding in Passive communication mode is defined in the ISO/IEC 18092 standard.

#### Table 14. Framing and coding overview

Transfer speed	Framing and Coding
106 kbit/s	According to the ISO/IEC 14443 type A and MIFARE scheme
212 kbit/s	According to the FeliCa scheme
424 kbit/s	According to the FeliCa scheme

#### 7.3.7.3 ISO/IEC 18092 protocol support

The ISO/IEC 18092 protocol is not described in this document. For a detailed explanation of the protocol, refer to the ISO/IEC 18092 standard.

## 7.4 Host interfaces

## 7.4.1 Host interface configuration

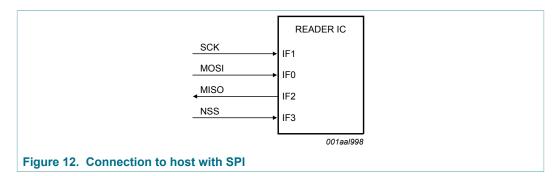
The CLRC663 supports direct interfacing of various hosts as the SPI, I<sup>2</sup>C, I<sup>2</sup>CL and serial UART interface type. The CLRC663 resets its interface and checks the current host interface type automatically having performed a power-up or resuming from power down. The CLRC663 identifies the host interface by the means of the logic levels on the control pins after the Cold Reset Phase. This is done by a combination of fixed pin connections. The following table shows the possible configurations defined by IFSEL1,IFSEL0:

Pin	Pin Symbol	UART	SPI	I <sup>2</sup> C	I <sup>2</sup> C-L
28	IF0	RX	MOSI	ADR1	ADR1
29	IF1	n.c.	SCK	SCL	SCL
30	IF2	тх	MISO	ADR2	SDA
31	IF3	PAD_VDD	NSS	SDA	ADR2
26	IFSEL0	VSS	VSS	PAD_VDD	PAD_VDD
27	IFSEL1	VSS	PAD_VDD	VSS	PAD_VDD

#### Table 15. Connection scheme for detecting the different interface types

## 7.4.2 SPI interface

#### 7.4.2.1 General



The CLRC663 acts as a slave during the SPI communication. The SPI clock SCK has to be generated by the master. Data communication from the master to the slave uses the Line MOSI. Line MISO is used to send data back from the CLRC663 to the master.

A serial peripheral interface (SPI compatible) is supported to enable high-speed communication to a host. The implemented SPI compatible interface is according to a standard SPI interface. The SPI compatible interface can handle data speed of up to 10 Mbit/s. In the communication with a host, CLRC663 acts as a slave receiving data from the external host for register settings and to send and receive data relevant for the communication on the RF interface.

NSS (Not Slave Select) enables or disables the SPI interface. When NSS is logical high, the interface is disabled and reset. Between every SPI command, the NSS must go to logical high to be able to start the next command read or write.

On both data lines (MOSI, MISO) each data byte is sent by MSB first. Data on MOSI line shall be stable on rising edge of the clock line (SCK) and is allowed to change on

falling edge. The same is valid for the MISO line. Data is provided by the CLRC663 on the falling edge and is stable on the rising edge. The polarity of the clock is low at SPI idle.

#### 7.4.2.2 Read data

To read out data from the CLRC663 by using the SPI compatible interface, the following byte order has to be used.

The first byte that is sent defines the mode (LSB bit) and the address.

Table 16.	<b>Bvte</b>	Order for	MOSI	and MISO

	byte 0	byte 1	byte 2	byte 3 to n-1	byte n	byte n+1
MOSI	address 0	address 1	address 2		address n	00h
MISO	Х	data 0	data 1		data n - 1	data n

Remark: The Most Significant Bit (MSB) has to be sent first.

#### 7.4.2.3 Write data

To write data to the CLRC663 using the SPI interface, the following byte order has to be used. It is possible to write more than one byte by sending a single address byte (see.8.5.2.4).

The first send byte defines both, the mode itself and the address byte.

#### Table 17. Byte Order for MOSI and MISO

	byte 0	byte 1	byte 2	3 to n-1	byte n	byte n + 1
MOSI	address 0	data 0	data 1		data n - 1	data n
MISO	Х	Х	Х		Х	Х

**Remark:** The Most Significant Bit (MSB) has to be sent first.

#### 7.4.2.4 Address byte

The address byte has to fulfill the following format:

The LSB bit of the first byte defines the used mode. To read data from the CLRC663, the LSB bit is set to logic 1. To write data to the CLRC663, the LSB bit has to be cleared. The bits 6 to 0 define the address byte.

NOTE: When writing the sequence [address byte][data0][data1][data2]..., [data0] is written to address [address byte], [data1] is written to address [address byte + 1] and [data2] is written to [address byte + 2].

Exception: This auto increment of the address byte is not performed if data is written to the FIFO address

7	6	5	4	3	2	1	0
address 6	address 5	address 4	address 3	address 2	address 1	address 0	1 (read) 0 (write)
MSB							LSB

#### Table 18. Address byte 0 register; address MOSI

#### 7.4.2.5 Timing Specification SPI

The timing condition for SPI interface is as follows:

Table 19. Timin	g conditions SPI				
Symbol	Parameter	Min	Тур	Мах	Unit
t <sub>SCKL</sub>	SCK LOW time	50	-	-	ns
t <sub>scкн</sub>	SCK HIGH time	50	-	-	ns
t <sub>h(SCKH-D)</sub>	SCK HIGH to data input hold time	25	-	-	ns
t <sub>su(D-SCKH)</sub>	data input to SCK HIGH set-up time	25	-	-	ns
t <sub>h(SCKL-Q)</sub>	SCK LOW to data output hold time	-	-	25	ns
$t_{(SCKL-NSSH)}$	SCK LOW to NSS HIGH time	0	-	-	ns
t <sub>NSSH</sub>	NSS HIGH time	50	-	-	ns

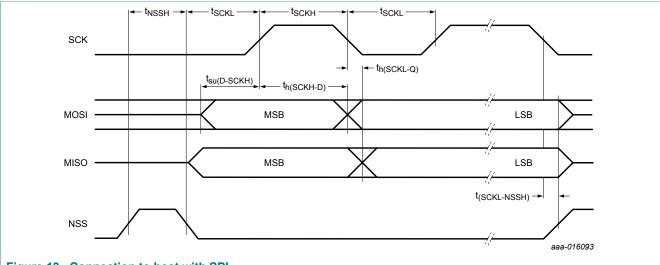


Figure 13. Connection to host with SPI

**Remark:** To send more bytes in one data stream, the NSS signal must be LOW during the send process. To send more than one data stream, the NSS signal must be HIGH between each data stream.

## 7.4.3 RS232 interface

#### 7.4.3.1 Selection of the transfer speeds

The internal UART interface is compatible to an RS232 serial interface. The levels supplied to the pins are between VSS and PVDD. To achieve full compatibility of the voltage levels to the RS232 specification, an RS232 level shifter is required.

<u>Table 21</u> describes examples for different transfer speeds and relevant register settings. The resulting transfer speed error is less than 1.5 % for all described transfer speeds. The default transfer speed is 115.2 kbit/s.

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To change the transfer speed, the host controller has to write a value for the new transfer speed to the register SerialSpeedReg. The bits BR\_T0 and BR\_T1 define factors to set the transfer speed in the SerialSpeedReg.

Table 20 describes the settings of BR\_T0 and BR\_T1.

Table 20.	Settinas	of BR	T0 and BR T	1

BR_T0	0	1	2	3	4	5	6	7
factor BR_T0	1	1	2	4	8	16	32	64
range BR_T1	1 to 32	33 to 64						

#### Table 21. Selectable transfer speeds

Transfer speed (kbit/s)	Serial SpeedReg	Transfer speed accuracy (%)
	(Hex.)	
7.2	FA	-0.25
9.6	EB	0.32
14.4	DA	-0.25
19.2	СВ	0.32
38.4	AB	0.32
57.6	9A	-0.25
115.2	7A	-0.25
128	74	-0.06
230.4	5A	-0.25
460.8	3A	-0.25
921.6	1C	1.45
1228.8	15	0.32

The selectable transfer speeds as shown are calculated according to the following formulas:

if BR\_T0 = 0: transfer speed = 27.12 MHz / (BR\_T1 + 1) if BR\_T0 > 0: transfer speed = 27.12 MHz / (BR\_T1 + 33)/ $2^{(BR_T0 - 1)}$ 

Remark: Transfer speeds above 1228.8 kBits/s are not supported.

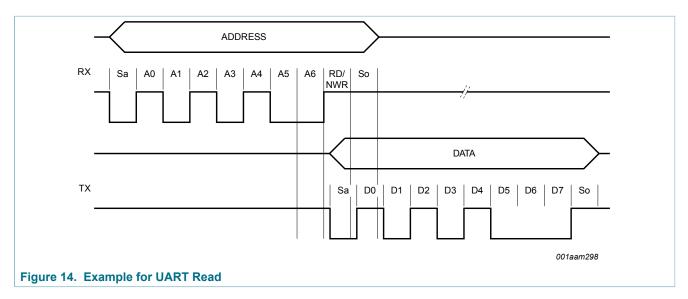
#### 7.4.3.2 Framing

Table 22. UART framing		
Bit	Length	Value
Start bit (Sa)	1 bit	0
Data bits	8 bit	Data
Stop bit (So)	1 bit	1

**Remark:** For data and address bytes, the LSB bit has to be sent first. No parity bit is used during transmission.

**Read data:** To read out data using the UART interface, the flow described below has to be used. The first send byte defines both the mode itself and the address. The Trigger on pin IF3 has to be set, otherwise no read of data is possible.

Table 23. Byte Order to Read Data					
Mode	byte 0	byte 1			
RX	address	-			
ТХ	-	data 0			



#### Write data:

To write data to the CLRC663 using the UART interface, the following sequence has to be used.

The first send byte defines both, the mode itself and the address.

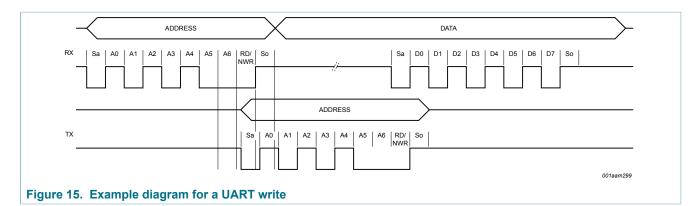
#### Table 24. Byte Order to Write Data

Mode	byte 0	byte 1
RX	address 0	data 0
ТХ		address 0

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Remark: Data can be sent before address is received.

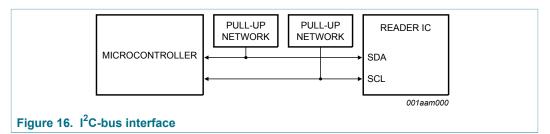
## 7.4.4 I<sup>2</sup>C-bus interface

#### 7.4.4.1 General

An Inter IC ( $I^2C$ ) bus interface is supported to enable a low cost, low pin count serial bus interface to the host. The implemented  $I^2C$  interface is mainly implemented according to the NXP Semiconductors  $I^2C$  interface specification, rev. 3.0, June 2007. The CLRC663 can act as a slave receiver or slave transmitter in standard mode, fast mode and fast mode plus.

The following features defined by the NXP Semiconductors I<sup>2</sup>C interface specification, rev. 3.0, June 2007 are not supported:

- The CLRC663 I2C interface does not stretch the clock
- The CLRC663 I2C interface does not support the general call. This means that the CLRC663 does not support a software reset
- · The CLRC663 does not support the I2C device ID
- The implemented interface can only act in slave mode. Therefore no clock generation and access arbitration is implemented in the CLRC663.
- · High-speed mode is not supported by the CLRC663



The voltage level on the I2C pins is not allowed to be higher than PVDD.

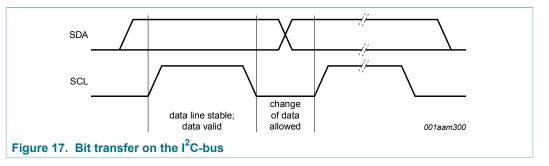
SDA is a bidirectional line, connected to a positive supply voltage via a pull-up resistor. Both lines SDA and SCL are set to HIGH level if no data is transmitted. Data on the  $l^2$ C-bus can be transferred at data rates of up to 400 kbit/s in fast mode, up to 1 Mbit/s in the fast mode+.

If the  $I^2C$  interface is selected, a spike suppression according to the  $I^2C$  interface specification on SCL and SDA is automatically activated.

For timing requirements, refer to Table 254.

## 7.4.4.2 I<sup>2</sup>C Data validity

Data on the SDA line shall be stable during the HIGH period of the clock. The HIGH state or LOW state of the data line shall only change when the clock signal on SCL is LOW.



## 7.4.4.3 I<sup>2</sup>C START and STOP conditions

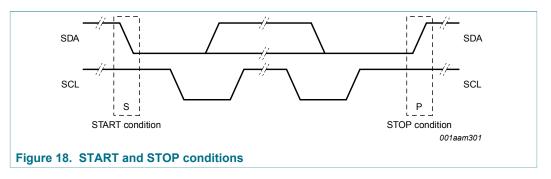
To handle the data transfer on the  $I^2$ C-bus, unique START (S) and STOP (P) conditions are defined.

A START condition is defined with a HIGH-to-LOW transition on the SDA line while SCL is HIGH.

A STOP condition is defined with a LOW-to-HIGH transition on the SDA line while SCL is HIGH.

The master always generates the START and STOP conditions. The bus is considered to be busy after the START condition. The bus is considered to be free again a certain time after the STOP condition.

The bus stays busy if a repeated START (Sr) is generated instead of a STOP condition. In this respect, the START (S) and repeated START (Sr) conditions are functionally identical. Therefore, the S symbol will be used as a generic term to represent both the START and repeated START (Sr) conditions.



## 7.4.4.4 I<sup>2</sup>C byte format

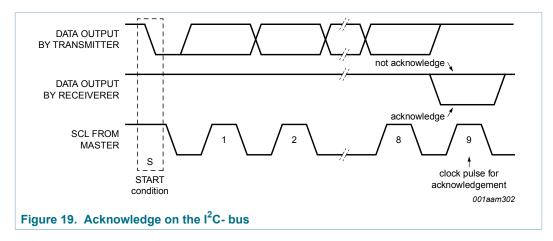
Each byte has to be followed by an acknowledge bit. Data is transferred with the MSB first, see <u>Figure 18</u>. The number of transmitted bytes during one data transfer is unrestricted but shall fulfill the read/write cycle format.

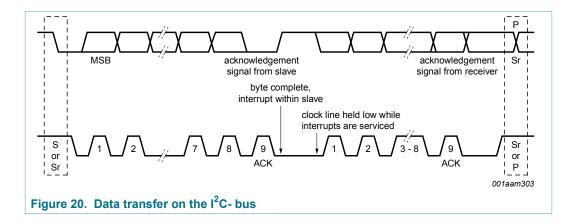
## 7.4.4.5 I<sup>2</sup>C Acknowledge

An acknowledge at the end of one data byte is mandatory. The acknowledge-related clock pulse is generated by the master. The transmitter of data, either master or slave, releases the SDA line (HIGH) during the acknowledge clock pulse. The receiver shall pull down the SDA line during the acknowledge clock pulse so that it remains stable LOW during the HIGH period of this clock pulse.

The master can then generate either a STOP (P) condition to stop the transfer, or a repeated START (Sr) condition to start a new transfer.

A master-receiver shall indicate the end of data to the slave- transmitter by not generating an acknowledge on the last byte that was clocked out by the slave. The slave-transmitter shall release the data line to allow the master to generate a STOP (P) or repeated START (Sr) condition.





## 7.4.4.6 I<sup>2</sup>C 7-bit addressing

During the I<sup>2</sup>C-bus addressing procedure, the first byte after the START condition is used to determine which slave will be selected by the master.

Alternatively the I<sup>2</sup>C address can be configured in the EEPROM. Several address numbers are reserved for this purpose. During device configuration, the designer has to ensure, that no collision with these reserved addresses in the system is possible. Check the corresponding I<sup>2</sup>C specification for a complete list of reserved addresses.

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For all CLRC663 devices, the upper 5 bits of the device bus address are reserved by NXP and set to 01010(bin). The remaining 2 bits (ADR\_2, ADR\_1) of the slave address can be freely configured by the customer in order to prevent collisions with other  $I^2C$  devices by using the interface pins (refer to <u>Table 15</u>) or the value of the  $I^2C$  address EEPROM register (refer to <u>Table 37</u>).

MSB							LSB
Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	R/W
		— sla	ve addre	ess —		00	1aam304

Figure 21. First byte following the START procedure

## 7.4.4.7 I<sup>2</sup>C-register write access

To write data from the host controller via  $I^2C$  to a specific register of the CLRC663, the following frame format shall be used.

The read/write bit shall be set to logic 0.

The first byte of a frame indicates the device address according to the I<sup>2</sup>C rules. The second byte indicates the register address followed by up to n-data bytes. In case the address indicates the FIFO, in one frame all n-data bytes are written to the FIFO register address. This enables for example a fast FIFO access.

## 7.4.4.8 I<sup>2</sup>C-register read access

To read out data from a specific register address of the CLRC663, the host controller shall use the procedure:

First a write access to the specific register address has to be performed as indicated in the following frame:

The first byte of a frame indicates the device address according to the  $I^2C$  rules. The second byte indicates the register address. No data bytes are added.

The read/write bit shall be logic 0.

Having performed this write access, the read access starts. The host sends the device address of the CLRC663. As an answer to this device address, the CLRC663 responds with the content of the addressed register. In one frame n-data bytes could be read using the same register address. The address pointing to the register is incremented automatically (exception: FIFO register address is not incremented automatically). This enables a fast transfer of register content. The address pointer is incremented automatically and data is read from the locations [address], [address+1], [address+2]... [address+(n-1)]

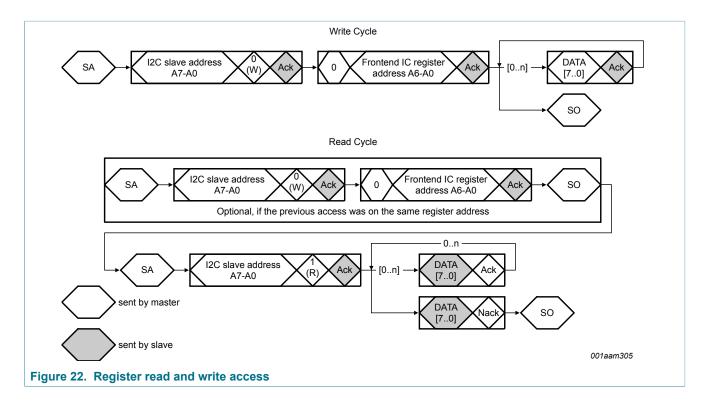
In order to support a fast FIFO data transfer, the address pointer is not incremented automatically in case the address is pointing to the FIFO.

The read/write bit shall be set to logic 1.

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### 7.4.4.9 I<sup>2</sup>CL-bus interface

The CLRC663 provides an additional interface option for connection of a SAM. This logical interface fulfills the  $I^2C$  specification, but the rise/fall timings will not be compliant to the  $I^2C$  standard. The  $I^2CL$  interface uses standard I/O pads, and the communication speed is limited to 5 MBaud. The protocol itself is equivalent to the fast mode protocol of  $I^2C$ . The SCL levels are generated by the host in push/pull mode. The RC663 does not stretch the clock. During the high period of SCL, the status of the line is maintained by a bus keeper.

The address is 01010xxb, where the last two bits of the address can be defined by the application. The definition of these bits can be done by two options. With a pin, where the higher bit is fixed to 0 or the configuration can be defined via EEPROM. Refer to the EEPROM configuration in Section 7.7.

Parameter	Min	Max	Unit
f <sub>SCL</sub>	0	5	MHz
t <sub>HD;STA</sub>	80	-	ns
t <sub>LOW</sub>	100	-	ns
t <sub>HIGH</sub>	100	-	ns
t <sub>SU;SDA</sub>	80	-	ns
t <sub>HD;DAT</sub>	0	50	ns
t <sub>SU;DAT</sub>	0	20	ns
t <sub>SU;STO</sub>	80	-	ns
t <sub>BUF</sub>	200	-	ns

#### Table 25. Timing parameter I<sup>2</sup>CL

The pull-up resistor is not required for the  $I^2CL$  interface. Instead, a on chip buskeeper is implemented in the CLRC663 for SDA of the  $I^2CL$  interface. This protocol is intended to be used for a point-to-point connection of devices over a short distance and does not support a bus capability. The driver of the pin must force the line to the desired logic voltage. To avoid that two drivers are pushing, the line at the same time following regulations must be fulfilled:

SCL: As there is no clock stretching, the SCL is always under control of the Master.

SDA: The SDA line is shared between master and slave. Therefore the master and the slave must have the control over the own driver enable line of the SDA pin. The following rules must be followed:

- In the idle phase, the SDA line is driven high by the master
- In the time between start and stop condition, the SDA line is driven by master or slave when SCL is low. If SCL is high, the SDA line is not driven by any device
- To keep the value on the SDA line a on chip, buskeeper structure is implemented for the line

## 7.4.5 SAM interface

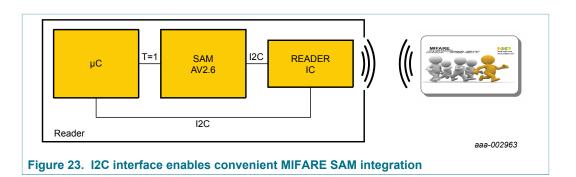
#### 7.4.5.1 SAM functionality

The CLRC663 implements a dedicated I2C or SPI interface to integrate a MIFARE SAM (Secure Access Module) in a very convenient way into applications (e.g. a proximity reader).

The SAM can be connected to the microcontroller to operate like a cryptographic coprocessor. For any cryptographic task, the microcontroller requests an operation from the SAM, receives the answer and sends it over a host interface (e.g. I2C, SPI) interface to the connected reader IC.

The MIFARE SAM supports an optimized method to integrate the SAM in a very efficient way to reduce the protocol overhead. In this system configuration, the SAM is integrated between the microprocessor and the reader IC, connected by one interface to the reader IC and by another interface to the microcontroller. In this application, the microcontroller accesses the SAM using the T=1 protocol and the SAM accesses the reader IC using an I2C interface. The I2C SAM address is always defined by EEPROM register. Default value is 0101100. As the SAM is directly communicating with reader IC, the communication overhead is reduced. In this configuration, a performance boost of up to 40 % can be achieved for a transaction time.

The MIFARE SAM supports applications using MIFARE product-based cards. For multiapplication purposes, an architecture connecting the microcontroller additionally directly to the reader IC is recommended. This is possible by connecting the CLRC663 on one interface (SAM Interface SDA, SCL) with the MIFARE SAM AV2.6 (P5DF081XX/ T1AR1070) and by connecting the microcontroller to the S2C or SPI interface.



#### 7.4.5.2 SAM connection

The CLRC663 provides an interface to connect a SAM dedicated to the CLRC663. Both interface options of the CLRC663, I<sup>2</sup>C, I<sup>2</sup>CL or SPI can be used for this purpose. The interface option of the SAM itself is configured by a host command sent from the host to the SAM.

The I<sup>2</sup>CL interface is intended to be used as connection between two ICs over a short distance. The protocol fulfills the I<sup>2</sup>C specification, but does support a single device connected to the bus only.

The SPI block for SAM connection is identical with the SPI host interface block.

The pins used for the SAM SPI are described in the following table:

SPI functionality	PIN			
MISO	SDA2			
SCL	SCL2			
MOSI	IFSEL1			
NSS	IFSEL0			

#### Table 26. SPI SAM connection

## 7.4.6 Boundary scan interface

The CLRC663 provides a boundary scan interface according to the IEEE 1149.1. This interface allows testing interconnections without using physical test probes. This is done by test cells, assigned to each pin, which override the functionality of this pin.

To be able to program the test cells, the following commands are supported:

Value (decimal)	Command	Parameter in	Parameter out
0	bypass	-	-
1	preload	data (24)	-
2	sample	-	data (24)
3	ID code (default)	-	data (32)
4	USER code	-	data (32)
5	Clamp	-	-

Value (decimal)	Command	Parameter in	Parameter out
6	HIGH Z	-	-
7	extest	data (24)	data (24)
8	interface on/off	interface (1)	-
9	register access read	address (7)	data (8)
10	register access write	address (7) - data (8)	-

The Standard IEEE 1149.1 describes the four basic blocks necessary to use this interface: Test Access Port (TAP), TAP controller, TAP instruction register, TAP data register;

#### 7.4.6.1 Interface signals

The boundary scan interface implements a four line interface between the chip and the environment. There are three Inputs: Test Clock (TCK); Test Mode Select (TMS); Test Data Input (TDI) and one output Test Data Output (TDO). TCK and TMS are broadcast signals, TDI to TDO generate a serial line called Scan path.

Advantage of this technique is that independent of the numbers of boundary scan devices the complete path can be handled with four signal lines.

The signals TCK, TMS are directly connected with the boundary scan controller. Because these signals are responsible for the mode of the chip, all boundary scan devices in one scan path will be in the same boundary scan mode.

#### 7.4.6.2 Test Clock (TCK)

The TCK pin is the input clock for the module. If this clock is provided, the test logic is able to operate independent of any other system clocks. In addition, it ensures that multiple boundary scan controllers that are daisy-chained together can synchronously communicate serial test data between components. During normal operation, TCK is driven by a free-running clock. When necessary, TCK can be stopped at 0 or 1 for extended periods of time. While TCK is stopped at 0 or 1, the state of the boundary scan controller does not change and data in the Instruction and Data Registers is not lost.

The internal pull-up resistor on the TCK pin is enabled. This assures that no clocking occurs if the pin is not driven from an external source.

#### 7.4.6.3 Test Mode Select (TMS)

The TMS pin selects the next state of the boundary scan controller. TMS is sampled on the rising edge of TCK. Depending on the current boundary scan state and the sampled value of TMS, the next state is entered. Because the TMS pin is sampled on the rising edge of TCK, the IEEE Standard 1149.1 expects the value on TMS to change on the falling edge of TCK.

Holding TMS high for five consecutive TCK cycles drives the boundary scan controller state machine to the Test-Logic-Reset state. When the boundary scan controller enters the Test-Logic-Reset state, the Instruction Register (IR) resets to the default instruction, IDCODE. Therefore, this sequence can be used as a reset mechanism.

The internal pull-up resistor on the TMS pin is enabled.

#### 7.4.6.4 Test Data Input (TDI)

The TDI pin provides a stream of serial information to the IR chain and the DR chains. TDI is sampled on the rising edge of TCK and, depending on the current TAP state and the current instruction, presents this data to the proper shift register chain. Because the TDI pin is sampled on the rising edge of TCK, the IEEE Standard 1149.1 expects the value on TDI to change on the falling edge of TCK.

The internal pull-up resistor on the TDI pin is enabled.

#### 7.4.6.5 Test Data Output (TDO)

The TDO pin provides an output stream of serial information from the IR chain or the DR chains. The value of TDO depends on the current TAP state, the current instruction, and the data in the chain being accessed. In order to save power when the port is not being used, the TDO pin is placed in an inactive drive state when not actively shifting out data. Because TDO can be connected to the TDI of another controller in a daisy-chain configuration, the IEEE Standard 1149.1 expects the value on TDO to change on the falling edge of TCK.

#### 7.4.6.6 Data register

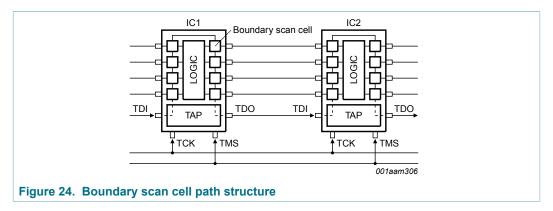
According to the IEEE1149.1 standard, there are two types of data register defined: bypass and boundary scan

The bypass register enable the possibility to bypass a device when part of the scan path.Serial data is allowed to be transferred through a device from the TDI pin to the TDO pin without affecting the operation of the device.

The boundary scan register is the scan-chain of the boundary cells. The size of this register is dependent on the command.

#### 7.4.6.7 Boundary scan cell

The boundary scan cell opens the possibility to control a hardware pin independent of its normal use case. Basically the cell can only do one of the following: control, output and input.



#### 7.4.6.8 Boundary scan path

This chapter shows the boundary scan path of the CLRC663.

Table 28. Boundary sc Number (decimal)	Cell	Port	Function
23			Control
	BC_1	-	
22	BC_8	CLKOUT	Bidir
21	BC_1	-	Control
20	BC_8	SCL2	Bidir
19	BC_1	-	Control
18	BC_8	SDA2	Bidir
17	BC_1	-	Control
16	BC_8	IFSEL0	Bidir
15	BC_1	-	Control
14	BC_8	IFSEL1	Bidir
13	BC_1	-	Control
12	BC_8	IF0	Bidir
11	BC_1	-	Control
10	BC_8	IF1	Bidir
9	BC_1	-	Control
8	BC_8	IF2	Bidir
7	BC_1	IF2	Output2
6	BC_4	IF3	Bidir
5	BC_1	-	Control
4	BC_8	IRQ	Bidir
3	BC_1	-	Control
2	BC_8	SIGIN	Bidir
1	BC_1	-	Control
	—		

Table 28. Boundary scan path of the CLRC663

Refer to the CLRC663 BSDL file.

#### 7.4.6.9 Boundary Scan Description Language (BSDL)

All of the boundary scan devices have a unique boundary structure which is necessary to know for operating the device. Important components of this language are:

- available test bus signal
- compliance pins
- · command register
- · data register
- boundary scan structure (number and types of the cells, their function and the connection to the pins.)

The CLRC663 is using the cell BC\_8 for the IO-Lines. The  $I^2C$  Pin is using a BC\_4 cell. For all pad enable lines, the cell BC1 is used.

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The manufacturer's identification is 02Bh.

- attribute IDCODEISTER of CLRC663: entity is "0001" and -- version
- "0011110010000010b" and -- part number (3C82h)
- "00000010101b" and -- manufacturer (02Bh)
- "1b"; -- mandatory

The user code data is coded as followed:

- product ID (3 bytes)
- version

These four bytes are stored as the first four bytes in the EEPROM.

#### 7.4.6.10 Non-IEEE1149.1 commands

#### Interface on/off

With this command, the host/SAM interface can be deactivated and the Read and Write command of the boundary scan interface is activated. (Data = 1). With Update-DR, the value is taken over.

#### Register Access Read

At Capture-DR, the actual address is read and stored in the DR. Shifting the DR is shifting in a new address. With Update-DR, this address is taken over into the actual address.

#### **Register Access Write**

At the Capture-DR, the address and the data is taken over from the DR. The data is copied into the internal register at the given address.

## 7.5 Buffer

#### 7.5.1 Overview

A 512 × 8-bit FIFO buffer is implemented in the CLRC663. It buffers the input and output data stream between the host and the internal state machine of the CLRC663. Thus, it is possible to handle data streams with lengths of up to 512 bytes without taking timing constraints into account. The FIFO can also be limited to a size of 255 bytes. In this case all the parameters (FIFO length, Watermark...) require a single byte only for definition. In case of a 512 byte FIFO length, the definition of this value requires 2 bytes.

#### 7.5.2 Accessing the FIFO buffer

When the  $\mu$ -Controller starts a command, the CLRC663 may, while the command is in progress, access the FIFO-buffer according to that command. Physically only one FIFO-buffer is implemented, which can be used in input and output direction. Therefore the  $\mu$ -Controller has to take care, not to access the FIFO buffer in a way that corrupts the FIFO data.

## 7.5.3 Controlling the FIFO buffer

Besides writing to and reading from the FIFO buffer, the FIFO-buffer pointers might be reset by setting the bit FIFOFlush in FIFOControl to 1. Consequently, the FIFOLevel bits are set to logic 0, the actually stored bytes are not accessible any more and the FIFO buffer can be filled with another 512 bytes (or 255 bytes if the bit FIFOSize is set to 1) again.

#### 7.5.4 Status Information about the FIFO buffer

The host may obtain the following data about the FIFO-buffers status:

- Number of bytes already stored in the FIFO-buffer. Writing increments, reading decrements the FIFO level: FIFOLength in register FIFOLength (and FIFOControl Register in 512 byte mode)
- Warning, that the FIFO-buffer is almost full: HiAlert in register FIFOControl according to the value of the water level in register WaterLevel (Register 02h bit [2], Register 03h bit[7:0])
- Warning, that the FIFO-buffer is almost empty: LoAlert in register FIFOControl according to the value of the water level in register WaterLevel (Register 02h bit [2], Register 03h bit[7:0])
- FIFOOvI bit indicates, that bytes were written to the FIFO buffer although it was already full: ErrIRQ in register IRQ0.

WaterLevel is one single value defining both HiAlert (counting from the FIFO top) and LoAlert (counting from the FIFO bottom). The CLRC663 can generate an interrupt signal if:

- LoAlertIRQEn in register IRQ0En is set to logic 1 it will activate pin IRQ when LoAlert in the register FIFOControl changes to 1.
- HiAlertIRQEN in register IRQ0En is set to logic 1 it will activate pin IRQ when HiAlert in the register FIFOControl changes to 1.

The bit HiAlert is set to logic 1 if maximum water level bytes (as set in register WaterLevel) or less can be stored in the FIFO-buffer. It is generated according to the following equation:

$$HiAlert = (FiFoSize + - FiFoLength) \leq WaterLevel$$

(2)

The bit LoAlert is set to logic 1 if water level bytes (as set in register WaterLevel) or less are actually stored in the FIFO-buffer. It is generated according to the following equation:

$$LoAlert = FIFOLength \le WaterLevel$$

(3)

## 7.6 Analog interface and contactless UART

#### 7.6.1 General

The integrated contactless UART supports the external host online with framing and error checking of the protocol requirements up to 848 kbit/s. An external circuit can be connected to the communication interface pins SIGIN and SIGOUT to modulate and demodulate the data.

The contactless UART handles the protocol requirements for the communication schemes in co-operation with the host. The protocol handling itself generates bit- and byte-oriented framing and handles error detection like Parity and CRC according to the different contactless communication schemes.

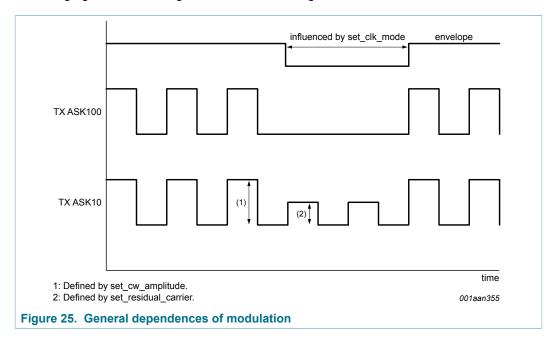
The size, the tuning of the antenna, and the supply voltage of the output drivers have an impact on the achievable field strength. The operating distance between reader and card depends additionally on the type of card used.

#### 7.6.2 TX transmitter

The signal delivered on pin TX1 and pin TX2 is the 13.56 MHz carrier modulated by an envelope signal for energy and data transmission. It can be used to drive an antenna directly, using a few passive components for matching and filtering, see <u>Section 13</u>. The signal on TX1 and TX2 can be configured by the register DrvMode, see <u>Section 8.8.1</u>.

The modulation index can be set by the TxAmp.

Following figure shows the general relations during modulation



Note: When changing the continuous carrier amplitude, the residual carrier amplitude also changes, while the modulation index remains the same.

The registers <u>Section 8.8</u> and <u>Section 8.10</u> control the data rate, the framing during transmission and the setting of the antenna driver to support the requirements at the different specified modes and transfer speeds.

#### Table 29. Settings for TX1 and TX2

TxClkMode (binary)	Tx1 and TX2 output	Remarks
000	High impedance	-
001	0	output pulled to 0 in any case
010	1	output pulled to 1 in any case
110	RF high side push	open-drain, only high side (push) MOS supplied with clock, clock parity defined by invtx; low side MOS is off
101	RF low side pull	open-drain, only low side (pull) MOS supplied with clock, clock parity defined by invtx; high side MOS is off
111	13.56 MHz clock derived from 27.12 MHz quartz divided by 2	push/pull Operation, clock polarity defined by invtx; setting for 10 % modulation

Register TXamp and the bits for set\_residual\_carrier define the modulation index:

set_residual_carrier (decimal)	residual carrier [%]	modulation index [%]
0	99	0.5
1	98	1.0
2	96	2.0
3	94	3.1
4	91	4.7
5	89	5.8
6	87	7.0
7	86	7.5
8	85	8.1
9	84	8.7
10	83	9.3
11	82	9.9
12	81	10.5
13	80	11.1
14	79	11.7
15	78	12.4
16	77	13.0
17	76	13.6
18	75	14.3

Table 30. Setting residual carrier and modulation index by TXamp.set residual carrier

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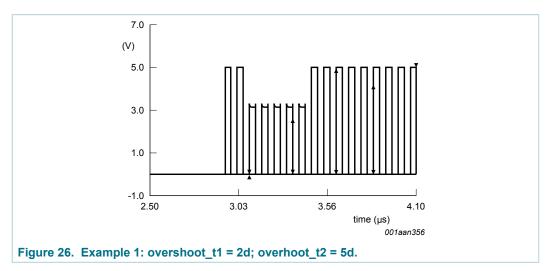
set_residual_carrier (decimal)	residual carrier [%]	modulation index [%]
19	74	14.9
20	72	16.3
21	70	17.6
22	68	19.0
23	65	21.2
24	60	25.0
25	55	29.0
26	50	33.3
27	45	37.9
28	40	42.9
29	35	48.1
30	30	53.8
31	25	60.0

Note: At VDD(TVDD) <5 V and residual carrier settings <50 %, the accuracy of the modulation index may be low in dependency of the antenna tuning impedance

#### 7.6.2.1 Overshoot protection

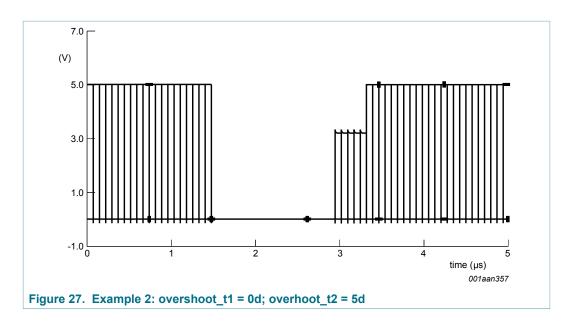
The CLRC663 provides an overshoot protection for 100 % ASK to avoid overshoots during a PCD communication. Therefore two timers overshoot\_t1 and overshoot\_t2 can be used.

During the timer overshoot\_t1 runs an amplitude defined by set\_cw\_amplitude bits is provided to the output driver. Followed by an amplitude denoted by set\_residual\_carrier bits with the duration of overshoot\_t2.



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#### 7.6.2.2 Bit generator

The default coding of a data stream is done by using the Bit-Generator. It is activated when the value of TxFrameCon.DCodeType is set to 0000 (bin). The Bit-Generator encodes the data stream byte-wise and can apply the following encoding steps to each data byte.

- 1. Add a start-bit of specified type at beginning of every byte
- 2. Add a stop-bit and EGT bits of a specified type. The maximum number of EGT bit is 6, only full bits are supported
- 3. Add a parity-bit of a specified type
- 4. TxLastBits (skips a given number of bits at the end of the last byte in a frame)
- 5. Encrypt data-bit (MIFARE Classic encryption)

It is not possible to skip more than 8 bit of a single byte!

By default, data bytes are always treated LSB first.

#### 7.6.3 Receiver circuitry

#### 7.6.3.1 General

The CLRC663 features a versatile quadrature receiver architecture with fully differential signal input at RXP and RXN. It can be configured to achieve optimum performance for reception of various 13.56 MHz based protocols.

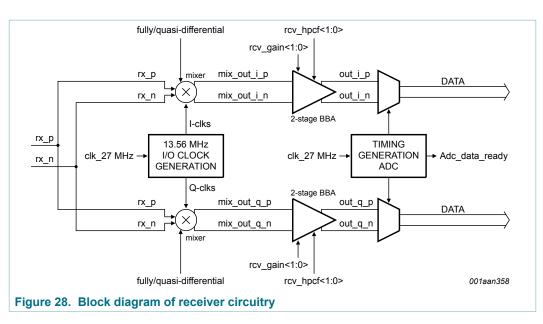
For all processing units various adjustments can be made to obtain optimum performance.

#### 7.6.3.2 Block diagram

The following figure shows the block diagram of the receiver circuitry. The receiving process includes several steps. First the quadrature demodulation of the carrier signal of 13.56 MHz is done. Several tuning steps in this circuit are possible.

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The receiver can also be operated in a single ended mode. In this case, the Rcv\_RX\_single bit has to be set. In the single ended mode, the two receiver pins RXP and RXN need to be connected together and will provide a single ended signal to the receiver circuitry.

When using the receiver in a single ended mode, the receiver sensitivity is decreased and the achievable reading distance might be reduced, compared to the fully differential mode.

Mode	rcv_rx_single	pins RXP and RXN
Fully differential	0	provide differential signal from differential antenna by separate rx-coupling branches
Quasi differential	1	connect RXP and RXN together and provide single ended signal from antenna by a single rx-coupling branch

#### Table 31. Configuration for single or differential receiver

The quadrature-demodulator uses two different clocks, Q-clock and I-clock, with a phase shift of 90° between them. Both resulting baseband signals are amplified, filtered, digitized and forwarded to a correlation circuitry.

The typical application is intended to implement the Fully differential mode and will deliver maximum reader/writer distance. The Quasi differential mode can be used together with dedicated antenna topologies that allow a reduction of matching components at the cost of overall reading performance.

During low-power card detection the DC levels at the I- and Q-channel mixer outputs are evaluated. This requires that mixers are directly connected to the ADC. This can be configured by setting the bit Rx\_ADCmode in register Rcv (38h).

#### 7.6.4 Active antenna concept

Two main blocks are implemented in the CLRC663. A digital circuitry, comprising state machines, coder and decoder logic and an analog circuitry with the modulator and antenna drivers, receiver and amplification circuitry. For example, the interface between

these two blocks can be configured in the way, that the interfacing signals may be routed to the pins SIGIN and SIGOUT. The most important use of this topology is the active antenna concept where the digital and the analog blocks are separated. This opens the possibility to connect e.g. an additional digital block of another CLRC663 device with a single analog antenna frontend.

	SIGIN	SIGOUT	
READER IC (DIGITAL)	SIGOUT	SIGIN	READER IC (ANTENNA)
			001aam307

Figure 29. Block diagram of the active Antenna concept

The <u>Table 32</u> and <u>Table 33</u> describe the necessary register configuration for the use case active antenna concept.

Register	Value (binary)	Description
SigOut.SigOutSel	0100	TxEnvelope
Rcv.SigInSel	10 11	Receive over SigIn (ISO/IEC14443A) Receive over SigIn (Generic Code)
DrvCon.TxSel	00	Low (idle)

#### Table 32. Register configuration of CLRC663 active antenna concept (DIGITAL)

Table 33	<b>Register configuratio</b>	n of CLRC663 active antenna	concept (Antenna)
----------	------------------------------	-----------------------------	-------------------

Register	Value (binary)	Description
SigOut.SigOutSel	0110 0111	Generic Code (Manchester) Manchester with Subcarrier (ISO/IEC14443A)
Rcv.SigInSel	01	Internal
DrvCon.TxSel	10	External (SigIn)
RxCtrl.RxMultiple	1	RxMultiple on

The interface between these two blocks can be configured in the way, that the interfacing signals may be routed to the pins SIGIN and SIGOUT (see <u>Figure 30</u>).

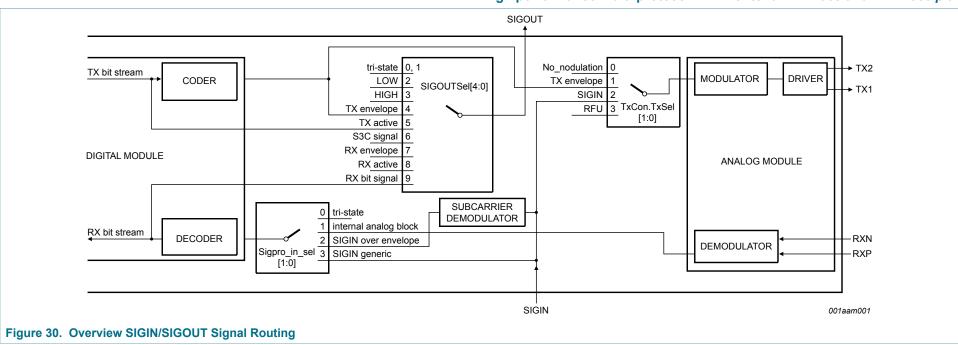
This topology supports, that some parts of the analog part of the CLRC663 may be connected to the digital part of another device.

The switch SigOutSel in registerSigOut can be used to measure signals. This is especially important during the design-in phase or for test purposes to check the transmitted and received data.

However, the most important use of SIGIN/SIGOUT pins is the active antenna concept. An external active antenna circuit can be connected to the digital circuit of the CLRC663. SigOutSel has to be configured in that way that the signal of the internal Miller Coder is sent to SIGOUT pin (SigOutSel = 4). SigInSel has to be configured to receive Manchester signal with subcarrier from SIGIN pin (SigInSel = 1).

It is possible, to connect a passive antenna to pins TX1, TX2 and RX (via the appropriate filter and matching circuit) and at the same time an active antenna to the pins SIGOUT and SIGIN. In this configuration, two RF-parts may be driven (one after another) by a single host processor.

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#### 7.6.5 Symbol generator

The symbol generator is used to create various protocol symbols. These can be e.g. SOF or EOF symbols as they are used by the ISO14443 protocols or proprietary protocol symbols like the CS symbol as used by the ICODE EPC protocol.

Symbols are defined by means of the symbol definition registers and the mode registers. Four different symbols can be used. Two of them, Symbol0 and Symbol1 have a maximum pattern length of 16 bit and feature a burst length of up to 256 bits of either logic "0" or logic "1". The Symbol2 and Symbol3 are limited to 8-bit pattern length and do not support a burst.

The definition of symbol patterns is done by writing the bit sequence of the pattern to the appropriate register. The last bit of the pattern to be sent is located at the LSB of the register. By setting the symbol length in the symbol-length register (TxSym10Len and TxSym32Len), the definition of the symbol pattern is completed. All other bits at bit-position higher than the symbol length in the definition register are ignored. (Example: length of Symbol2 = 5, bit7 and bit6 are ignored, bit5 to bit0 define the symbol pattern, bit5 is sent first)

Which symbol-pattern is sent can be configured in the TxFrameCon register. Symbol0, Symbol1 and Symbol2 can be sent before data packets, Symbol1, Symbol2 and Symbol3 can be sent after data packets. Each symbol is defined by a set of registers. Symbols are configured by a pair of registers. Symbol0 and Symbol1 share the same configuration and Symbol2 and Symbol3 share the same configuration. The configuration includes setting of bit-clock- and subcarrier-frequency, as well as selection of the pulse type/length and the envelope type.

## 7.7 Memory

#### 7.7.1 Memory overview

The CLRC663 implements three different memories: EEPROM, FIFO and Registers.

At startup, the initialization of the registers which define the behavior of the IC is performed by an automatic copy of an EEPROM area (read/write EEPROM section1 and section2, register reset) into the registers. The behavior of the CLRC663 can be changed by executing the command LoadProtocol, which copies a selected default protocol from the EEPROM (read-only EEPROM section4, register Set Protocol area) into the registers.

The read/write EEPROM section2 can be used to store any user data or predefined register settings. These predefined settings can be copied with the command "LoadRegister" into the internal registers.

The FIFO is used as Input/Out buffer and is able to improve the performance of a system with limited interface speed.

#### 7.7.2 EEPROM memory organization

The CLRC663 has implemented a EEPROM non-volatile memory with a size of 8 kB.The EEPROM is organized in pages of 64 bytes. One page of 64 bytes can be programmed at a time. Defined purposes had been assigned to specific memory areas

of the EEPROM, which are called Sections. Five sections 0..4 with different purpose do exist.

Section	Page	Byte addresses	Access rights	Memory content
0	0	00 to 31	r	product information and configuration
		32 to 63	r/w	product configuration
1	1 to 2	64 to 191	r/w	register reset
2	3 to 95	192 to 6143	r/w	free
3	96 to 111	6144 to 7167	w	MIFARE Classic key
4	112 to 128	7168 to 8191	r	Register Set Protocol (RSP)

The following figure shows the structure of the EEPROM:

	Section 0:	Production and config
	Section 1:	Register reset
	Section 2:	Free
	Section 3:	MIFARE Classic key area (MKA)
	Section 4_TX:	RSP-Area for TX
	Section 4_RX:	RSP-Area for RX
		001aan35
Figure 31. Sector arrangen	nent of the EEF	PROM

#### 7.7.2.1 Product information and configuration - Page 0

The first EEPROM page includes production data as well as configuration information.

Table 35.	Production are	ea (Page 0)		,	i .		i.	
Address (Hex.)	0	1	2	3	4	5	6	7
00	ProductID Version Unique Identifier							
08						Manufacturer Data		
10	ManufacturerData							
18	ManufacturerData							

ProductID: Identifier for this CLRC663 product or derivative, only address 01h shall be evaluated for identifying the Product CLRC663, address 00h and 02h shall be ignored by software.

Please note, that the silicon version CLRC66301, CLRC66302, CLRC66303 and derivatives can be identified on register address 7Fh, it is not coded in the EEPROM production area.

#### Table 36. Product ID overview of CLRC663 family

Address 01h	Product ID
CLRC663	01h
MFRC631	C0h
MFRC630	80h
SLRC610	20h

Version: This register indicates the version of the EEPROM initialization data during production.

Unique IDentifier: Unique serial number code for this device

Manufacturer Data: This data is programmed during production. The content is not intended to be used by any application and might not be constant for different devices. Therefore the content needs to be considered to be undefined.

#### Table 37. Configuration area (Page 0)

Address (Hex.)	0	1	2	3	4	5	6	7
20	I <sup>2</sup> C_Address	Interface	I <sup>2</sup> C SAM_Address	DefaultProtRx	DefaultProtTx	-	TxCRCPrese	et
28	RxCRCPreset		-	-	-	-	-	-
30	-							
38	-							

### I<sup>2</sup>C-Address

Two possibilities exist to define the address of the  $I^2C$  interface. This can be done either by configuring the pins IF0, IF2 (address is then 10101xx, xx is defined by the interface pins IF0, IF2) or by writing value into the  $I^2C$  address area. The selection, which of this 2-information pin configuration or EEPROM content - is used as  $I^2C$ -address is done at EEPROM address 21h (Interface, bit4)

#### Interface

This section describes the interface byte configuration.

#### Table 38. Interface byte

Bit	7	6	5	4	3	2 1		0
	I <sup>2</sup> C_HSP	-	-	I2C_Address	Boundary Scan	Host		
access rights	r/w	RFU	RFU	r/w	r/w	-	-	-

Bit	Symbol	Description
7	I <sup>2</sup> C_HSP	when cleared, the high-speed mode is used when set, the high speed+ mode is used (default)
6, 5	RFU	-
4	I <sup>2</sup> C_Address	when cleared, the pins are used (default) when set, the EEPROM is used
3	Boundary Scan	when cleared, the boundary scan interface is ON (default) when set, the boundary scan is OFF
2 to 0	Host	000b - RS232 001b - $I^{2}C$ 010b - SPI 011b - $I^{2}CL$ 1xxb - pin selection

## I<sup>2</sup>C\_SAM\_Address

The I<sup>2</sup>C SAM Address is always defined by the EEPROM content.

The Register Set Protocol (RSP) Area contains settings for the TX registers (16 bytes) and for the RX registers (8 bytes).

Section								
Section 4 TX	Tx0		Tx1		TX2		Tx3	
Section 4 TX	Tx4		Tx5		TX6		ТХ7	
Section 4 Rx	RX0	RX1	RX2	RX3	RX4	RX5	RX6	RX7
Section 4 Rx	RX8	RX9	RX10	RX11	RX12	RX13	RX14	RX15

#### **TxCrcPreset**

The data bits are sent by the analog module and are automatically extended by a CRC.

#### 7.7.3 EEPROM initialization content LoadProtocol

The CLRC663 EEPROM is initialized at production with values which are used to reset certain registers of the CLRC663 to default settings by copying the EEPROM content to the registers. Only registers or bits with "read/write" or "dynamic" access rights are initialized with this default values copied from the EEPROM.

Note that the addresses used for copying reset values from EEPROM to registers are dependent on the configured protocol and can be changed by the user.

Address	0 (8)	1 (9)	2 (A)	3 (B)	4 (C)	5 (D)	6 (E)	7 (F)
Function	n Product ID			Version	Unique Identi	fier		

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CLRC663

Address	0 (8)	1 (9)	2 (A)	3 (B)	4 (C)	5 (D)	6 (E)	7 (F)
00	XX	see table 34	ХХ	XX	XX	XX	XX	XX
Function	Unique Iden	tifier	1	I				Factory trim value
08	XX	XX	XX	XX	XX	XX	XX	XX
Function	TrimLFO Factory trim values							
10	XX	XX	XX	XX	XX	XX	XX	XX
Function	Factory trim	values		1			I	
18	XX	XX	XX	XX	XX	XX	XX	XX
	Factory trim	values	1	i	i	i	1	
38	XX	XX	XX	XX	XX	XX	XX	XX

The register reset values are configuration parameters used after startup of the IC. They can be changed to modify the default behavior of the device. In addition to these register reset values, is the possibility to load settings for various users implemented protocols.The load protocol command is used for this purpose.

#### Table 42. Register reset values (Hex.)(Page1 and page 2)

	rtogiotor rooo		./(i ugo i unu	pago 2)				
Address	0 (8)	1 (9)	2 (A)	3 (B)	4 (C)	5 (D)	6 (E)	7 (F)
	Command	HostCtrl	FiFoControl	WaterLevel	FiFoLength	FiFoData	IRQ0	IRQ1
40	40	00	80	05	00	00	00	00
	IRQ0En	IRQ1En	Error	Status	RxBitCtrl	RxColl	TControl	T0Control
48	10	00	00	00	00	00	00	00
	T0ReloadHi	T0ReloadLo	T0Counter ValHi	T0Counter ValLo	T1Control	T1ReloadHi	T1ReloadLo	T1Counter ValHi
50	00	80	00	00	00	00	80	00
	T1Counter ValLo	T2Control	T2ReloadHi	T2ReloadLo	T2Counter ValHi	T2Counter ValLo	T3Control	T3ReloadHi
58	00	00	00	80	00	00	00	00
	T3ReloadLo	T3Counter ValHi	T3Counter ValHi	T4Control	T4ReloadHi	T4ReloadLo	T4Counter ValHi	T4Counter ValLo
60	80	00	00	00	00	80	00	00
	DrvMode	TxAmp	DrvCon	Txl	TxCRC Preset	RxCRC Preset	TxDataNum	TxModWith
68	86	15	11	06	18	18	08	27
	TxSym10 BurstLen	TxWaitCtrl	TxWaitLo	FrameCon	RxSofD	RxCtrl	RxWait	RxThres hold
70	00	C0	12	CF	00	04	90	3F
	Rcv	RxAna	RFU	SerialSpeed	LFO_trimm	PLL_Ctrl	PLL_Div	LPCD_QMin
78	12	0A	00	7A	80	04	20	48
	1	1	1	1	1	1	1	

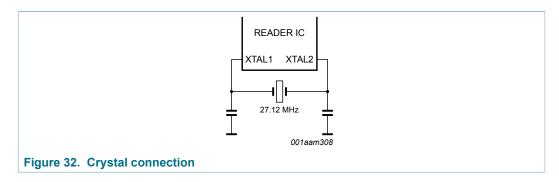
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Address	0 (8)	1 (9)	2 (A)	3 (B)	4 (C)	5 (D)	6 (E)	7 (F)
	LPCD_ QMax	LPCD_IMin	LPCD _result_I	LPCD _result_Q			PadIn	SigOut
80	12	88	00	00	00	00	00	00
	TxBitMod	RFU	TxDataCon	TxDataMod	TxSymFreq	TxSym0H	TySym0L	TxSym1H
88	20	хх	04	50	40	00	00	00
	TxSym1L	TxSym2	TxSym3	TxSym10Le ngth	TxSym32Le ngth	TxSym32Bu rstCtrl	TxSym10M od	TxSym32M od
90	0x00	0x00	0x00	0x00	0x00	0x00	0x00	0x50
	RxBitMod	RxEOFSym	RxSyncValH	RxSyncValL	RxSyncMod	RxMod	RXCorr	FabCal
98	0x02	0x00	0x00	0x01	0x00	0x08	0x08	0xB2

### 7.8 Clock generation

### 7.8.1 Crystal oscillator

The clock applied to the CLRC663 acts as time basis for generation of the carrier sent out at TX and for the quadrature mixer I and Q clock generation as well as for the coder and decoder of the synchronous system. Therefore stability of the clock frequency is an important factor for proper performance. To obtain highest performance, clock jitter has to be as small as possible. This is best achieved by using the internal oscillator buffer with the recommended circuitry.



#### Table 43. Crystal requirements recommendations

Symbol	Parameter	Conditions	Min	Тур	max	Unit
f <sub>xtal</sub>	crystal frequency		-	27.12	-	MHz
$\Delta f_{xtal}/f_{xtal}$	relative crystal frequency variation		-250	-	+250	ppm
ESR	equivalent series resistance		-	50	100	Ω
CL	load capacitance		-	10	-	pF
P <sub>xtal</sub>	crystal power dissipation		-	50	100	μW

### 7.8.2 IntegerN PLL clock line

The CLRC663 is able to provide a clock with configurable frequency at CLKOUT from 1 MHz to 24 MHz (PLL\_Ctrl and PLL\_DIV). There it can serve as a clock source to a microcontroller which avoids the need of a second crystal oscillator in the reader system. Clock source for the IntegerN-PLL is the 27.12 MHz crystal oscillator.

Two dividers are determining the output frequency. First a feedback integer-N divider configures the VCO frequency to be N × fin/2 (control signal pll\_set\_divfb). As supported Feedback Divider Ratios are 23, 27 and 28, VCO frequencies can be  $23 \times fin / 2$  (312 MHz),  $27 \times fin / 2$  (366 MHz) and  $28 \times fin / 2$  (380 MHz).

The VCO frequency is divided by a factor which is defined by the output divider (pll\_set\_divout). The following table shows the accuracy achieved for various frequencies (integer multiples of 1 MHz and some typical RS232 frequencies) and the divider ratios to be used. The register bit ClkOutEn enables the clock at CLKOUT pin.

The following formula can be used to calculate the output frequency:

f<sub>out</sub> = 13.56 MHz × PLLDiv\_FB /PLLDiv\_Out

Frequency [MHz]	4	6	8	10	12	20	24	1.8432	3.6864
PLLDiv_FB	23	27	23	28	23	28	23	28	28
PLLDiv_Out	78	61	39	38	26	19	16	206	103
accuracy [%]	0.04	0.03	0.04	0.08	0.04	0.08	0.04	0.01	0.01

#### Table 44. Divider values for selected frequencies using the integerN PLL

#### 7.8.3 Low Frequency Oscillator (LFO)

The Low-Frequency (LFO) is implemented to drive a wake-up counter (WUC). This wakes up the system in regular time intervals and eases the design of a reader that is regularly polling for card presence or implements a low-power card detection.

The LFO is trimmed during production to run at 16 kHz. Unless a high accuracy of the LFO is required by the application and the device is operated in an environment with changing ambient temperatures, trimming of the LFO is not required. For a typical application making use of the LFO for wake-up from power down, the trim value set during production can be used. Optional trimming to achieve a higher accuracy of the 16 kHz LFO clock is supported by a digital state machine which compares LFO-clock to a reference clock. As reference clockfrequency the 13.56 MHz crystal clock is available.

## 7.9 Power management

#### 7.9.1 Supply concept

The CLRC663 is supplied by  $V_{DD}$  (Supply Voltage), PVDD (Pad Supply) and TVDD (Transmitter Power Supply). These three voltages are independent from each other.

To connect the CLRC663 to a Microcontroller supplied by 3.3 V, PVDD and  $V_{DD}$  shall be at a level of 3.3 V, TVDD can be in a range from 3.3 V to 5.0 V. A higher supply voltage at TVDD results in a higher field strength.

Independent of the voltage it is recommended to buffer these supplies with blocking capacitances close to the terminals of the package.  $V_{DD}$  and PVDD are recommended to be blocked with a capacitor of 100 nF min, TVDD is recommended to be blocked with 2 capacitors, 100 nF parallel to 1.0  $\mu$ F

AVDD and DVDD are not supplied input pins. They are output pins and shall be connected to blocking capacitors 470 nF each.

#### 7.9.2 Power reduction mode

#### 7.9.2.1 Power-down

A hard power-down is enabled with HIGH level on pin PDOWN. This turns off the internal 1.8 V voltage regulators for the analog and digital core supply as well as the oscillator. All digital input buffers are separated from the input pads and clamped internally (except pin PDOWN itself). The output pins are switched to high impedance. HardPowerDown is performing a reset of the IC. All registers will be reset, the Fifo will be cleared.

To leave the power-down mode the level at the pin PDOWN as to be set to LOW. This starts the internal start-up sequence.

#### 7.9.2.2 Standby mode

The standby mode is entered immediately after setting the bit PowerDown in the register Command. All internal current sinks are switched off. Voltage references and voltage regulators are set into standby mode.

In opposition to the power-down mode, the digital input buffers are not separated by the input pads and keep their functionality. The digital output pins do not change their state.

During standby mode, all registers values, the FIFO's content and the configuration itself keeps its current content.

To leave the standby mode, the bit PowerDown in the register Command is cleared. This triggers the internal start-up sequence. The reader IC is in full operation mode again when the internal start-up sequence is finalized.

A value of 55h must be sent to the CLRC663 using the RS232 interface to leave the standby mode. This is must at RS232, but cannot be used for the  $I^2C/SPI$  interface. Then read accesses shall be performed at address 00h until the device returns the content of this address. The return of the content of address 00h indicates that the device is ready to receive further commands and the internal start-up sequence is finalized.

#### 7.9.2.3 Modem off mode

When the ModemOff bit in the register Control is set the antenna transmitter and the receiver are switched off.

To leave the modem off mode, clears the ModemOff bit in the register Control.

#### 7.9.3 Low-Power Card Detection (LPCD)

The low-power card detection is an energy saving mode in which the CLRC663 is not fully powered permanently.

The LPCD works in two phases. First the standby phase is controlled by the wake-up counter (WUC), which defines the duration of the standby of the CLRC663. Second phase is the detection-phase. In this phase, the values of the I and Q channel are detected and stored in the register map. (LPCD\_I\_Result, LPCD\_Q\_Result). This time period can be handled with Timer3. The value is compared with the min/max values in the registers (LPCD\_IMin, LPCD\_IMax; LPCD\_QMin, LPCD\_QMax). If it exceeds the limits, an LPCDIRQ is raised.

After the command LPCD the standby of the CLRC663 is activated, if selected. The wake-up Timer4 can activate the system after a given time. For the LPCD, it is recommended to set T4AutoWakeUp and T4AutoRestart, to start the timer and then go to standby. If a card is detected, the communication can be started. If T4AutoWakeUp is not set, the IC will not enter Standby mode in case no card is detected.

#### 7.9.4 Reset and start-up time

A 10 µs constant high level at the PDOWN pin starts the internal reset procedure.

The following figure shows the internal voltage regulator:

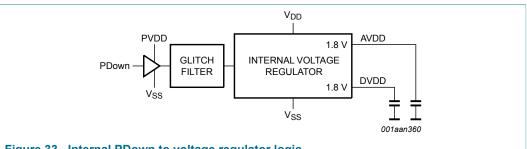


Figure 33. Internal PDown to voltage regulator logic

When the CLRC663 has finished, the reset phase and the oscillator has entered a stable working condition the IC is ready to be used.

## 7.10 Command set

#### 7.10.1 General

The behavior is determined by a state machine capable to perform a certain set of commands. By writing a command-code to the command register, the command is executed.

Arguments and/or data necessary to process a command, are exchanged via the FIFO buffer.

- Each command that needs a certain number of arguments will start processing only when it has received the correct number of arguments via the FIFO buffer.
- The FIFO buffer is not cleared automatically at command start. It is recommended to write the command arguments and/or the data bytes into the FIFO buffer and start the command afterwards.
- Each command may be stopped by the host by writing a new command code into the command register e.g.: the Idle-Command.

#### 7.10.2 Command set overview

#### Table 45. Command set Command Parameter (bytes) **Short description** No. Idle 00h no action. cancels current command execution LPCD 01h low-power card detection LoadKey 02h (keybyte1),(keybyte2), (keybyte3), reads a MIFARE Classic key (size of 6 bytes) from (keybyte4), (keybyte5),(keybyte6); FIFO buffer ant puts it into Key buffer MFAuthent 03h 60h or 61h, (block address), (card performs the MIFARE Classic authentication serial number byte0),(card serial number byte1), (card serial number byte2),(card serial number byte3); performs a query, an Ack and a Req-Rn for ISO/IEC AckReg 04h \_ 18000-3 mode 3/ EPC Class-1 HF 05h Receive activates the receive circuit 06h transmits data from the FIFO buffer Transmit bytes to send: byte1, byte2,.... Transceive 07h transmits data from the FIFO buffer and automatically bytes to send: byte1, byte2,.... activates the receiver after transmission finished WriteE2 08h addressH, addressL, data; gets one byte from FIFO buffer and writes it to the internal EEPROM WriteE2Page (page Address), data0, gets up to 64 bytes (one EEPROM page) from the FIFO 09h [data1 ..data63]; buffer and writes it to the EEPROM ReadE2 0Ah addressH, address L, length; reads data from the EEPROM and copies it into the FIFO buffer LoadReg 0Ch (EEPROM addressH), (EEPROM reads data from the internal EEPROM and initializes the CLRC663 registers. EEPROM address needs to be addressL), RegAdr, (number of within FEPROM sector 2 Register to be copied); LoadProtocol 0Dh (Protocol number RX), (Protocol reads data from the internal EEPROM and initializes the number TX); CLRC663 registers needed for a Protocol change LoadKeyE2 0Fh KevNr: copies a key from the EEPROM into the key buffer StoreKeyE2 0Fh KeyNr, byte1,byte2, byte3, byte4, stores a MIFARE Classic key (size of 6 bytes) into the byte5,byte6; EEPROM ReadRNR 1Ch Copies bytes from the Random Number generator into the FIFO until the FiFo is full Soft Reset 1Fh resets the CLRC663

#### 7.10.3 Command functionality

#### 7.10.3.1 Idle command

Command (00h);

This command indicates that the CLRC663 is in idle mode. This command is also used to terminate the actual command.

#### 7.10.3.2 LPCD command

Command (01h);

This command performs a low-power card detection and/or an automatic trimming of the LFO. After wake-up from standby, the values of the sampled I and Q channels are compared with the min/max threshold values in the registers. If it exceeds the limits, an LPCD\_IRQ will be raised. After the LPCD command the standby is activated, if selected.

#### 7.10.3.3 Load key command

Command (02h), Parameter1 (key byte1),..., Parameter6 (key byte6);

Loads a MIFARE Classic key (6 bytes) for Authentication from the FIFO into the cryptounit.

Abort condition: Less than 6 bytes written to the FIFO.

#### 7.10.3.4 MFAuthent command

Command (03h), Parameter1 (Authentication command code 60h or 61h), Parameter2 (block address), Parameter3 (card serial number byte0), Parameter4 (card serial number byte1), Parameter5 (card serial number byte2), Parameter6 (card serial number byte3);

This command handles the MIFARE Classic authentication in Reader/Writer mode to ensure a secure communication to any MIFARE classic card.

When the MFAuthent command is active, any FIFO access is blocked. Anyhow if there is access to the FIFO, the bit WrErr in the Error register is set.

This command terminates automatically when the MIFARE Classic card is authenticated and the bit MFCrypto1On is set to logic 1.

This command does not terminate automatically, when the card does not answer, therefore the timer should be initialized to automatic mode. In this case, beside the bit IdleIRQ the bit TimerIRQ can be used as termination criteria. During authentication processing the bits RxIRQ and TxIRQ are blocked. The Crypto1On shows if the authentication was successful. The Crypto1On is always valid.

In case, there is an error during authentication, the bit ProtocolErr in the Error register is set to logic 1 and the bit Crypto1On in register Status2Reg is set to logic 0.

#### 7.10.3.5 AckReq command

Command (04h);

Performs a Query (Full command must be written into the FIFO); an Ack and a ReqRn command. All answers to the command will be written into the FIFO. The error flag is copied after the answer into the FIFO.

This command terminates automatically and the then active state is idle.

#### 7.10.3.6 Receive command

Command (05h);

The CLRC663 activates the receiver path and waits for any data stream to be received, according to its register settings. The registers must be set before starting this command

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according to the used protocol and antenna configuration. The correct settings have to be chosen before starting the command.

This command terminates automatically when the received data stream ends. This is indicated either by the end of frame pattern or by the length byte depending on the selected framing and speed.

**Remark:** If the bit RxMultiple in the RxModeReg register is set to logic 1, the Receive command does not terminate automatically. It has to be terminated by activating any other command in the CommandReg register (see <u>Section 8.17.6</u>).

#### 7.10.3.7 Transmit command

Command (06h); data to transmit

The content of the FIFO is transmitted immediately after starting the command. Before transmitting the FIFO, all relevant registers have to be set to transmit data.

This command terminates automatically when the FIFO gets empty. It can be terminated by any other command written to the command register.

#### 7.10.3.8 Transceive command

Command (07h); data to transmit

This command transmits data from FIFO buffer and automatically activates the receiver after a transmission is finished.

Each transmission process starts by writing the command into CommandReg.

**Remark:** If the bit RxMultiple in register RxModeReg is set to logic 1, this command will never leave the receiving state, because the receiving will not be cancelled automatically.

#### 7.10.3.9 WriteE2 command

Command (08h), Parameter1 (addressH), Parameter2 (addressL), Parameter3 (data);

This command writes one byte into the EEPROM. If the FIFO contains no data, the command will wait until the data is available.

Abort condition: Address-parameter outside of allowed range 0x00 – 0x7F.

#### 7.10.3.10 WriteE2PAGE command

Command (09h), Parameter1 (page address), Parameter2..63 (data0, data1...data63);

This command writes up to 64 bytes into the EEPROM. The addresses are not allowed to wrap over a page border. If this is the case, this additional data be ignored and stays in the fifo. The programming starts after 64 bytes are read from the FIFO or the FIFO is empty.

Abort condition: Insufficient parameters in FIFO; Page address parameter outside of range 0x00 - 0x7F.

#### 7.10.3.11 ReadE2 command

Command (0Ah), Parameter1 (addressH), Parameter2 (addressL), Parameter3 (length);

Reads up to 256 bytes from the EEPROM to the FIFO. If a read operation exceeds the address 1FFFh, the read operation continues from address 0000h.

Abort condition: Insufficient parameter in FIFO; Address parameter outside of range.

#### 7.10.3.12 LoadReg command

Command (0Ch), Parameter1 (EEPROM addressH), Parameter2 (EEPROM addressL), Parameter3 (RegAdr), Parameter4 (number);

Read a defined number of bytes from the EEPROM and copies the value into the Register set, beginning at the given address RegAdr.

Abort condition: Insufficient parameter in FIFO; Address parameter outside of range.

#### 7.10.3.13 LoadProtocol command

Command (0Dh), Parameter1 (Protocol number RX), Parameter2 (Protocol number TX);

Reads out the EEPROM Register Set Protocol Area and overwrites the content of the Rx- and Tx- related registers. These registers are important for a Protocol selection.

Abort condition: Insufficient parameter in FIFO	
Table 46. Predefined protocol overview RX <sup>[1]</sup>	

Protocol Number (decimal)	Protocol	Receiver speed [kbits/s]	Receiver Coding
00	ISO/IEC14443 A	106	Manchester SubC
01	ISO/IEC14443 A	212	BPSK
02	ISO/IEC14443 A	424	BPSK
03	ISO/IEC14443 A	848	BPSK
04	ISO/IEC14443 B	106	BPSK
05	ISO/IEC14443 B	212	BPSK
06	ISO/IEC14443 B	424	BPSK
07	ISO/IEC14443 B	848	BPSK
08	FeliCa	212	Manchester
09	FeliCa	424	Manchester
10	ISO/IEC15693	26	SSC
11	ISO/IEC15693	52	SSC
12	ISO/IEC15693	26	DSC
13	EPC/UID	26	SSC
14	ISO/IEC 18000-3 mode 3/ EPC Class-1 HF	212	2/424
15	ISO/IEC 18000-3 mode 3/ EPC Class-1 HF	106	4/424
16	ISO/IEC 18000-3 mode 3/ EPC Class-1 HF	424	2/848

Protocol Number (decimal)	Protocol	Receiver speed [kbits/s]	Receiver Coding
17	ISO/IEC 18000-3 mode 3/ EPC Class-1 HF	212	4/848
18	Jewel	-	-

[1] For more protocol details, please refer to <u>Section 7</u>.

Protocol Number (decimal)	Protocol	Transmitter speed [kbits/s]	Transmitter Coding
00	ISO/IEC14443 A	106	Miller
01	ISO/IEC14443 A	212	Miller
02	ISO/IEC14443 A	424	Miller
03	ISO/IEC14443 A	848	Miller
04	ISO/IEC14443 B	106	NRZ
05	ISO/IEC14443 B	212	NRZ
06	ISO/IEC14443 B	424	NRZ
07	ISO/IEC14443 B	848	NRZ
08	FeliCa	212	Manchester
09	FeliCa	424	Manchester
10	ISO/IEC15693	26	1/4
11	ISO/IEC15693	26	1/4
12	ISO/IEC15693	1,66	1/256
13	EPC/UID	53	Unitray
14	ISO/IEC 18000-3 mode 3/ EPC Class-1 HF	-	based on Tari value, ASK, PIE
15	ISO/IEC 18000-3 mode 3/ EPC Class-1 HF	-	based on Tari value, ASK, PIE
16	ISO/IEC 18000-3 mode 3/ EPC Class-1 HF	-	based on Tari value, ASK, PIE
17	ISO/IEC 18000-3 mode 3/ EPC Class-1 HF	-	based on Tari value, ASK, PIE
18	Jewel	-	-

#### Table 47. Predefined protocol overview TX<sup>[1]</sup>

[1] For more protocol details, please refer to <u>Section 7</u>.

#### 7.10.3.14 LoadKeyE2 command

Command (0Eh), Parameter1 (key number);

Loads a MIFARE Classic key for authentication from the EEPROM into the crypto 1 unit.

Abort condition: Insufficient parameter in FIFO; KeyNr is outside the MIFARE Classic key area.

#### 7.10.3.15 StoreKeyE2 command

Command (0Fh), Parameter1 (KeyNr), Parameter2(keybyte1), Parameter3(keybyte2), Parameter4(keybyte3), Parameter5(keybyte4), Parameter6(keybyte5), Parameter7 (keybyte6);

Stores MIFARE Classic keys into the EEPROM. The key number parameter indicates the first key (n) in the MKA that will be written. If more than one MIFARE Classic key is available in the FIFO then the next key (n+1) will be written until the FIFO is empty. If an incomplete key (less than 6 bytes) is written into the FIFO, this key will be ignored and will remain in the FIFO.

Abort condition: Insufficient parameter in FIFO; KeyNr is outside the MKA;

#### 7.10.3.16 GetRNR command

Command (1Ch);

This command is reading Random Numbers from the random number generator of the CLRC663. The Random Numbers are copied to the FIFO until the FIFO is full.

#### 7.10.3.17 SoftReset command

Command (1Fh);

This command is performing a soft reset. Triggered by this command all the default values for the register setting will be read from the EEPROM and copied into the register set.

## 8 CLRC663 registers

### 8.1 Register bit behavior

Depending on the functionality of a register, the access conditions to the register can vary. In principle, bits with same behavior are grouped in common registers. The access conditions are described in the following table: 5 m

Abbreviation	Behavior	Description
r/w	read and write	These bits can be written and read via the host interface. Since they are used only for control purposes, the content is not influenced by the state machines but can be read by internal state machines.
dy	dynamic	These bits can be written and read via the host interface. They can also be written automatically by internal state machines, for example Command register changes its value automatically after the execution of the command.
r	read only	These register bits indicate hold values which are determined by internal states only.
w	write only	Reading these register bits always returns zero.
RFU	-	These bits are reserved for future use and must not be changed. In case of a required write access, it is recommended to write a logic 0.

#### Table 48. Behavior of register bits and their designation

#### Table 49. CLRC663 registers overview

Address	Register name	Function
00h	Command	Starts and stops command execution
01h	HostCtrl	Host control register
02h	FIFOControl	Control register of the FIFO
03h	WaterLevel	Level of the FIFO underflow and overflow warning
04h	FIFOLength	Length of the FIFO
05h	FIFOData	Data In/Out exchange register of FIFO buffer
06h	IRQ0	Interrupt register 0
07h	IRQ1	Interrupt register 1
08h	IRQ0En	Interrupt enable register 0
09h	IRQ1En	Interrupt enable register 1
0Ah	Error	Error bits showing the error status of the last command execution
0Bh	Status	Contains status of the communication
0Ch	RxBitCtrl	Control register for anticollision adjustments for bit oriented protocols
0Dh	RxColl	Collision position register
0Eh	TControl	Control of Timer 03

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Address	Register name	Function
0Fh	T0Control	Control of Timer0
10h	T0ReloadHi	High register of the reload value of Timer0
11h	T0ReloadLo	Low register of the reload value of Timer0
12h	T0CounterValHi	Counter value high register of Timer0
13h	T0CounterValLo	Counter value low register of Timer0
14h	T1Control	Control of Timer1
15h	T1ReloadHi	High register of the reload value of Timer1
16h	T1ReloadLo	Low register of the reload value of Timer1
17h	T1CounterValHi	Counter value high register of Timer1
18h	T1CounterValLo	Counter value low register of Timer1
19h	T2Control	Control of Timer2
1Ah	T2ReloadHi	High byte of the reload value of Timer2
1Bh	T2ReloadLo	Low byte of the reload value of Timer2
1Ch	T2CounterValHi	Counter value high byte of Timer2
1Dh	T2CounterValLo	Counter value low byte of Timer2
1Eh	T3Control	Control of Timer3
1Fh	T3ReloadHi	High byte of the reload value of Timer3
20h	T3ReloadLo	Low byte of the reload value of Timer3
21h	T3CounterValHi	Counter value high byte of Timer3
22h	T3CounterValLo	Counter value low byte of Timer3
23h	T4Control	Control of Timer4
24h	T4ReloadHi	High byte of the reload value of Timer4
25h	T4ReloadLo	Low byte of the reload value of Timer4
26h	T4CounterValHi	Counter value high byte of Timer4
27h	T4CounterValLo	Counter value low byte of Timer4
28h	DrvMod	Driver mode register
29h	TxAmp	Transmitter amplifier register
2Ah	DrvCon	Driver configuration register
2Bh	Txl	Transmitter register
2Ch	TxCrcPreset	Transmitter CRC control register, preset value
2Dh	RxCrcPreset	Receiver CRC control register, preset value
2Eh	TxDataNum	Transmitter data number register
2Fh	TxModWidth	Transmitter modulation width register
30h	TxSym10BurstLen	Transmitter symbol 1 + symbol 0 burst length register
31h	TXWaitCtrl	Transmitter wait control
32h	TxWaitLo	Transmitter wait low
33h	FrameCon	Transmitter frame control

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Address	Register name	Function
34h	RxSofD	Receiver start of frame detection
35h	RxCtrl	Receiver control register
36h	RxWait	Receiver wait register
37h	RxThreshold	Receiver threshold register
38h	Rcv	Receiver register
39h	RxAna	Receiver analog register
3Ah	RFU	No function implemented for CLRC66301 and CLRC66302
	LPCD_Options	For CLRC66303: Options for LPCD configuration
3Bh	SerialSpeed	Serial speed register
3Ch	LFO_Trimm	Low-power oscillator trimming register
3Dh	PLL_Ctrl	IntegerN PLL control register, for microcontroller clock output adjustment
3Eh	PLL_DivOut	IntegerN PLL control register, for microcontroller clock output adjustment
3Fh	LPCD_QMin	Low-power card detection Q channel minimum threshold
40h	LPCD_QMax	Low-power card detection Q channel maximum threshold
41h	LPCD_IMin	Low-power card detection I channel minimum threshold
42h	LPCD_I_Result	Low-power card detection I channel result register
43h	LPCD_Q_Result	Low-power card detection Q channel result register
44h	PadEn	PIN enable register
45h	PadOut	PIN out register
46h	PadIn	PIN in register
47h	SigOut	Enables and controls the SIGOUT Pin
48h	TxBitMod	Transmitter bit mode register
49h	RFU	-
4Ah	TxDataCon	Transmitter data configuration register
4Bh	TxDataMod	Transmitter data modulation register
4Ch	TxSymFreq	Transmitter symbol frequency
4Dh	TxSym0H	Transmitter symbol 0 high register
4Eh	TxSym0L	Transmitter symbol 0 low register
4Fh	TxSym1H	Transmitter symbol 1 high register
50h	TxSym1L	Transmitter symbol 1 low register
51h	TxSym2	Transmitter symbol 2 register
52h	TxSym3	Transmitter symbol 3 register
53h	TxSym10Len	Transmitter symbol 1 + symbol 0 length register
54h	TxSym32Len	Transmitter symbol 3 + symbol 2 length register
55h	TxSym10BurstCtrl	Transmitter symbol 1 + symbol 0 burst control register
56h	TxSym10Mod	Transmitter symbol 1 + symbol 0 modulation register
57h	TxSym32Mod	Transmitter symbol 3 + symbol 2 modulation register

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Address	Register name	Function
58h	RxBitMod	Receiver bit modulation register
59h	RxEofSym	Receiver end of frame symbol register
5Ah	RxSyncValH	Receiver synchronisation value high register
5Bh	RxSyncValL	Receiver synchronisation value low register
5Ch	RxSyncMod	Receiver synchronisation mode register
5Dh	RxMod	Receiver modulation register
5Eh	RxCorr	Receiver correlation register
5Fh	FabCal	Calibration register of the receiver, calibration performed at production
48h-5Fh	RFU	-
7Fh	Version	Version and subversion register

## 8.2 Command configuration

### 8.2.1 Command

Starts and stops command execution.

#### Table 50. Command register (address 00h)

Bit	7	6	5	4	3	2	1	0
Symbol	Standby	Modem Off	RFU		Co	ommand		
Access rights	dy	r/w	-			dy		

#### Table 51. Command bits

	oommana bito	
Bit	Symbol	Description
7	Standby	Set to 1, the IC is entering power-down mode.
6	ModemOff	Set to logic 1, the receiver and the transmitter circuit is powering down.
5	RFU	-
4 to 0	Command	Defines the actual command for the CLRC663.

## 8.3 SAM configuration register

#### 8.3.1 HostCtrl

Via the HostCtrl Register the interface access right can be controlled

Fable 52. HostCtrl register (address 01h);								
Bit	7	6	5	4	3	2	1	0
Symbol	RegEn	BusHost	BusSAM	RFU	SAMInterface	SAMInterface	RFU	RFU
Access rights	dy	r/w	r/w	-	r/w	r/w	-	-

#### Table 53. HostCtrl bits

Bit	Symbol	Description
7	RegEn	If this bit is set to logic 1, the register HostCtrl_reg can be changed at the next register access. The next write access clears this bit automatically.
6	BusHost	Set to logic 1, the bus is controlled by the host. This bit cannot be set together with the bit BusSAM. This bit can only be set if the bit RegEn is previously set.
5	BusSAM	Set to logic 1, the bus is controlled by the SAM. This bit cannot be set together with BusHost. This bit can only be set if the bit RegEn is previously set.
4	RFU	-
3 to 2	SAMInterface	0h:SAM Interface switched off 1h:SAM Interface SPI active 2h:SAM Interface I <sup>2</sup> CL active 3h:SAM Interface I <sup>2</sup> C active
1 to 0	RFU	-

## 8.4 FIFO configuration register

#### 8.4.1 FIFOControl

FIFOControl defines the characteristics of the FIFO

Table 54.	<b>FIFOControl</b>	register	(address 02h);

Bit	7	6	5	4	3	2	1	0
Symbol	FIFOSize	HiAlert	LoAlert	FIFOFlush	RFU	WaterLe velExtBit	FIFOLengthExtBits	
Access rights	r/w	r	r	w	-	r/w	r	

#### Table 55. FIFOControl bits

Bit	Symbol	Description
7	FIFOSize	Set to logic 1, FIFO size is 255 bytes; Set to logic 0, FIFO size is 512 bytes. It is recommended to change the FIFO size only, when the FIFO content had been cleared.

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Bit	Symbol	Description
6	HiAlert	Set to logic 1, when the number of bytes stored in the FIFO buffer fulfils the following equation: HiAlert = (FIFOSize - FIFOLength) <= WaterLevel
5	LoAlert	Set to logic 1, when the number of bytes stored in the FIFO buffer fulfils the following conditions: LoAlert =1 if FIFOLength <= WaterLevel
4	FIFOFlush	Set to logic 1 clears the FIFO buffer. Reading this bit will always return 0
3	RFU	-
2	WaterLevelExtBit	Defines the bit 8 (MSB) for the waterlevel (extension of register WaterLevel). This bit is only evaluated in the 512-byte FIFO mode. Bits 70 are defined in register WaterLevel.
1 to 0	FIFOLengthExtBits	Defines the bit9 (MSB) and bit8 for the FIFO length (extension of FIFOLength). These two bits are only evaluated in the 512-byte FIFO mode. The bits 70 are defined in register FIFOLength.

#### 8.4.2 WaterLevel

Defines the level for FIFO under- and overflow warning levels. This register is extended by 1 bit in FIFOControl in case the 512-byte FIFO mode is activated by setting bit FIFOControl.FIFOSize.

#### Table 56. WaterLevel register (address 03h);

Bit	7	6	5	4	3	2	1	0
Symbol	WaterLevelBits							
Access rights	r/w	r/w	r/w	r/w	r/w	r/w	r/w	r/w

#### Table 57. WaterLevel bits

Bit	Symbol	Description
7 to 0	WaterLevelBits	Sets a level to indicate a FIFO-buffer state which can be read from bits HighAlert and LowAlert in the FifoControl. In 512-byte FIFO mode, the register is extended by bit WaterLevelExtBit in the FIFOControl. This functionality can be used to avoid a FIFO buffer overflow or underflow: The bit HiAlert bit in FIFO Control is read logic 1, if the number of but a in the FIFO buffer is equal or less than the number defined bu
		bytes in the FIFO-buffer is equal or less than the number defined by the waterlevel configuration.
		The bit LoAlert bit in FIFO control is read logic 1, if the number of bytes in the FIFO buffer is equal or less than the number defined by the waterlevel configuration.
		Note: For the calculation of HiAlert and LoAlert, see register description of these bits (see section <u>Section 8.4.1</u> ).

#### 8.4.3 FIFOLength

Number of bytes in the FIFO buffer. In 512-byte mode, this register is extended by FIFOControl.FifoLength.

Table 58. FIFOLength register (address 04h); reset value: 00h										
Bit	7	6	5	4	3	2	1	0		
Symbol		FIFOLength								
Access rights				d	Ý					

#### Table 59. FIFOLength bits

Bit	Symbol	Description
7 to 0	FIFOLength	Indicates the number of bytes in the FIFO buffer. In 512-byte mode this register is extended by the bits FIFOLength in the FIFOControl register. Writing to the FIFOData register increments, reading decrements the number of available bytes in the FIFO.

#### 8.4.4 FIFOData

In- and output of FIFO buffer. Contrary to any read/write access to other addresses, reading or writing to the FIFO address does not increment the address pointer. Writing to the FIFOData register increments, reading decrements the number of bytes present in the FIFO.

#### Table 60. FIFOData register (address 05h);

Bit	7	6	5	4	3	2	1	0
Symbol	FIFOData							
Access rights	dy	dy	dy	dy	dy	dy	dy	dy

#### Table 61. FIFOData bits

Bit	Symbol	Description			
7 to 0	FIFOData	Data input and output port for the internal FIFO buffer. Refer to Section 7.5 "Buffer".			

## 8.5 Interrupt configuration registers

The Registers IRQ0 register and IRQ1 register implement a special functionality to avoid the unintended modification of bits.

The mechanism of changing register contents requires the following consideration: IRQ(x).Set indicates, if a set bit on position 0 to 6 shall be cleared or set. Depending on the content of IRQ(x).Set, a write of a 1 to positions 0 to 6 either clears or sets the corresponding bit. With this register, the application can modify the interrupt status which is maintained by the CLRC663.

0 RxSOF IRQ dy

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Bit 7 indicates, if the intended modification is a setting or clearance of a bit. Any 1 written to a bit position 6...0 will trigger the setting or clearance of this bit as defined by bit 7. Example: writing FFh sets all bits 6..0, writing 7Fh clears all bits 6..0 of the interrupt request register

#### 8.5.1 IRQ0 register

Interrupt request register 0.

Table 62. IF	RQ0 register	(address 06	h); reset valu	ıe: 00h			
Bit	7	6	5	4	3	2	1
Symbol	Set	HiAlertIRQ	LoAlertIRQ	IdleIRQ	TxIRQ	RxIRQ	ErrlRQ
Access rights	w	dy	dy	dy	dy	dy	dy

## Table 63 IRO0 bits

Table 03	S. IRQU DITS	
Bit	Symbol	Description
7	Set	<ol> <li>1: writing a 1 to a bit position 60 sets the interrupt request</li> <li>0: Writing a 1 to a bit position 60 clears the interrupt request</li> </ol>
6	HiAlertIRQ	Set, when bit HiAlert in register Status1Reg is set. In opposition to HiAlert, HiAlertIRQ stores this event.
5	LoAlertIRQ	Set, when bit LoAlert in register Status1 is set. In opposition to LoAlert, LoAlertIRQ stores this event.
4	IdleIRQ	Set, when a command terminates by itself e.g. when the Command changes its value from any command to the Idle command. If an unknown command is started, the Command changes its content to the idle state and the bit IdleIRQ is set. Starting the Idle command by the Controller does not set bit IdleIRQ.
3	TxIRQ	Set, when data transmission is completed, which is immediately after the last bit is sent.
2	RxIRQ	Set, when the receiver detects the end of a data stream. Note: This flag is no indication that the received data stream is correct. The error flags have to be evaluated to get the status of the reception.
1	ErrIRQ	Set, when the one of the following errors is set: FifoWrErr, FiFoOvI, ProtErr, NoDataErr, IntegErr.
0	RxSOFIrq	Set, when a SOF or a subcarrier is detected.

#### 8.5.2 IRQ1 register

Interrupt request register 1.

Table 64. IRQ	1 register	(address 07h)
---------------	------------	---------------

Bit	7	6	5	4	3	2	1	0
Symbol	Set	GlobalIRQ	LPCD_IRQ	Timer4IRQ	Timer3IRQ	Timer2IRQ	Timer1IRQ	Timer0IRQ
Access rights	w	dy	dy	dy	dy	dy	dy	dy

#### Table 65. IRQ1 bits

Bit	Symbol	Description
7	Set	<ol> <li>writing a 1 to a bit position 50 sets the interrupt request</li> <li>Writing a 1 to a bit position 50 clears the interrupt request</li> </ol>
6	GlobalIRQ	Set, if an enabled IRQ occurs.
5	LPCD_IRQ	Set if a card is detected in Low-power card detection sequence.
4	Timer4IRQ	Set to logic 1 when Timer4 has an underflow.
3	Timer3IRQ	Set to logic 1 when Timer3 has an underflow.
2	Timer2IRQ	Set to logic 1 when Timer2 has an underflow.
1	Timer1IRQ	Set to logic 1 when Timer1 has an underflow.
0	Timer0IRQ	Set to logic 1 when Timer0 has an underflow.

#### 8.5.3 IRQ0En register

Interrupt request enable register for IRQ0. This register allows defining if an interrupt request is processed by the CLRC663.

		3	/					
Bit	7	6	5	4	3	2	1	0
Symbol	IRQ_Inv	Hi AlertIRQEn	LoAlertIRQEn	ldlelRQEn	TxIRQEn	RxIRQEn	ErrlRQEn	RxSOF IRQEn
Access rights	r/w	r/w	r/w	r/w	r/w	r/w	r/w	r/w

#### Table 66. IRQ0En register (address 08h)

	7. IRQUEIT DILS	
Bit	Symbol	Description
7	IRQ_Inv	Set to one the signal of the IRQ pin is inverted
6	Hi AlerIRQEn	Set to logic 1, it allows the High Alert interrupt Request (indicated by the bit HiAlertIRQ) to be propagated to the GlobalIRQ
5	Lo AlertIRQEn	Set to logic 1, it allows the Low Alert Interrupt Request (indicated by the bit LoAlertIRQ) to be propagated to the GlobalIRQ
4	IdleIRQEn	Set to logic 1, it allows the Idle interrupt request (indicated by the bit IdleIRQ) to be propagated to the GlobalIRQ
3	TxIRQEn	Set to logic 1, it allows the transmitter interrupt request (indicated by the bit TxtIRQ) to be propagated to the GlobalIRQ
2	RxIRQEn	Set to logic 1, it allows the receiver interrupt request (indicated by the bit RxIRQ) to be propagated to the GlobalIRQ
1	ErrIRQEn	Set to logic 1, it allows the Error interrupt request (indicated by the bit ErrorIRQ) to be propagated to the GlobalIRQ
0	RxSOFIRQEn	Set to logic 1, it allows the RxSOF interrupt request (indicated by the bit RxSOFIRQ) to be propagated to the GlobalIRQ

#### Table 67. IRQ0En bits

0 Timer0 IRQEn r/w

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#### 8.5.4 IRQ1En

Interrupt request enable register for IRQ1.

Table 68	able 68. IRQ1EN register (address 09h);									
Bit	7	6	5	4	3	2	1			
Symbol	IRQPushPull	IRQPinEn	LPCD_IRQEn	Timer4 IRQEn	Timer3 IRQEn	Timer2 IRQEn	Timer1 IRQEn			
Access riahts	r/w	r/w	r/w	r/w	r/w	r/w	r/w			

#### Table 69. IRQ1EN bits

i ubic	OS. INGLENDIS	
Bit	Symbol	Description
7	IRQPushPull	Set to 1 the IRQ-pin acts as PushPull pin, otherwise it acts as OpenDrain pin
6	IRQPinEN	Set to logic 1, it allows the global interrupt request (indicated by the bit GlobalIRQ) to be propagated to the interrupt pin
5	LPCD_IRQEN	Set to logic 1, it allows the LPCDinterrupt request (indicated by the bit LPCDIRQ) to be propagated to the GlobalIRQ
4	Timer4IRQEn	Set to logic 1, it allows the Timer4 interrupt request (indicated by the bit Timer4IRQ) to be propagated to the GlobalIRQ
3	Timer3IRQEn	Set to logic 1, it allows the Timer3 interrupt request (indicated by the bit Timer3IRQ) to be propagated to the GlobalIRQ
2	Timer2IRQEn	Set to logic 1, it allows the Timer2 interrupt request (indicated by the bit Timer2IRQ) to be propagated to the GlobalIRQ
1	Timer1IRQEn	Set to logic 1, it allows the Timer1 interrupt request (indicated by the bit Timer1IRQ) to be propagated to the GlobalIRQ
0	Timer0IRQEn	Set to logic 1, it allows the Timer0 interrupt request (indicated by the bit Timer0IRQ) to be propagated to the GlobalIRQ

## 8.6 Contactless interface configuration registers

#### 8.6.1 Error

Error register.

Table 70. Error register (address 0Ah)

Bit	7	6	5	4	3	2	1	0
Symbol	EE_Err	FiFoWrErr	FIFOOvl	MinFrameErr	NoDataErr	CollDet	ProtErr	IntegErr
Access rights	dy	dy	dy	dy	dy	dy	dy	dy

#### Table 71. Error bits

Bit	Symbol	Description
7	EE_Err	An error appeared during the last EEPROM command. For details see the descriptions of the EEPROM commands
6	FIFOWrErr	Data was written into the FIFO, during a transmission of a possible CRC, during "RxWait", "Wait for data" or "Receiving" state, or during an authentication command. The Flag is cleared when a new CL command is started. If RxMultiple is active, the flag is cleared after the error flags have been written to the FIFO.
5	FIFOOvl	Data is written into the FIFO when it is already full. The data that is already in the FIFO remains untouched. All data that is written to the FIFO after this Flag is set to 1 will be ignored.
4	Min FrameErr	A valid SOF was received, but afterwards less than 4 bits of data were received. Note: Frames with less than 4 bits of data are automatically discarded and the RxDecoder stays enabled. Furthermore no RxIRQ is set. The same is valid for less than 3 bytes, if the EMD suppression is activated Note: MinFrameErr is automatically cleared at the start of a receive or transceive command. In case of a transceive command, it is cleared at the start of the receiving phase ("Wait for data" state)
3	NoDataErr	Data should be sent, but no data is in FIFO
2	CollDet	A collision has occurred. The position of the first collision is shown in the register RxColl. Note: CollDet is automatically cleared at the start of a receive or transceive command. In case of a transceive command, it is cleared at the start of the receiving phase ("Wait for data" state). Note: If a collision is part of the defined EOF symbol, CollDet is not set to 1.
1	ProtErr	A protocol error has occurred. A protocol error can be a wrong stop bit, a missing or wrong ISO/IEC14443B EOF or SOF or a wrong number of received data bytes. When a protocol error is detected, data reception is stopped. Note: ProtErr is automatically cleared at start of a receive or transceive command. In case of a transceive command, it is cleared at the start of the receiving phase ("Wait for data" state). Note: When a protocol error occurs the last received data byte is not written into the FIFO.
0	IntegErr	A data integrity error has been detected. Possible cause can be a wrong parity or a wrong CRC. In case of a data integrity error the reception is continued. Note: IntegErr is automatically cleared at start of a Receive or Transceive command. In case of a Transceive command, it is cleared at the start of the receiving phase ("Wait for data" state). Note: If the NoColl bit is set, also a collision is setting the IntegErr.

### 8.6.2 Status

Status register.

•	Table 72.	Status	register	(address	0Bh)
---	-----------	--------	----------	----------	------

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	Crypto1On	-	-	ComState		

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Bit	7	6	5	4	3	2	1	0
Access rights	RFU	RFU	dy	RFU	RFU	r		

#### Table 73. Status bits

Bit	Symbol	Description
7 to 6	-	RFU
5	Crypto1On	Indicates if the MIFARE encryption/decryption is on. Clearing this bit is switching the MIFARE encryption/decryption off. The bit can only be set by the MFAuthent command.
4 to 3	-	RFU
2 to 0	ComState	ComState shows the status of the transmitter and receiver state machine:
		000b Idle
		001b TxWait
		011b Transmitting
		101b RxWait
		110b Wait for data
		111b Receiving
		100b not used

### 8.6.3 RxBitCtrl

#### Receiver control register.

Table 74. RxBitCtrl register (address 0Ch);									
Bit	7	6	5	4	3	2	1	0	
Symbol	ValuesAfterColl	RxAlign			NoColl	RxLastBits			
Access rights	r/w	r/w	r/w	r/w	r/w	W	W		

#### Table 75. RxBitCtrl bits

Bit	Symbol	Description
7	ValuesAfter Coll	If cleared, every received bit after a collision is replaced by a zero. This function is needed for ISO/IEC14443 anticollision

Bit	Symbol	Description
6 to 4	RxAlign	Used for reception of bit oriented frames: RxAlign defines the bit position length for the first bit received to be stored. Further received bits are stored at the following bit positions. Example:
		RxAlign = 0h - the LSB of the received bit is stored at bit 0, the second received bit is stored at bit position 1.
		RxAlign = 1h - the LSB of the received bit is stored at bit 1, the second received bit is stored at bit position 2.
		RxAlign = 7h - the LSB of the received bit is stored at bit 7, the second received bit is stored in the following byte at position 0.
		Note: If RxAlign = 0, data is received byte-oriented, otherwise bit- oriented.
3	NoColl	If this bit is set, a collision will result in an IntegErr
2 to 0	RxLastBits	Defines the number of valid bits of the last data byte received in bit- oriented communications. If zero the whole byte is valid.
		Note: These bits are set by the RxDecoder in a bit-oriented communication at the end of the communication. They are reset at start of reception.

#### 8.6.4 RxColl

Receiver collision register.

### Table 76. RxColl register (address 0Dh);

Bit	7	6	5	4	3	2	1	0	
Symbol	CollPosValid		CollPos						
Access rights	r				r				

Bit	Symbol	Description				
7	CollPos Valid	If set to 1, the value of CollPos is valid. Otherwise no collision is detected or the position of the collision is out of the range of bits CollPos.				
6 to 0	CollPos	These bits show the bit position of the first detected collision in a received frame (only data bits are interpreted). CollPos can only be displayed for the first 8 bytes of a data stream. Example: 00h indicates a bit collision in the 1st bit 01h indicates a bit collision in the 2nd bit 08h indicates a bit collision in the 9th bit (1st bit of 2nd byte) 3Fh indicates a bit collision in the 64th bit (8th bit of the 8th byte) These bits shall only be interpreted in Passive communication mode at 106 kbit/s or ISO/IEC 14443 type A and read /write mode for MIFARE Classic or ISO/IEC 15693/ICODE SLI read/write mode if bit CollPosValid is set.				
		Note: If RxBitCtrl.RxAlign is set to a value different to 0, this value is included in the CollPos.				
		Example: RxAlign = 4h, a collision occurs in the 4th received bit (which is the last bit of that UID byte). The CollPos = 7h in this case.				

## 8.7 Timer configuration registers

#### 8.7.1 TControl

Control register of the timer section.

The TControl implements a special functionality to avoid the not intended modification of bits.

Bit 3..0 indicates, which bits in the positions 7..4 are intended to be modified.

Example: writing FFh sets all bits 7..4, writing F0h does not change any of the bits 7..4

 Table 78. TControl register (address 0Eh)

Bit	7	6	5	4	3	2	1	0
Symbol	T3Running	T2Running	T1Running	T0Running	T3Start StopNow	T2Start StopNow	T1Start StopNow	T0Start StopNow
Access rights	dy	dy	dy	dy	W	W	w	w

#### Table 79. TControl bits

Table 13.	I Control bits	
Bit	Symbol	Description
7	T3Running	Indicates Timer3 is running.If the bit T3startStopNow is set/reset, this bit and the timer can be started/stopped
6	T2Running	Indicates Timer2 is running. If the bit T2startStopNow is set/reset, this bit and the timer can be started/stopped
5	T1Running	Indicates tTmer1 is running. If the bit T1startStopNow is set/reset, this bit and the timer can be started/stopped
4	T0Running	Indicates Timer0 is running. If the bit T0startStopNow is set/reset, this bit and the timer can be started/stopped
3	T3StartStop Now	The bit 7 of TControl T3Running can be modified if set
2	T2StartStop Now	The bit 6of TControl T2Running can be modified if set
1	T1StartStop Now	The bit 5of TControl T1Running can be modified if set
0	T0StartStop Now	The bit 4 of TControl T0Running can be modified if set

#### 8.7.2 T0Control

Control register of the Timer0.

Table 80.	<b>T0Control</b>	register	(address	0Fh);
-----------	------------------	----------	----------	-------

Bit	7	6	5	4	3	2	1	0
Symbol	T0StopRx	-	T0Start		T0AutoRestart	-	тс	)Clk

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Bit	7	6	5	4	3	2	1	0
Access rights	r/w	RFU	r/w		r/w	RFU	n	/w

### Table 81. T0Control bits

Bit	Symbol	Description
7	T0StopRx	If set, the timer stops immediately after receiving the first 4 bits. If cleared the timer does not stop automatically. Note: If LFO Trimming is selected by T0Start, this bit has no effect.
6	-	RFU
5 to 4	T0Start	<ul> <li>00b: The timer is not started automatically</li> <li>01 b: The timer starts automatically at the end of the transmission</li> <li>10 b: Timer is used for LFO trimming without underflow (Start/Stop on PosEdge)</li> <li>11 b: Timer is used for LFO trimming with underflow (Start/Stop on PosEdge)</li> </ul>
3	T0AutoRestart	<ol> <li>the timer automatically restarts its count-down from TOReloadValue, after the counter value has reached the value zero.</li> <li>the timer decrements to zero and stops.</li> <li>The bit Timer1IRQ is set to logic 1 when the timer underflows.</li> </ol>
2	-	RFU
1 to 0	T0Clk	<ul> <li>00 b: The timer input clock is 13.56 MHz.</li> <li>01 b: The timer input clock is 211,875 kHz.</li> <li>10 b: The timer input clock is an underflow of Timer2.</li> <li>11 b: The timer input clock is an underflow of Timer1.</li> </ul>

### 8.7.2.1 T0ReloadHi

High byte reload value of the Timer0.

Table 82.	T0ReloadHi re	gister (address	s 10h);

Bit	7	6	5	4	3	2	1	0
Symbol	T0Reload Hi							
Access rights		r/w						

### Table 83. T0ReloadHi bits

Bit	Symbol	Description
7 to 0	T0ReloadHi	Defines the high byte of the reload value of the timer. With the start event, the timer loads the value of the registers T0ReloadValHi, T0ReloadValLo. Changing this register affects the timer only at the next start event.

### 8.7.2.2 T0ReloadLo

Low byte reload value of the Timer0.

Table 84. T0ReloadLo register (address 11h);										
Bit	7         6         5         4         3         2         1         0									
Symbol		T0ReloadLo								
Access rights		r/w								

#### Table 85. T0ReloadLo bits

Tuble 00.	I UICEIDAULO DIL	
Bit	Symbol	Description
7 to0	T0ReloadLo	Defines the low byte of the reload value of the timer. With the start event, the timer loads the value of the T0ReloadValHi, T0ReloadValLo. Changing this register affects the timer only at the next start event.

### 8.7.2.3 T0CounterValHi

High byte of the counter value of Timer0.

### Table 86. T0CounterValHi register (address 12h)

Bit	7 6 5 4 3 2 1							
Symbol		T0CounterValHi						
Access rights	dy							

#### Table 87. T0CounterValHi bits

Bit	Symbol	Description
7to0	T0Counter ValHi	High byte value of the Timer0. This value shall not be read out during reception.

### 8.7.2.4 T0CounterValLo

Low byte of the counter value of Timer0.

### Table 88. T0CounterValLo register (address 13h)

Bit	7	6	5	4	3	2	1	0
Symbol	T0CounterValLo							
Access rights				dy	,			

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### Table 89. T0CounterValLo bits

Bit	Symbol	Description
7 to 0	T0CounterValLo	Low byte value of the Timer0. This value shall not be read out during reception.

### 8.7.2.5 T1Control

Control register of the Timer1.

1	Table 90.	T	1Control	reg	ister	(addre	ess 14h);

Bit	7	6	5	4	3	2	1	0
Symbol	T1StopRx	-	T1Start		T1AutoRestart	-	T1Clk	
Access rights	r/w	RFU	r/w		r/w	RFU	r	/w

Bit	Symbol	Description
7	T1StopRx	If set, the timer stops after receiving the first 4 bits. If cleared, the timer is not stopped automatically. Note: If LFO trimming is selected by T1start, this bit has no effect.
6	-	RFU
5 to 4	T1Start	<ul> <li>00b: The timer is not started automatically</li> <li>01 b: The timer starts automatically at the end of the transmission</li> <li>10 b: Timer is used for LFO trimming without underflow (Start/Stop on PosEdge)</li> <li>11 b: Timer is used for LFO trimming with underflow (Start/Stop on PosEdge)</li> </ul>
3	T1AutoRestart	Set to logic 1, the timer automatically restarts its countdown from T1ReloadValue, after the counter value has reached the value zero. Set to logic 0 the timer decrements to zero and stops. The bit Timer1IRQ is set to logic 1 when the timer underflows.
2	-	RFU
1 to 0	T1Clk	00 b: The timer input clock is 13.56 MHz 01 b: The timer input clock is 211,875 kHz. 10 b: The timer input clock is an underflow of Timer0 11 b: The timer input clock is an underflow of Timer2

### 8.7.2.6 T1ReloadHi

High byte (MSB) reload value of the Timer1.

Table 92.	T0ReloadHi	register	(address 15h	)
	1 VI	regiotor		,

Bit	7	6	5	4	3	2	1	0
Symbol	T1ReloadHi							
Access rights		r/w						

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Table 93.	Table 93. T1ReloadHi bits						
Bit	Symbol	Description					
7 to 0	T1ReloadHi	Defines the high byte reload value of the Timer 1. With the start event, the timer loads the value of the T1ReloadValHi and T1ReloadValLo. Changing this register affects the Timer only at the next start event.					

### 8.7.2.7 T1ReloadLo

Low byte (LSB) reload value of the Timer1.

Table 94. T1	ReloadLo reg	ister (addre	ss 16h)					
Bit	7	6	5	4	3	2	1	0
Symbol	T1ReloadLo							
Access rights				r/w				_

Table 95.	Table 95. T1ReloadValLo bits						
Bit	Symbol	Description					
7 to 0	T1ReloadLo	Defines the low byte of the reload value of the Timer1. Changing this register affects the timer only at the next start event.					

### 8.7.2.8 T1CounterValHi

High byte (MSB) of the counter value of byte Timer1.

#### Table 96. T1CounterValHi register (address 17h)

Bit	7	6	5	4	3	2	1	0
Symbol	T1CounterValHi							
Access rights				dy	,			

#### Table 97. T1CounterValHi bits

Bit	Symbol	Description			
7 to 0	T1Counter ValHi	High byte of the current value of the Timer1. This value shall not be read out during reception.			

### 8.7.2.9 T1CounterValLo

Low byte (LSB) of the counter value of byte Timer1.

### Table 98. T1CounterValLo register (address 18h)

Bit	7	6	5	4	3	2	1	0
Symbol	T1CounterValLo							
Access rights	dy							

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### Table 99. T1CounterValLo bits

### 8.7.2.10 T2Control

Control register of the Timer2.

Table 100. T2Control register (address 19h)												
Bit	7	6	5	4	3	2	1	0				
Symbol	T2StopRx	-	T2Sta	T2Start		-	T2Clk					
Access rights	r/w	RFU	r/w		r/w	RFU	r	/w				

Bit	Symbol	Description
7	T2StopRx	If set the timer stops immediately after receiving the first 4 bits. If cleared indicates, that the timer is not stopped automatically. Note: If LFO Trimming is selected by T2Start, this bit has no effect.
6	-	RFU
5 to 4	T2Start	<ul> <li>00 b: The timer is not started automatically.</li> <li>01 b: The timer starts automatically at the end of the transmission.</li> <li>10 b: Timer is used for LFO trimming without underflow (Start/Stop on PosEdge).</li> <li>11 b: Timer is used for LFO trimming with underflow (Start/Stop on PosEdge).</li> </ul>
3	T2AutoRestart	Set to logic 1, the timer automatically restarts its countdown from T2ReloadValue, after the counter value has reached the value zero. Set to logic 0 the timer decrements to zero and stops. The bit Timer2IRQ is set to logic 1 when the timer underflows
2	-	RFU
1 to 0	T2Clk	00 b: The timer input clock is 13.56 MHz. 01 b: The timer input clock is 212 kHz. 10 b: The timer input clock is an underflow of Timer0 11b: The timer input clock is an underflow of Timer1

### 8.7.2.11 T2ReloadHi

High byte of the reload value of Timer2.

Table 102.	T2ReloadHi r	egister (add	lress 1Ah)

Bit	7	6	5	4	3	2	1	0	
Symbol		T2ReloadHi							
Access rights				r/w	1				

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Table 103.	T2Reload bits	
Bit	Symbol	Description
7 to 0	T2ReloadHi	Defines the high byte of the reload value of the Timer2. With the start event, the timer load the value of the T2ReloadValHi and T2ReloadValLo. Changing this register affects the timer only at the next start event.

### 8.7.2.12 T2ReloadLo

Low byte of the reload value of Timer2.

Table 104. T2ReloadLo register (address 1Bh)											
Bit	7	6	5	4	3	2	1	0			
Symbol		T2ReloadLo									
Access rights				r/w	I						

#### Table 105. T2ReloadLo bits

Bit	Symbol	Description						
7 to 0	T2ReloadLo	Defines the low byte of the reload value of the Timer2. With the start event, the timer load the value of the T2ReloadValHi and T2ReloadVaLo. Changing this register affects the timer only at the next start event.						

### 8.7.2.13 T2CounterValHi

High byte of the counter register of Timer2.

Table 106. T2CounterValHi register (address 1Ch)										
Bit	7	6	5	4	3	2	1	0		
Symbol		T2CounterValHi								
Access rights				dy						

Table 107.	Table 107. T2CounterValHi bits							
Bit	Symbol	Description						
7 to 0	T2Counter ValHi	High byte current counter value of Timer2. This value shall not be read out during reception.						

### 8.7.2.14 T2CounterValLoReg

Low byte of the current value of Timer 2.

	T2CounterVa	ILo register	(address 1Dh)								
Bit	7	6	5	4	3	2	1	0			
Symbol		T2CounterValLo									
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Bit	7	6	5	4	3	2	1	0
Access rights				dy				

### Table 109. T2CounterValLo bits

Bit	Symbol	Description
7 to 0	T2Counter	Low byte of the current counter value of Timer1Timer2.
	ValLo	This value shall not be read out during reception.

### 8.7.2.15 T3Control

Control register of the Timer 3.

#### Table 110. T3Control register (address 1Eh)

Bit	7	6	5	4	3	2	1	0
Symbol	T3StopRx	-	T3Start		T3AutoRestart	-	T3Clk	
Access rights	r/w	RFU	r/w	r/w		RFU	r	/w

### Table 111. T3Control bits

Table 111.	T3Control bits	
Bit	Symbol	Description
7	T3StopRx	If set, the timer stops immediately after receiving the first 4 bits. If cleared, indicates that the timer is not stopped automatically. Note: If LFO Trimming is selected by T3Start, this bit has no effect.
6	-	RFU
5 to 4	T3Start	<ul> <li>00b - timer is not started automatically</li> <li>01 b - timer starts automatically at the end of the transmission</li> <li>10 b - timer is used for LFO trimming without underflow (Start/Stop on PosEdge)</li> <li>11 b - timer is used for LFO trimming with underflow (Start/Stop on PosEdge).</li> </ul>
3	T3AutoRestart	Set to logic 1, the timer automatically restarts its countdown from T3ReloadValue, after the counter value has reached the value zero. Set to logic 0 the timer decrements to zero and stops. The bit Timer1IRQ is set to logic 1 when the timer underflows.
2	-	RFU
1 to 0	T3Clk	00 b - the timer input clock is 13.56 MHz. 01 b - the timer input clock is 211,875 kHz. 10 b - the timer input clock is an underflow of Timer0 11 b - the timer input clock is an underflow of Timer1

### 8.7.2.16 T3ReloadHi

High byte of the reload value of Timer3.

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### Table 112. T3ReloadHi register (address 1Fh);

Bit	7	6	5	4	3	2	1	0	
Symbol		T3ReloadHi							
Access rights				r/v	I				

### Table 113. T3ReloadHi bits

Bit	Symbol	Description
7 to 0	T3ReloadHi	Defines the high byte of the reload value of the Timer3. With the start event, the timer load the value of the T3ReloadValHi and T3ReloadValLo. Changing this register affects the timer only at the next start event.

### 8.7.2.17 T3ReloadLo

Low byte of the reload value of Timer3.

#### Table 114. T3ReloadLo register (address 20h)

Bit	7	6	5	4	3	2	1	0	
Symbol		T3ReloadLo							
Access rights				r/w	I				

#### Table 115. T3ReloadLo bits

Bit	Symbol	Description					
7 to 0	T3ReloadLo	Defines the low byte of the reload value of Timer3. With the start event, the timer load the value of the T3ReloadValHi and T3ReloadValLo. Changing this register affects the timer only at the next start event.					

#### 8.7.2.18 T3CounterValHi

High byte of the current counter value the 16-bit Timer3.

#### Table 116. T3CounterValHi register (address 21h)

Bit	7	6	5	4	3	2	1	0		
Symbol		T3CounterValHi								
Access rights				dy						

### Table 117. T3CounterValHi bits

Bit	Symbol	Description
7 to 0	T3Counter ValHi	High byte of the current counter value of Timer3. This value shall not be read out during reception.

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### 8.7.2.19 T3CounterValLo

Low byte of the current counter value the 16-bit Timer3.

### Table 118. T3CounterValLo register (address 22h)

Bit	7	6	5	4	3	2	1	0		
Symbol		T3CounterValLo								
Access rights				dy	,					

#### Table 119. T3CounterValLo bits

Tuble 110.	1000uniter van	
Bit	Symbol	Description
7 to 0	T3Counter ValLo	Low byte current counter value of Timer3. This value shall not be read out during reception.

### 8.7.2.20 T4Control

The wake-up timer T4 activates the system after a given time. If enabled, it can start the low-power card detection function.

<b>Table 120</b> .	<b>T4Control</b>	register	(address 23h)	
--------------------	------------------	----------	---------------	--

Bit	7	6	5	4	3	2	1	0
Symbol	T4Running	T4Start StopNow	T4Auto Trimm	T4Auto LPCD	T4Auto Restart	T4AutoWakeUp	T4Clk	
Access rights	dy	w	r/w	r/w	r/w	r/w	r	/w

#### Table 121. T4Control bits

Bit	Symbol	Description
7	T4Running	Shows if the timer T4 is running. If the bit T4StartStopNow is set, this bit and the timer T4 can be started/stopped.
6	T4Start StopNow	if set, the bit T4Running can be changed.
5	T4AutoTrimm	If set to one, the timer activates an LFO trimming procedure when it underflows. For the T4AutoTrimm function, at least one timer (T0 to T3) has to be configured properly for trimming (T3 is not allowed if T4AutoLPCD is set in parallel).
4	T4AutoLPCD	If set to one, the timer activates a low-power card detection sequence. If a card is detected an interrupt request is raised and the system remains active if enabled. If no card is detected the CLRC663 enters the Power down mode if enabled. The timer is automatically restarted (no gap). Timer 3 is used to specify the time where the RF field is enabled to check if a card is present. Therefore you may not use Timer 3 for T4AutoTrimm in parallel.

Bit	Symbol	Description
3	T4AutoRestart	Set to logic 1, the timer automatically restarts its countdown from T4ReloadValue, after the counter value has reached the value zero. Set to logic 0 the timer decrements to zero and stops. The bit Timer4IRQ is set to logic 1 at timer underflow.
2	T4AutoWakeUp	If set, the CLRC663 wakes up automatically, when the timer T4 has an underflow. This bit has to be set if the IC should enter the Power down mode after T4AutoTrimm and/or T4AutoLPCD is finished and no card has been detected. If the IC should stay active after one of these procedures, this bit has to be set to 0.
1 to 0	T4Clk	00b - the timer input clock is the LFO clock 01b - the timer input clock is the LFO clock/8 10b - the timer input clock is the LFO clock/16 11b - the timer input clock is the LFO clock/32

### 8.7.2.21 T4ReloadHi

High byte of the reload value of the 16-bit timer 4.

#### Table 122. T4ReloadHi register (address 24h)

Bit	7	6	5	4	3	2	1	0
Symbol		T4ReloadHi						
Access rights				r/w	I			

Table 123.	T4ReloadHi bi	ts
Bit	Symbol	Description
7 to 0	T4ReloadHi	Defines high byte for the reload value of timer 4. With the start event, the timer 4 loads the T4ReloadVal. Changing this register affects the timer only at the next start event.

#### 8.7.2.22 T4ReloadLo

Low byte of the reload value of the 16-bit timer 4.

#### Table 124. T4ReloadLo register (address 25h)

Bit	7	6	5	4	3	2	1	0		
Symbol		T4ReloadLo								
Access rights				r/w	I					

Table 125.	Table 125. T4ReloadLo bits								
Bit	Symbol	Description							
7 to 0	T4ReloadLo	Defines the low byte of the reload value of the timer 4. With the start event, the timer loads the value of the T4ReloadVal. Changing this register affects the timer only at the next start event.							

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### 8.7.2.23 T4CounterValHi

High byte of the counter value of the 16-bit timer 4.

Table 126. T4CounterValHi register (address 26h)										
Bit	7	7         6         5         4         3         2         1								
Symbol		T4CounterValHi								
Access rights				dy						

### Table 127. T4CounterValHi bits

Bit	Symbol	Description			
7 to 0	T4CounterValHi	High byte of the current counter value of timer 4.			

### 8.7.2.24 T4CounterValLo

Low byte of the counter value of the 16-bit timer 4.

Table 128.	T4CounterVa	ILo register	(address 27h)					
Bit	7	6	5	4	3	2	1	
Symbol				T4Counte	erValLo			
Access rights				dy	,			

#### Table 129. T4CounterValLo bits

Bit	Symbol	Description
7 to 0	T4CounterValLo	Low byte of the current counter value of the timer 4.

### 8.8 Transmitter driver configuration registers

### 8.8.1 TxMode

#### Table 130. TXMode register (address 28h)

Bit	7	6	5	4	3	2	1	0
Symbol	Tx2Inv	Tx1Inv	-	-	TxEn	TxClk Mode		
Access rights	r/w	r/w	RFU	RFU	r/w		r/w	

#### Table 131. TXMode bits

Bit	Symbol	Description
7	Tx2Inv	Inverts transmitter 2 at TX2 pin

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Bit	Symbol	Description
6	Tx1Inv	Inverts transmitter 1 at TX1 pin
5		RFU
4	-	RFU
3	TxEn	If set to 1 both transmitter pins are enabled
2 to 0	TxClkMode	Transmitter clock settings. Codes 011b and 0b110 are not supported. This register defines, if the output is operated in open-drain, push-pull, at high impedance or pulled to a fix high or low level.

### 8.8.2 TxAmp

With the set\_cw\_amplitude register, output power can be traded off against power supply rejection. Spending more headroom leads to better power supply rejection ration and better accuracy of the modulation degree.

With CwMax set, the voltage of TX1 will be pulled to the maximum possible. This register overrides the settings made by set\_cw\_amplitude.

Table 132. TxAmp register (address 29h)

Bit	7	6	5	4	3	2	1	0
Symbol	set_cw_a	mplitude	-		set_	residual_carri	er	
Access rights	r/w		RFU	r/w				

#### Table 133. TxAmp bits

	5. TXAIIIP DILS						
Bit	Symbol	Description					
7 to 6	set_cw_amplitude	Allows reducing the output amplitude of the transmitter by a fix value.					
		Four different preset values that are subtracted from TVDD can be selected:					
		0: TVDD -100 mV					
		1: TVDD -250 mV					
		2: TVDD -500 mV					
		3: T <sub>VDD</sub> -1000 mV					
5	RFU	-					
4 to 0	set_residual_ carrier	Set the residual carrier percentage. refer to section <u>Section 7.6.2</u> .					

### 8.8.3 TxCon

Table 134.	Table 134. TxCon register (address 2Ah)							
Bit	7	6	5	4	3	2	1	0
Symbol		Ove	rshootT2	CwMax	TxInv	T	Sel	
Access rights	r/w				r/w	r/w	I	/w

Bit	Symbol	Description
7 to 4	OvershootT2	Specifies the length (number of carrier clocks) of the additional modulation for overshoot prevention. Refer to section <u>Section 7.6.2.1</u> .
3	Cwmax	Set amplitude of continuous wave carrier to the maximum. If set, set_cw_amplitude in Register TxAmp has no influence on the continuous amplitude.
2	TxInv	If set, the resulting modulation signal defined by TxSel is inverted
1 to 0	TxSel	Defines which signal is used as source for modulation 00b no modulation 01b TxEnvelope 10b SigIn 11b RFU

### 8.8.4 Txl

Table 136. Txl register (address 2Bh)									
Bit	7	6	5	4	3	2	1	0	
Symbol	OvershootT1				tx_set_iLoad				
Access rights	r/w				r/w				

Bit	Symbol	Description						
7 to 4	OvershootT1	Overshoot value for Timer1. Refer to Section Section 7.6.2.1.						
3 to 0	tx_set_iLoad	Factory trim value, sets the expected Tx load current. This value is used to control the modulation index in an optimized way dependent on the expected TX load current.						

## 8.9 Transmitter CRC configuration registers

### 8.9.1 TxCrcPreset

<b>Table 138.</b>	<b>TXCrcPreset</b>	register	(address 2Ch)
-------------------	--------------------	----------	---------------

Bit	7	6	5	4	3	2	1	0
Symbol	RFU		TXPresetVal		TxCR	Ctype	TxCRCInvert	TxCRCEn
Access rights	-		r/w		r/\	N	r/w	r/w

#### Table 139. TxCrcPreset bits

Bit	Symbol	Description			
7	RFU	-			

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Bit	Symbol	Description
6 to 4	TXPresetVal	Specifies the CRC preset value for transmission (see following table).
3 to 2	TxCRCtype	Defines which type of CRC (CRC8/CRC16/CRC5) is calculated: • 00h CRC5 • 01h CRC8 • 02h CRC16 • 03h RFU
1	TxCRCInvert	if set, the resulting CRC is inverted and attached to the data frame (ISO/IEC 3309)
0	TxCRCEn	if set, a CRC is appended to the data stream

### Table 140. Transmitter CRC preset value configuration

TXPresetVal[64]	CRC16	CRC8	CRC5
0h	0000h	00h	00h
1h	6363h	12h	12h
2h	A671h	BFh	-
3h	FFFEh	FDh	-
4h	-	-	-
5h	-	-	-
6h	User defined	User defined	User defined
7h	FFFFh	FFh	1Fh

**Remark:** User-defined CRC preset values can be configured by EEPROM (see section <u>Section 7.7.2.1</u>, <u>Table 37</u>.

### 8.9.2 RxCrcCon

Table 141. RxCrcCon register (address 2Dh)										
Bit	7	6	5	4	3	2	1	0		
Symbol	RxForceCRCWrite	RXPresetVal			RXCRCtype		RxCRCInvert	RxCRCEn		
Access rights	r/w		r/w		r/w		r/w	r/w		

### Table 142. RxCrcCon bits

Bit	Symbol	Description
7	RxForceCrc Write	If set, the received CRC byte(s) are copied to the FIFO. If cleared CRC Bytes are only checked, but not copied to the FIFO. This bit has to be always set in case of a not byte aligned CRC (e.g. ISO/IEC 18000-3 mode 3/ EPC Class-1HF)
6 to 4	RXPresetVal	Defines the CRC preset value (Hex.) for transmission. (see following table).

Bit	Symbol	Description
3 to 2	RxCRCtype	Defines which type of CRC (CRC8/CRC16/CRC5) is calculated: • 00h CRC5 • 01h CRC8 • 02h CRC16 • 03h RFU
1	RxCrcInvert	If set, the CRC check is done for the inverted CRC.
0	RxCrcEn	If set, the CRC is checked and in case of a wrong CRC an error flag is set. Otherwise the CRC is calculated but the error flag is not modified.

### Table 143. Receiver CRC preset value configuration

RXPresetVal[64]	CRC16	CRC8	CRC5
0h	0000h	00h	00h
1h	6363h	12h	12h
2h	A671h	BFh	-
3h	FFFEh	FDh	-
4h	-	-	-
5h	-	-	-
6h	User defined	User defined	User defined
7h	FFFFh	FFh	1Fh

### 8.10 Transmitter data configuration registers

### 8.10.1 TxDataNum

Table 144.	Table 144. TxDataNum register (address 2Eh)									
Bit	7	6	5	4	3	2	1	0		
Symbol	RFU	RFU-	RFU-	KeepBitGrid	DataEn	TxLastBits				
Access rights				r/w	r/w		r/w			

### Table 145. TxDataNum bits

Tuble 140.							
Bit	Symbol	Description					
7 to 5	RFU	-					
4	KeepBitGrid	If set, the time between consecutive transmissions starts is a multiple of one ETU. If cleared, consecutive transmissions can even start within one ETU					
3	DataEn	If cleared - it is possible to send a single symbol pattern. If set - data is sent.					

Bit	Symbol	Description
2 to 0	TxLastBits	Defines how many bits of the last data byte to be sent. If set to 000b, all bits of the last data byte are sent. Note - bits are skipped at the end of the byte.
		Example - Data byte B2h (sent LSB first). TxLastBits = 011b (3h) => 010b (LSB first) is sent
		$TxLastBits = 110b (6h) \Rightarrow 010011b (LSB first) is sent$

### 8.10.2 TxDATAModWidth

Transmitter data modulation width register

#### Table 146. TxDataModWidth register (address 2Fh)

Bit	7	6	5	4	3	2	1	0
Symbol	DModWidth							
Access rights		r/w						

### Table 147. TxDataModWidth bits

Bit	Symbol	Description					
7 to 0	DModWidth	Specifies the length of a pulse for sending data with enabled pulse modulation. The length is given by the number of carrier clocks + 1.					
		A pulse can never be longer than from the start of the pulse to the end of the bit. The starting position of a pulse is given by the setting of TxDataMod.DPulseType. Note: This register is only used if Miller modulation (ISO/IEC 14443A PCD) is used. The settings are also used for the modulation width of start and/or stop symbols.					

### 8.10.3 TxSym10BurstLen

If a protocol requires a burst (an unmodulated subcarrier) the length can be defined with this TxSymBurstLen, the value high or low can be defined by TxSym10BurstCtrl.

Table 148. TxSym10BurstLen register (address 30h)									
Bit	7	6	5	4	3	2	1	0	
Symbol	RFU		Sym1Burst Len			Sym0Burst Len			
Access rights	-		r/w				r/w		

### Table 149. TxSym10BurstLen bits

Bit	Symbol	Description			
7	RFU	-			

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Bit	Symbol	Description
6 to 4	Sym1BurstLen	Specifies the number of bits issued for symbol 1 burst. The 3 bits encodes a range from 8 to 256 bit: 00h - 8bit 01h - 16bit 02h - 32bit 04h - 48bit 05h - 64bit 06h - 96bit 07h - 128bit 08h - 256bit
3	RFU	-
2 to 0	Sym0BurstLen	Specifies the number of bits issued for symbol 1 burst. The 3 bits encodes a range from 8 to 256 bit: 00h - 8bit 01h - 16bit 02h - 32bit 03h - 48bit 04h - 64bit 05h - 96bit 06h - 128bit 07h - 256bit

### 8.10.4 TxWaitCtrl

Table 150. TxWaitCtrl register (address 31h); res	set value: C0h
---	----------------

Bit	7	6	5	4	3	2	1	0
Symbol	TxWaitStart	TxWaitEtu	-	TxWait High		Tx	StopBitLen	gth
Access rights	r/w	r/w	r/w		r/w			

|--|

Dit	O	Description
Bit	Symbol	Description
7	TxWaitStart	If cleared, the TxWait time is starting at the End of the send data (TX). If set, the TxWait time is starting at the End of the received data (RX).
6	TxWaitEtu	If cleared, the TxWait time is TxWait × 16/13.56 MHz. If set, the TxWait time is TxWait × 0.5 / DBFreq (DBFreq is the frequency of the bit stream as defined by TxDataCon).
5 to 3	TxWait High	Bit extension of TxWaitLo. TxWaitCtrl bit 5 is MSB.

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Bit	Symbol	Description
2 to 0	TxStopBitLength	Defines stop-bits and EGT (= stop-bit + extra guard time EGT) to be sent:
		0h: no stop-bit, no EGT
		1h: 1 stop-bit, no EGT
		2h: 1 stop-bit + 1 EGT
		3h: 1 stop-bit + 2 EGT
		4h: 1 stop-bit + 3 EGT
		5h: 1 stop-bit + 4 EGT
		6h: 1 stop-bit + 5 EGT
		7h: 1 stop-bit + 6 EGT
		Note: This is only valid for ISO/IEC14443 Type B

### 8.10.5 TxWaitLo

Table 152.	152. TxWaitLo register (address 32h)							
Bit	7	7         6         5         4         3         2         1         0						
Symbol		TxWaitLo						
Access rights	r/w							

Table 153.	TxWaitLo bits	
Bit	Symbol	Description
7 to 0	TxWaitLo	Defines the minimum time between receive and send or between two send data streams Note: TxWait is a 11bit register (additional 3 bits are in the TxWaitCtrl register)! See also TxWaitEtu and TxWaitStart.

### 8.11 FrameCon

Table 154.	FrameCon	register	(address 33h)
------------	----------	----------	---------------

Bit	7	6	5	4	3	2	1	0
Symbol	TxParityEn	RxParityEn	-	-	StopS	Sym	Start	Sym
Access rights	r/w	r/w	RFU	RFU	r/w	1	r/v	W

### Table 155. FrameCon bits

Bit	Symbol	Description
7	TxParityEn	If set, a parity bit is calculated and appended to each byte transmitted.
6	RxParityEn	If set, the parity calculation is enabled. The parity is not transferred to the FIFO.

Bit	Symbol	Description
5 to 4	-	RFU
3 to 2	StopSym	<ul> <li>Defines which symbol is sent as stop-symbol:</li> <li>Oh: No symbol is sent</li> <li>1h: Symbol0 is sent</li> <li>2 h symbol1 is sent</li> <li>3h Symbol2 is sent</li> </ul>
1 to 0	StartSym	<ul> <li>Defines which symbol is sent as start-symbol:</li> <li>Oh: No Symbol is sent</li> <li>1h: Symbol0 is sent</li> <li>2 h: Symbol1 is sent</li> <li>3h: Symbol2 is sent</li> </ul>

### 8.12 Receiver configuration registers

### 8.12.1 RxSofD

Table 156. RxSofD register (address 34h)											
Bit	7	6	5	4	3	2	1	0			
Symbol	RF	U	SOF_En	SOFDetected	RFU	SubC_En	SubC_Detected	SubC_Present			
Access rights	-		r/w	dy	-	r/w	dy	r			

### Table 157. RxSofD bits

Bit	Symbol	Description
7 to 6	RFU	-
5	SOF_En	If set and a SOF is detected an RxSOFIRQ is raised.
4	SOF_Detected	Shows that a SOF is or was detected. Can be cleared by SW.
3	RFU	-
2	SubC_En	If set and a subcarrier is detected an RxSOFIRQ is raised.
1	SubC_Detected	Shows that a subcarrier is or was detected. Can be cleared by SW.
0	SubC_Present	Shows that a subcarrier is currently detected.

### 8.12.2 RxCtrl

Table 158. RxCtrl register (address 35h)											
Bit	7	6	5	4	3	2	1	0			
Symbol	RxAllowBits	RxMultiple	RxEOFType	EGT_Check	EMD_Sup	Baudrate					
Access rights	r/w	r/w	r/w	r/w	r/w		r/w				

Bit	Symbol	Description
7	RxAllowBits	If set, data is written into FIFO even if CRC is enabled, and no complete byte has been received.
6	RxMultiple	If set, RxMultiple is activated and the receiver will not terminate automatically (refer Section <u>Section 7.10.3.6</u> . If set to logic 1, at the end of a received data stream an error byte is added to the FIFO. The error byte is a copy of the Error register.
5 to 4	RFU	-
5	RxEOFType	<ul> <li>0: EOF as defined in the RxEOFSymbolReg is expected.</li> <li>1: ISO/IEC14443B EOF is expected.</li> <li>Note: Clearing this bit to 0 and clearing bit 0 and bit 1 in the RxEOFSymbolReg disables the EOF check.</li> </ul>
4	EGT_Check	If set to 1, the EGT is checked and if it is too long a protocol error is set. (This is only valid for ISO/IEC14443 Type B).
3	EMD_Sup	Enables the EMD suppression according to ISO/IEC14443. If an error occurs within the first three bytes, these three bytes are assumed to be EMD, ignored and the FIFO is reset. A collision is treated as an error as well If a valid SOF was received, the EMD_Sup is set and a frame of less than 3 bytes had been received. RX_IRQ is not set in this EMD error cases. If RxForceCRCWrite is set, the FIFO should not be read out before three bytes are written into.
2 to 0	Baudrate	Defines the baud rate of the receiving signal. 2h: 26 kBd 3h: 52 kBd 4h: 106 kBd 5h: 212 kBd 6h: 424 kBd 7h: 847 kBd all remaining values are RFU

### 8.12.3 RxWait

Selects internal receiver settings.

#### Table 160. RxWait register (address 36h)

Bit	7	6	5	4	3	2	1	0		
Symbol	RxWaitEtu		RxWait							
Access rights	r/w				r/w					

#### Table 161. RxWait bits

Bit	Symbol	Description
7	RXWaitEtu	If set to 0, the RxWait time is RxWait × 16/13.56 MHz. If set to 1, the RxWait time is RxWait × (0.5/DBFreq).
6 to 0	RxWait	Defines the time after sending, where every input is ignored.

### 8.12.4 RxThreshold

Selects minimum threshold level for the bit decoder.

#### Table 162. RxThreshold register (address 37h)

Bit	7	6	5	4	3	2	1	0	
Symbol		MinL	evel		MinLevelP				
Access rights	r/w					r/	W		

#### Table 163. RxThreshold bits

Bit	Symbol	Description					
7 to 4	MinLevel	Defines the MinLevel of the reception. Note: The MinLevel should be higher than the noise level in the system.					
3 to 0	MinLevelP	Defines the MinLevel of the phase shift detector unit.					

### 8.12.5 Rcv

#### Table 164. Rcv register (address 38h)

Bit	7	6	5	4	3	2	1	0
Symbol	Rcv_Rx_single	Rx_ADCmode	SigInSel		RFU		CollLevel	
Access rights	r/w	r/w	r/w	r/w			r	/w

#### Table 165. Rcv bits

Bit	Symbol	Description
7	Rcv_Rx_single	Single RXP Input Pin Mode; 0: Fully Differential 1: Quasi-Differential
6	Rx_ADCmode	Defines the operation mode of the Analog Digital Converter (ADC) 0: normal reception mode for ADC 1: LPCD mode for ADC
5 to 4	SigInSel	Defines input for the signal processing unit: 0h - idle 1h - internal analog block (RX) 2h - signal in over envelope (ISO/IEC14443A) 3h - signal in over s3c-generic
3 to 2	RFU	-
1 to 0	CollLevel	Defines the strength of a signal to be interpreted as a collision: 0h - Collision has at least 1/8 of signal strength 1h - Collision has at least 1/4 of signal strength 2h - Collision has at least 1/2 of signal strength 3h - Collision detection is switched off

### 8.12.6 RxAna

This register allows setting the gain (rcv\_gain) and high pass corner frequencies (rcv\_hpcf).

### Table 166. RxAna register (address 39h)

Bit	7	6	5	4	3	2	1	0
Symbol	VMid_	r_sel	RFU		rcv_hpcf		rcv_gain	
Access rights	r/w			-	r/v	V	r/	w

### Table 167. RxAna bits

Bit	Symbol	Description
7, 6	VMid_r_sel	Factory trim value, needs to be 0.
5, 4	RFU	
3, 2	rcv_hpcf	The rcv_hpcf [1:0] signals allow 4 different settings of the base band amplifier high pass cut-off frequency from ~40 kHz to ~300 kHz.
1 to 0	rcv_gain	With rcv_gain[1:0] four different gain settings from 30 dB and 60 dB can be configured (differential output voltage/differential input voltage).

#### Table 168. Effect of gain and high-pass corner register settings

rcv_gain (Hex.)	rcv_hpcf (Hex.)	fl (kHz)	fU (MHz)	gain (dB20)	bandwidth (MHz)
03	00	38	2.3	60	2.3
03	01	79	2.4	59	2.3
03	02	150	2.6	58	2.5
03	03	264	2.9	55	2.6
02	00	41	2.3	51	2.3
02	01	83	2.4	50	2.3
02	02	157	2.6	49	2.4
02	03	272	3.0	41	2.7
01	00	42	2.6	43	2.6
01	01	84	2.7	42	2.6
01	02	157	2.9	41	2.7
01	03	273	3.3	39	3.0
00	00	43	2.6	35	2.6
00	01	85	2.7	34	2.6
00	02	159	2.9	33	2.7
00	03	276	3.4	30	3.1

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### 8.13 Clock configuration

#### 8.13.1 SerialSpeed

This register allows setting speed of the RS232 interface. The default speed is set to 115.2 kbit/s. The transmission speed of the interface can be changed by modifying the entries for BR\_T0 and BR\_T1. The transfer speed can be calculated by using the following formulas:

 $BR_T0 = 0$ : transfer speed = 27.12 MHz / ( $BR_T1 + 1$ )

BR\_T0 > 0: transfer speed = 27.12 MHz / (BR\_T1 + 33) / 2^(BR\_T0 - 1)

The framing is implemented with 1 start bit, 8 data bits and 1 stop bit. A parity bit is not used. Transfer speeds above 1228.8 kbit/s are not supported.

Table 169.	SerialSpeed	register	(address3Bh):	reset value: 7Ah
	oonalopood	10giotor		

Bit	7	6	5	4	3	2	1	0
Symbol	BR_T0							
Access rights		r/w			r/w			

 Table 170.
 SerialSpeed bits

Bit	Symbol	Description
7 to 5	BR_T0	BR_T0 = 0: transfer speed = 27.12 MHz / (BR_T1 + 1) BR_T0 > 0: transfer speed = 27.12 MHz / (BR_T1 + 33) / 2^(BR_T0 - 1)
4 to 0	BR_T1	BR_T0 = 0: transfer speed = 27.12 MHz / (BR_T1 + 1) BR_T0 > 0: transfer speed = 27.12 MHz / (BR_T1 + 33) / 2^(BR_T0 - 1)

#### Table 171. RS232 speed settings

Transfer speed (kbit/s)	SerialSpeed register content (Hex.)
7.2	FA
9.6	EB
14.4	DA
19.2	СВ
38.4	AB
57.6	9A
115.2	7A
128.0	74
230.4	5A
460.8	3A
921.6	1C
1228.8	15

### 8.13.2 LFO\_Trimm

#### Table 172. LFO\_Trim register (address 3Ch)

Bit	7	6	5	4	3	2	1	0	
Symbol		LFO_trimm							
Access rights				r/w	I				

#### Table 173. LFO\_Trim bits

Bit	Symbol	Description
7 to 0	LFO_trimm	Trimm value. Refer to Section <u>Section 7.8.3</u> . Note: If the trimm value is increased, the frequency of the oscillator decreases.

### 8.13.3 PLL\_Ctrl Register

The PLL\_Ctrl register implements the control register for the IntegerN PLL. Two stages exist to create the ClkOut signal from the 27.12 MHz input. In the first stage, the 27.12 MHz input signal is multiplied by the value defined in PLLDiv\_FB and divided by two, and the second stage divides this frequency by the value defined by PLLDIV\_Out.

#### Table 174. PLL\_Ctrl register (address3Dh)

Bit	7	6	5	4	3	2	1	0
Symbol		Clk	OutSel		ClkOut_En	PLL_PD	PLLE	Div_FB
Access rights	r/w			r/w	r/w	r	/w	

#### Table 175. PLL\_Ctrl register bits

Bit	Symbol	Description
7 to 4	CLkOutSel	<ul> <li>0h - pin CLKOUT is used as I/O</li> <li>1h - pin CLKOUT shows the output of the analog PLL</li> <li>2h - pin CLKOUT is hold on 0</li> <li>3h - pin CLKOUT is hold on 1</li> <li>4h - pin CLKOUT shows 27.12 MHz from the crystal</li> <li>5h - pin CLKOUT shows 13.56 MHz derived from the crystal</li> <li>6h - pin CLKOUT shows 6.78 MHz derived from the crystal</li> <li>7h - pin CLKOUT shows 3.39 MHz derived from the crystal</li> <li>8h - pin CLKOUT is toggled by the Timer0 overflow</li> <li>9h - pin CLKOUT is toggled by the Timer1 overflow</li> <li>Ah - pin CLKOUT is toggled by the Timer2 overflow</li> <li>Bh - pin CLKOUT is toggled by the Timer3 overflow</li> <li>ChFh - RFU</li> </ul>
3	ClkOut_En	Enables the clock at Pin CLKOUT
2	PLL_PD	PLL power down
1-0	PLLDiv_FB	PLL feedback divider (see table 174)

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Table 176. Setting of feedback divider PLLDiv_FB [1:0]						
Bit 1	Bit 0	Division				
0	0	23 (VCO frequency 312 MHz)				
0	1	27 (VCO frequency 366 MHz)				
1	0	28 (VCO frequency 380 Mhz)				
1	1	23 (VCO frequency 312 Mhz)				

### Table 176 Setting of feedback divider BLL Div. EP [1:0]

### 8.13.4 PLLDiv\_Out

Table 177. PLLDiv_Out register (address 3Eh)											
Bit	7	7 6 5 4 3 2 1 0									
Symbol		PLLDiv_Out									
Access rights		r/w									

#### Table 178. PLLDiv Out bits

Bit	Symbol	Description				
7 to 0	PLLDiv_Out	PLL output divider factor; refer to Section Section 7.8.2.				

#### Table 179. Setting for the output divider ratio PLLDiv\_Out [7:0]

Value	Division
0	RFU
1	RFU
2	RFU
3	RFU
4	RFU
5	RFU
6	RFU
7	RFU
8	8
9	9
10	10
253	253
254	254

### 8.14 Low-power card detection configuration registers

The LPCD registers contain the settings for the low-power card detection. The setting for LPCD\_IMax (6 bits) is done by the two highest bits (bit 7, bit 6) of the registers LPCD\_QMin, LPCD\_QMax and LPCD\_IMin each.

### 8.14.1 LPCD\_QMin

Table 180. LPCD_QMin register (address 3Fh)								
Bit	7	6	5	4	3	2	1	0
Symbol	LPCD_IMax.5	LPCD_IMax.4			LPCD_Q	Min		
Access rights	r/w	r/w			r/w			

#### Table 181. LPCD\_QMin bits

Bit	Symbol	Description
7, 6	LPCD_IMax	Defines the highest two bits of the higher border for the LPCD. If the measurement value of the I channel is higher than LPCD_IMax, an LPCD interrupt request is indicated by bit IRQ0.LPCDIRQ.
5 to 0	LPCD_QMin	Defines the lower border for the LPCD. If the measurement value of the Q channel is higher than LPCD_QMin, an LPCDinterrupt request is indicated by bit IRQ0.LPCDIRQ.

### 8.14.2 LPCD\_QMax

#### Table 182. LPCD\_QMax register (address 40h)

		··· (··· ··· ···	<b>,</b>					
Bit	7	6	5	4	3	2	1	0
Symbol	LPCD_IMax.3	LPCD_IMax.2			LPCD_Q	Max		
Access rights	r/w	r/w			r/w			

### Table 183. LPCD\_QMax bits

10010 1001								
Bit	Symbol	Description						
7	LPCD_IMax.3	Defines the bit 3 of the high border for the LPCD. If the measurement value of the I channel is higher than LPCD IMax, an LPCD IRQ is raised.						
6	LPCD_IMax.2	Defines the bit 2 of the high border for the LPCD. If the measurement value of the I channel is higher than LPCD IMax, an LPCD IRQ is raised.						
5 to 0	LPCD_QMax	Defines the high border for the LPCD. If the measurement value of the Q channel is higher than LPCD QMax, an LPCD IRQ is raised.						

### 8.14.3 LPCD\_IMin

Table 184.	Table 184. LPCD_IMin register (address 41h)								
Bit	7	6	5	4	3	2	1	0	
Symbol	LPCD_IMax.1	LPCD_IMax.0	LPCD_IMin						
Access rights	r/w	r/w			r/w				

#### Table 185. LPCD\_IMin bits

Bit	Symbol	Description				
7 to 6	LPCD_IMax	Defines lowest two bits of the higher border for the low-power card detection (LPCD). If the measurement value of the I channel is higher than LPCD IMax, an LPCD IRQ is raised.				
5 to 0	LPCD_IMin	Defines the lower border for the low power card detection. If the measurement value of the I channel is lower than LPCD IMin, an LPCD IRQ is raised.				

### 8.14.4 LPCD\_Result\_I

Table 186. LPCD_Result_I register (address 42h)								
Bit	7	6	5	4	3	2	1	0
Symbol	RFU-	RFU-		LPCD_Result_I				
Access rights	-	-	r					

#### Table 187. LPCD\_I\_Result bits

B	Bit	Symbol	Description
7	' to 6	RFU	-
5	i to 0	LPCD_Result_I	Shows the result of the last low-power card detection (I-Channel).

### 8.14.5 LPCD\_Result\_Q

### Table 188. LPCD\_Result\_Q register (address 43h)

Bit	7	6	5	4	3	2	1	0
Symbol	RFU	LPCD_I RQ_Clr	LPCD_Reslult_Q					
Access rights	-	r/w	r					

#### Table 189. LPCD\_Q\_Result bits

Bit	Symbol	Description
7	RFU	-
6	LPCD_IRQ_Clr	If set no LPCD IRQ is raised any more until the next low-power card detection procedure. Can be used by software to clear the interrupt source.
5 to 0	LPCD_Result_Q	Shows the result of the last low power card detection (Q-Channel).

### 8.14.6 LPCD\_Options

This register is available on the CLRC66303 only. For silicon versions CLRC66301 and CLRC66302 this register on address 3AH is RFU.

Table 190. LPCD\_Options register (address 3Ah)

Bit	7	6	5	4	3	2	1	0
Symbol		RF	-		LPCD_TX_HIGH	LPCD_FILTER	LPCD_Q_ UNSTABLE	LPCD_I_UNSTABLE
Access rights					r/w	r/w	r	r

Table 191. LPCD_Options						
Bit	Symbol	Description				
7 to 4	RFU	-				
3	LPCD_TX_HIGH	If set, the TX-driver will be the same as $V_{TVDD}$ during LPCD. This will allow for a better LPCD detection range (higher transmitter output voltage) at the cost of a higher current consumption. If this bit is cleared, the output voltage at the TX drivers will be = $T_{VDD}$ - 0.4V. If this bit is set, the output voltage at the TX drivers will be = $V_{TVDD}$ .				
2	LPCD_FILTER	If set, The LPCD decision is based on the result of a filter which allows to remove noise from the evaluated signal in I and Q channel. Enabling LPCD_FILTER allows compensating for noisy conditions at the cost of a longer RF-ON time required for sampling. The total maximum LPCD sampling time is 4.72us.				
1	LPCD_Q_UNSTABLE	If bit 2 of this register is set, bit 1 indicates that the Q-channel ADC value was changing during the LPCD measuring time. Note: Only valid if LPCD_FILTER (bit 2) = 1. This information can be used by the host application for configuration of e.g. the threshold LPCD_QMax or inverting the TX drivers.				
0	LPCD_I_UNSTABLE	If bit 2 of this register is set, bit 0 Indicates that the I-channel ADC value was changing during the LPCD measuring time. Note: Only valid if LPCD_FILTER (bit2) = 1. This information can be used by the host application for configuration of e.g. the threshold LPCD_IMax or inverting the TX drivers.				

### 8.15 Pin configuration

### 8.15.1 PinEn

Table 192. PinEn register (address 44h)										
Bit	7	6	5	4	3	2	1	0		
Symbol	SIGIN_EN / OUT7	CLKOUT_EN / OUT6	IFSEL1_EN / OUT5	IFSEL0_EN / OUT4	TCK_EN / OUT 3	TMS_EN / OUT2	TDI_EN / OUT1	TDO_EN / OUT0		
Access rights	r/w	r/w	r/w	r/w	r/w	r/w	r/w	r/w		

Table 193.	Table 193. PinEn bits					
Bit	Symbol	Description				
7	SIGIN_EN / OUT7	Enables the output functionality on SIGIN (pin 5). The pin is then used as output.				
6	CLKOUT_EN / OUT6	Enables the output functionality of the CLKOUT (pin 22). The pin is then used as output. The CLKOUT function is switched off.				
5	IFSEL1_EN / OUT5	Enables the output functionality of the IFSEL1 (pin 27). The pin is then used as output.				
4	IFSEL0_EN / OUT4	Enables the output functionality of the IFSEL0 (pin 26). The pin is then used as output.				
3	TCK_EN / OUT3	Enables the output functionality of the TCK (pin 4) of the boundary scan interface. The pin is then used as output. If the boundary scan is activated in EEPROM, this bit has no function.				
2	TMS_EN / OUT2	Enables the output functionality of the TMS (pin 2) of the boundary scan interface. The pin is then used as output. If the boundary scan is activated in EEPROM, this bit has no function.				
1	TDI_EN / OUT1	Enables the output functionality of the TDI (pin 1) of the boundary scan interface. The pin is then used as output. If the boundary scan is activated in EEPROM, this bit has no function.				
0	TDO_EN / OUT0	Enables the output functionality of the TDO(pin 3) of the boundary scan interface. The pin is then used as output. If the boundary scan is activated in EEPROM, this bit has no function.				

### 8.15.2 PinOut

Table 194.	PinOut register	(address 45h)
	i mout register	(audiess +511)

Bit	7	6	5	4	3	2	1	0
Symbol	SIGIN_OUT	CLKOUT_OUT	IFSEL1_OUT	IFSEL0_OUT	TCK_OUT	TMS_OUT	TDI_OUT	TDO_OUT
Access rights	r/w	r/w	r/w	r/w	r/w	r/w	r/w	r/w

Bit	Symbol	Description
7	SIGIN_OUT	Output buffer of the SIGIN pin
6	CLKOUT_OUT	Output buffer of the CLKOUT pin
5	IFSEL1_OUT	Output buffer of the IFSEL1 pin
4	IFSEL0_OUT	Output buffer of the IFSEL0 pin
3	TCK_OUT	Output buffer of the TCK pin
2	TMS_OUT	Output buffer of the TMS pin
1	TDI_OUT	Output buffer of the TDI pin
0	TDO_OUT	Output buffer of the TDO pin

### 8.15.3 PinIn

### Table 196. PinIn register (address 46h)

Bit	7	6	5	4	3	2	1	0
Symbol	SIGIN_IN	CLKOUT_IN	IFSEL1_IN	IFSEL0_IN	TCK_IN	TMS_IN	TDI_IN	TDO_IN
Access rights	r	r	r	r	r	r	r	r

Table 197	Fable 197. PinIn bits					
Bit	Symbol	Description				
7	SIGIN_IN	Input buffer of the SIGIN pin				
6	CLKOUT_IN	Input buffer of the CLKOUT pin				
5	IFSEL1_IN	Input buffer of the IFSEL1 pin				
4	IFSEL0_IN	Input buffer of the IFSEL0 pin				
3	TCK_IN	Input buffer of the TCK pin				
2	TMS_IN	Input buffer of the TMS pin				
1	TDI_IN	Input buffer of the TDI pin				
0	TDO_IN	Input buffer of the TDO pin				

### 8.15.4 SigOut

### Table 198. SigOut register (address 47h)

Bit	7	6	5	4	3	2	1	0	
Symbol	Pad Speed	RFU			SigOutSel				
Access rights	r/w		-			r/w			

#### Table 199. SigOut bits

Bit	Symbol	Description
7	PadSpeed	If set, the I/O pins are supporting a fast switching mode. The fast mode for the I/O's will increase the peak current consumption of the device, especially if multiple I/Os are switching at the same time. The power supply needs to be designed to deliver this peak current.
6 to 4	RFU	-
3 to 0	SIGOutSel	0h, 1h - The pin SIGOUT is 3-state 2h - The pin SIGOUT is 0 3h - The pin SIGOUT is 1 4h - The pin SIGOUT shows the TX-envelope 5h - The pin SIGOUT shows the TX-active signal 6h - The pin SIGOUT shows the S3C (generic) signal 7h - The pin SIGOUT shows the RX-envelope (only valid for ISO/IEC 14443A, 106 kBd) 8h - The pin SIGOUT shows the RX-active signal 9h - The pin SIGOUT shows the RX-bit signal

### 8.16 Protocol configuration registers

### 8.16.1 TxBitMod

Table 200.	Table 200. TxBitMod register (address 48h)										
Bit	7	6	5	4	3	2	1	0			
Symbol	TxMSBFirst	RFU	TxParity Type	RFU	TxStopBitType	RFU	TxStartBitType	TxStartBitEn			
Access rights	r/w	-	r/w	-	r/w	-	r/w	r/w			

#### Table 201. TxBitMod bits

	TADILINOU DILS	
Bit	Symbol	Description
7	TxMSBFirst	If set, data is interpreted MSB first for data transmission. If cleared, data is interpreted LSB first.
6	RFU	-
5	TxParityType	Defines the type of the parity bit. If set to 1, odd parity is calculated, otherwise even parity is calculated.
4	RFU	-
3	TxStopBitType	Defines the type of the stop-bit (0b: logic zero / 1b: logic one).
2	RFU	-
1	TxStartBitType	Defines the type of the start-bit (0b: logic zero / 1b: logic one).
0	TxStartBitEn	If set to 1, a start-bit will be sent.

### 8.16.2 TxDataCon

Table 202.	able 202. TxDataCon (address 4Ah)							
Bit	7	7 6 5 4			3	2	1	0
Symbol	DCodeType			DSCFreq	DBFreq			
Access rights	r/w				r/w		r/w	

Table 203. TXDataCon bits	Table 203.	TxDataCon bits	
---------------------------	------------	----------------	--

Bit	Symbol	Description
7 to 4	DCodeType	Specifies the type of encoding of data to be used:0h - no special coding1h - collider datastream is decoded2h - RFU 3h - RFU4h - return to zero code - pulse at first position5h - return to zero code - pulse at 2nd position6h - return to zero code - pulse at 3rd position7h - return to zero code - pulse at 4th position8h - 1 out of 4 coding9h - 1 out of 256 code (range 0 - 255) [ICODE SLI]Ah - 1 out of 256 code (range 0 - 255; 00h is encoded with no modulation, value 256 not used) [ICODE 1]Bh - 1 out of 256 code (range 0 - 255; 00h is encoded with a pulse on last position) [ICODE quite value]Ch- Pulse internal encoded (PIE) [ISO/IEC 18000-3 mode 3/ EPC Class-1HF]Dh - RFUEh - RFUFh - RFU
3	DSCFreq	Specifies the subcarrier frequency of the used envelope. 0h - 424 kHz 1h - 848 kHz Note: This setting is only relevant, if an envelope is used which involves a subcarrier, e.g. Manchester with subcarrier coding.
2 to 0	DBFreq	Specifies the frequency of the bit stream: 0h - RFU 1h - RFU 2h - 26 kHz 3h - 53 kHz 4h - 106 kHz 5h - 212 kHz 6h - 424 kHz 7h - 848 kHz

### 8.16.3 TxDataMod

Bit	7	6	5	4	3	2	1	0
Symbol	Frame step	DMillerEn	DPulse	еТуре	DInvert		DEnvType	
Access rights	r/w	r/w	r/v	V	r/w		r/w	

Bit	Symbol	Description
7	Framestep	If set to 1, at every start of transmission, each byte of data is sent in a separate frame. SOF and EOF are appended to the data byte according to the framing settings. After one byte is transmitted, the TxEncoder waits for a new start trigger to continue with the next byte (trigger is generated automatically). If set to 0, transmission is done in the used way, where after a start trigger all data bytes are sent and the framing is done for the complete data stream only once.
6	DMillerEn	If set, pulse modulation is applied according to modified miller code. Note: This bit is intended to be set if DPulseType is 1h.
5 to 4	DPulseType	Specifies which type of pulse modulation is selected. 0 h - no pulse modulation 1h - pulse starts at beginning of bit 2h - pulse starts at beginning of second bit half 3h - pulse starts at beginning of third bit quarter Note: If DMillerEn is set, DPulseType must be set to 1h.
3	DInvert	If set the envelope of data is inverted.
2 to 0	DEnvType	Specifies the type of envelope used for transmission of data packets. The selected envelope type is applied to the pseudo bit stream. 0h - Direct output

1h - Manchester code

3h - BPSK

6h - RFU 7h - RFU

2h - Manchester code with subcarrier

4h - RZ (pulse of half bit length at beginning of second half of bit)

5h - RZ (pulse of half bit length at beginning of bit)

### 8.16.4 TxSymFreq

Table 206.	TxSymFreq (ad	dress 4Ch)						
Bit	7	6	5	4	3	2	1	0
Symbol	S32SCFreq	S32BFreq		S10SCFreq	S10BFreq			
Access rights	r/w		r/w		r/w	r/w		

Table 207.	le 207. TxSymFreq bits				
Bit	Symbol	Description			
7	S32SCFreq	Specifies the frequency of the subcarrier of symbol2 and symbol3: 0b 424 kHz 1b 848 kHz			

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Bit	Symbol	Description
6 to 4	S32BFreq	Specifies the frequency of the bit stream of symbol2 and symbol3: 000b RFU 001b RFU 010b 26 kHz 011b 53 kHz 100b 106 kHz 101b 212 kHz 110b 424 kHz 111b 848 kHz
3	S10SCFreq	Specifies the frequency of the subcarrier of symbol0 and symbol1: 0b424 kHz 1b848 kHz
2 to 0	S10BFreq	Specifies the frequency of the bit stream of symbol0 and symbol1: 000b RFU 001b RFU 010b 26 kHz 011b 53 kHz 100b 106 kHz 101b 212 kHz 110b 424 kHz 111b 848 kHz

### 8.16.5 TxSym0

The two Registers TxSym0H and TxSym0L create a 16-bit register that contains the pattern for Symbol0.

Table 208.	le 208. TxSym0H (address 4Dh)							
Bit	7	7 6 5 4 3 2 1 0						0
Symbol	Symbol0_H							
Access rights		r/w						

Table 209. TXSTWUH DIts	Table	e 209.	TxSYM0H bits
-------------------------	-------	--------	--------------

10010 2001		
Bit	Symbol	Description
7 to 0	Symbol0H	Higher 8 bits of symbol definition for Symbol0.

Table 210.	TxSym0L (ad	dress 4Eh)						
Bit	7	7 6 5 4 3 2 1						
Symbol	Symbol0_L							
Access rights	r/w							

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#### Table 211. TxSYM0L bits

Bit	Symbol	Description
7 to 0	Symbol0_L	Lower 8 bits of symbol definition for Symbol0.

### 8.16.6 TxSym1

The two Registers TxSym1H and TxSym1L create a 16 bit register that contains the pattern for Symbol1.

#### Table 212. TxSym1H (address 4Fh)

Bit	7	6	5	4	3	2	1	0
Symbol	Symbol1_H							
Access rights	r/w							

Table 213.	TxSym1H bits	
Bit	Symbol	Description
7 to 0	Symbol1_H	Higher 8 bits of symbol definition for Symbol1.

### Table 214. TxSym1L (address 50h)

Bit	7	6	5	4	3	2	1	0
Symbol		Symbol1_L						
Access rights	r/w							

#### Table 215. TxSym1L bits

Bit	Symbol	Description
7 to 0	Symbol1_L	Lower 8 bits of symbol definition for Symbol1.

### 8.16.7 TxSym2

Table 216.	TxSYM2 (add	ress 51h)							
Bit	7	7 6 5 4 3 2 1 0							
Symbol		Symbol2							
Access rights				r/w	1				

#### Table 217. TxSym2 bits

Bit	Symbol	Description
7 to 0	Symbol2	Symbol definition for Symbol2.

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## 8.16.8 TxSym3

Table 218.	TxSym3 (add	ress 52h)						
Bit	7	6	5	4	3	2	1	0
Symbol		Symbol3						
Access rights				r/w	1			

#### Table 219. TxSym3 bits

Bit	Symbol	Description
7 to 0	Symbol3	Symbol definition for Symbol3.

## 8.16.9 TxSym10Len

Table 220. TxSym10Len (address 53h)								
Bit	7	6	5	4	3	2	1	0
Symbol	Sym1Len			Sym0Len				
Access rights			r/w			r/w	,	

#### Table 221. TxSym10Len bits

Bit	Symbol	Description				
7 to 4	Sym1Len	Specifies the number of valid bits of the symbol definition of Symbol1. The range is from 1 bit (0h) to 16 bits (Fh).				
3 to 0	Sym0Len	Specifies the number of valid bits of the symbol definition of Symbol0. The range is from 1 bit (0h) to 16 bits (Fh).				

#### 8.16.10 TxSym32Len

#### Table 222. TxSym32Len (address 54h)

Bit	7	6	5	4	3	2	1	0
Symbol	RFU	Sym3Len			RFU	Sym2Len		
Access rights	-	r/w	r/w	r/w	-	r/w	r/w	r/w

#### Table 223. TxSym32Len bits

Bit	Symbol	Description
7	RFU	-
6 to 4	Sym3Len	Specifies the number of valid bits of the symbol definition of Symbol3. The range is from 1 bit (0h) to 8 bits (7h).
3	RFU	-

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Bit	Symbol	Description
2 to 0	Sym2Len	Specifies the number of valid bits of the symbol definition of Symbol2. The range is from 1 bit (0h) to 8 bits (7h).

## 8.16.11 TxSym10BurstCtrl

#### Table 224. TxSym10BurstCtrl register (address 55h)

Bit	7	6	5	4	3	2	1	0
Symbol	RFU	Sym1BurstType	Sym1BurstOnly	Sym1BurstEn	RFU	Sym0Burst Type	Sym0B urstOnly	Sym0B urstEn
Access rights	-	r/w	r/w	r/w	-	r/w	r/w	r/w

### Table 225. TxSym10BurstCtrl bits

Bit	Symbol	Description
7	RFU	-
6	Sym1BurstType	Specifies the type of the burst of Symbol1 (logical zero / logical one).
5	Sym1BurstOnly	If set to 1 Symbol1 consists only of a burst and no symbol pattern.
4	Sym1BurstEn	Enables the burst of symbol 1 of the length defined in TxSym10BurstLen.
3	RFU	-
2	Sym0BurstType	Specifies the type of the burst of symbol 0 (logical zero / logical one).
1	Sym0BurstOnly	If set to 1, symbol 0 consists only of a burst and no symbol pattern.
0	Sym0BurstEn	Enables the burst of symbol 0 of the length defined in TxSym10BurstLen.

#### 8.16.12 TxSym10Mod Reg

Table 226	TxSym10Mod	register	address 56	h)
	1 AO y III I U MIOU	register	(audi 633 30	,

Bit	7	6	5	4	3	2	1	0
Symbol	RFU	S10MillerEn	S10Pulse	Туре	S10Inv	S10E	InvType	
Access rights	-	r/w	r/w		r/w	r/w		

### Table 227. TxSym10Mod bits

Bit	Symbol	Description
7	RFU	-

Bit	Symbol	Description					
6	S10MillerEn	If set, pulse modulation is applied according to modified miller code. Note: This bit shall be set only if S10PulseType is set to 1h.					
5 to 4	S10PulseType	Specifies which type of pulse modulation is selected: 0h - no pulse modulation 1h - pulse starts at beginning of bit 2h - pulse starts at beginning of second bit half 3h - pulse starts at beginning of third bit quarter					
3	S10Inv	If set. the output of Symbol0 and Symbol1 is inverted.					
2 to 0	S10EnvType	Specifies the type of envelope used for transmission of Symbol0 and Symbol1. The pseudo bit stream is logically combined with the selected envelope type. Oh - Direct output 1h - Manchester code 2h - Manchester code with subcarrier 3h - BPKSK 4h - RZ return zero, pulse of half bit length at beginning of second half of bit 5h - RZ return zero, pulse of half bit length at beginning of second half of bit 6h - RFU 7h - RFU					

## 8.16.13 TxSym32Mod

#### Table 228. TxSym32Mod register (address 57h)

Bit	7	6	5	4	3	2	1	0
Symbol	RFU	S32MillerEn	S32PulseType		S32Inv	S32EnvType		
Access rights	-	r/w	r/w		r/w	r/w		

Table 229. TxSym32Mod bits
----------------------------

Bit	Symbol	cription				
BR	Cymbol					
7	RFU	-				
6	S32MillerEn	If set, pulse modulation is applied according to modified miller code. Note: This bit shall be set only if S32PulseType is set to 1h.				
5 to 4	S32PulseType	Specifies which type of pulse modulation is selected: 0h - no pulse modulation 1h - pulse starts at beginning of bit 2h - pulse starts at beginning of second bit half 3h - pulse starts at beginning of third bit quarter				
3	S32Inv	If set. the output of Symbol2 and Symbol3 is inverted.				

Bit	Symbol	Description
2 to 0	S32EnvType	Specifies the type of envelope used for transmission of symbol 0 and symbol 1. The bit stream is logically combined with the selected envelope type. Oh - Direct output 1h - Manchester code 2h - Manchester code with subcarrier 3h - BPSK 4h - RZ return zero, pulse of half bit length at beginning of second half of bit)
		5h - RZ return zero, pulse of half bit length at beginning of bit) 6h to 7h RFU

## 8.17 Receiver configuration

### 8.17.1 RxBitMod

#### Table 230. RxBitMod (address 58h)

Bit	7	6	5	4	3	2	1	0
Symbol	RFU	RFU	RxStopOnInvPar	RxStopOnLength	RxMSBFirst	RxStopBitEn	RxParityType	RFU
Access rights	-	-	r/w	r/w	r/w	r/w	r/w	-

#### Table 231. RxBitMod bits

Bit	Symbol	Description
7 to 6	RFU	-
5	RxStopOnInvPar	If set to 1, inverse parity bit is a stop condition.
4	RxStopOnLength	If set to 1, data reception stops when the number of received bytes reach the defined frame length. The value for the frame length is taken from the first data-byte received.
3	RxMSBFirst	If set to 1, data bytes are interpreted MSB first for data reception, which means data is converted at the CLCoPro interface. If this bit is set to 0, data is interpreted LSB first.
2	RxStopBitEn	If set, a stop-bit is expected and will be checked and extracted from data stream. Additionally on detection of a stop-bit a reset signal for the demodulator is generator to enable a resynchronization of the demodulator. If the expected stop-bit is incorrect, a frame error flag is set and the reception is aborted. Note: A stop bit is always considered to be a logic 1
1	RxParityType	Defines which type of the parity-bit is calculated: If cleared: Even parity If set: Odd parity
0	RFU	-

## 8.17.2 RxEofSym

Table 232. RxEofSym (address 59h)										
Bit	7	6	5	4	3	2	1	0		
Symbol		RxEOFSymbol								
Access rights		r/w								

#### Table 233. RxEOFSym bits

Bit	Symbol	Description
7 to 0	RxEOF Symbol	This value defines the pattern of the EOF symbol with a maximum length of 4 bit. Every tuple of 2 bits of the RxEOFSymbol encodes one bit of the EOF symbol. A 00 tuple closes the symbol. In this way symbols with less than 4 bits can be defined, starting with the bit0 and bit1. The leftmost active symbol pattern is processed first, which means the pattern is expected first. If the bit0 and bit1 are both zero, the EOF symbol is disabled. The following mapping is defined: 0h - no symbol bit 1h - zero value 2h - one value 3h - collision Example: 1Dh: Zero-Collision-Zero E8h: No symbol because two LSBits are zero

### 8.17.3 RxSyncValH

Table 234. RxSyncValH register (address5Ah)										
Bit	7	6	5	4	3	2	1	0		
Symbol		RxSyncValH								
Access rights		r/w								

#### Table 235. RxSyncValH bits

10010 200.	it to yric valit b		
Bit	Symbol	Description	
15 to 0	•	Defines the high byte of the Start Of Frame (SOF) pattern, which must be in front of the receiving data.	

## 8.17.4 RxSyncValL

Table 236. RxSyncValL register (address 5Bh)									
Bit	7	6	5	4	3	2	1	0	
Symbol		RxSyncValL							

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Bit	7	6	5	4	3	2	1	0
Access rights				r/w	I			

Table 237.	RxSyncValL b	its
Bit	Symbol	Description
7 to 0	RxSyncValL	Defines the low byte of the Start Of Frame (SOF) Pattern, which must be in front of the receiving data.

## 8.17.5 RxSyncMod

#### Table 238. RxSyncMode register (address 5Ch)

Bit	7	6	5	4	3	2	1	0	
Symbol	SyncLen				SyncNegEdge	LastSyncHalf	tSyncHalf SyncType		
Access rights	r/w		r/w	r/w	r	/w			

#### Table 239. RxSyncMod bits

Bit	Symbol	Description
7 to 4	SyncLen	Defines how many Bits of registers RxSyncValH and RxSyncValL are valid. For ISO/IEC 14443B set to 0.
3	SyncNegEdge	Is used for SOF with no correlation peak. The first negative edge of the correlation is used for defining the bit grid.
2	LastSyncHalf	The last Bit of the Sync mode has only half of the length compared to all other bits. (ISO/IEC 18000-3 mode 3/ EPC Class-1HF).
1 to 0	SyncType	<ul> <li>0: all 16 bits of SyncVal are interpreted as burst.</li> <li>1: a nibble of bits is interpreted as one bit in following way: {data, coll} data = zero or one; coll = 1 means a collision on this bit. Note: if Coll = 1 the value of data is ignored.</li> <li>2: the synchronization is done at every start bit of each byte (type B)</li> <li>3: RFU</li> </ul>

### 8.17.6 RxMod

Table 240.	RxMod reg	ister (address	5Dh)					
Bit	7	6	5	4	3	2	1	0
Symbol	RFU	RFU	PreFilter	RectFilter	SyncHigh	CorrInv	FSK	BPSK
Access rights	-	-	r/w	r/w	r/w	r/w	r/w	r/w

Table 241	I. RxMod bits	
Bit	Symbol	Description
7 to 6	-	RFU
5	PreFilter	If set 4 samples are combined to one data. (average).
4	RectFilter	If set, the ADC-values are changed to a more rectangular wave shape.
3	SyncHigh	Defines if the bit grid is fixed at maximum (1) or at minimum (0) value of the correlation.
2	CorrInv	Defines a logical for Manchester coding: 0: subcarrier / no subcarrier.
1	FSK	If set to 1, the demodulation scheme is set to FSK.
0	BPSK	If set to 1, the modulation scheme is BPSK.

## 8.17.7 RxCorr

#### Table 242. RxCorr register (address 5Eh)

Bit	7	6	5	4	3	2	1	0
Symbol	Corr	Freq	CorrSpeed		CorrLen	RFU		
Access rights	r/w	r/w	r/w	r/w	r/w		-	

## Table 243. RxCorr bits

Table 24	3. RXCorr Dits	
Bit	Symbol	Description
7, 6	CorrFreq	0h - 212 kHz 1h - 424 kHz 2h - 848 kHz 3h - 848 kHz
5, 4	CorrSpeed	Defines the number of clocks used for one correlation. 0h - ISO/IEC 14443 1h - ICODE 53 kBd, FeliCa 424 kBd 2h - ICODE 26 kBd, FeliCa 212 kBd 3h - RFU
3	CorrLen	Defines the length of the correlation data. (64 or 32 values). If set the lengths of the correlation data is 32 values. (ISO/IEC 18000-3 mode 3/ EPC Class-1HF, 2 Pulse Manchester 848 kHz subcarrier).
2 to 0	RFU	-

## 8.17.8 FabCali

Table 244.	FabCali	register	(address	5Fh)
------------	---------	----------	----------	------

Bit	7	6	5	4	3	2	1	0
Symbol				FabCali				

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Bit	7	6	5	4	3	2	1	0
Access rights				r/w				

Table 245.	FabCali bits	
Bit	Symbol	Description
7 to 0	FabCali	Fabrication calibration of the receiver. NOTE: do not change boot value.

## 8.18 Version register

## 8.18.1 Version

Table 246.	Version regis	ter (address	s 7Fh)								
Bit	7	6	5	4	3 2 1 0						
Symbol		V	ersion			SubVer	SubVersion				
Access rights			r			r					

Table 24	7. Version bits	
Bit	Symbol	Description
7 to 4	Version	Includes the version of the CLRC663 silicon.
		CLRC66301: 0x1
		CLRC66302: 0x1
		CLRC66303: 0x1
3 to 0	SubVersion	Includes the subversion of the CLRC663 silicon.
		CLRC66301: 0x8
		CLRC66302: 0x8 -No difference of the silicon between versions CLRC66301 and CLRC66302
		CLRC66303: 0xA
		LPCD_OPTIONS register had been added compared to earlier version of the CLRC663. Default configuration for LoadProtocol updated for improved performance. User EEPROM initialized with data. Transmitter driver allows higher I <sub>TVDD</sub> than lower SubVersions.

## 9 Limiting values

#### Table 248. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
V <sub>DD</sub>	supply voltage		-0.5	+ 6.0	V
V <sub>DD(PVDD)</sub>	PVDD supply voltage		-0.5	+ 6.0	V
V <sub>DD(TVDD)</sub>	TVDD supply voltage		-0.5	+ 6.0	V
I <sub>DD(TVDD)</sub>	TVDD supply current	CLRC66301, CLRC66302	-	250	mA
,		CLRC66303	-	500	
V <sub>i(RXP)</sub>	input voltage on pin RXP		-0.5	+ 2.0	V
V <sub>i(RXN)</sub>	input voltage on pin RXN		-0.5	+ 2.0	V
P <sub>tot</sub>	total power dissipation	per package	-	1125	mW
V <sub>ESD</sub> (HBM)	electrostatic discharge voltage	Human Body Model (HBM); 1500 Ω, 100 pF; JESD22-A114- B	-2000	2000	V
V <sub>ESD</sub> (CDM)	electrostatic discharge voltage	Charge Device Model (CDM);	-500	500	V
T <sub>j(max)</sub>	maximum junction temperature		-	125	°C
T <sub>stg</sub>	storage temperature	no supply voltage applied	-55	+150	°C

## **10** Recommended operating conditions

Exposure of the device to other conditions than specified in the Recommended Operating Conditions section for extended periods may affect device reliability.

Electrical parameters (minimum, typical and maximum) of the device are guaranteed only when it is used within the recommended operating conditions.

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>DD</sub>	supply voltage			3.0	5.0	5.5	V
V <sub>DD(TVDD)</sub>	TVDD supply voltage		[1]	3.0	5.0	5.5	V
V <sub>DD(PVDD)</sub>	PVDD supply voltage			3.0	5.0	5.5	V
T <sub>amb</sub>	operating ambient temperature	in still air with exposed pin soldered on a 4 layer JEDEC PCB		-25	+25	+85	°C
T <sub>stg</sub>	storage temperature	no supply voltage applied, relative humidity 4575%		-40	+25	+125	°C

Table 249. Operating conditions CLRC66301HN and CLRC66302HN

[1]  $V_{DD(PVDD)}$  must always be the same or lower than  $V_{DD}$ .

#### Table 250. Operating conditions CLRC66303HN

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
V <sub>DD</sub>	supply voltage			2.5	5.0	5.5	V
V <sub>DD(TVDD)</sub>	TVDD supply voltage		[1]	2.5	5.0	5.5	V
V <sub>DD(PVDD)</sub>	PVDD supply voltage	all host interfaces except I2C interface		2.5	5.0	5.5	V
		all host interfaces incl. I2C interface		3.0	5.0	5.5 5.5	V
T <sub>amb</sub>	operating ambient temperature	in still air with exposed pin soldered on a 4 layer JEDEC PCB		-40	+25	+105	°C
T <sub>stg</sub>	storage temperature	no supply voltage applied, relative humidity 4575%		-45	+25	+125	°C

 $\label{eq:VDD} \mbox{[1]} \quad \mbox{$V_{DD(PVDD)}$ must always be the same or lower than $V_{DD}$.}$ 

1.  $V_{DD(PVDD)}$  must always be the same or lower than  $V_{DD}$ .

## **11** Thermal characteristics

#### Table 251. Thermal characteristics

Symbol	Parameter	Conditions	Package	Тур	Unit
R <sub>th(j-a)</sub>	thermal resistance from junction to ambient	in still air with exposed pin soldered on a 4 layer JEDEC PCB	HVQFN32	40	K/W

## **12 Characteristics**

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
Current c	onsumption						
I <sub>DD</sub>	supply current	$I_{DD} = A_{VDD}+D_{VDD}$ ; modem on (transmitter and receiver are switched on)		-	17	20	mA
		$I_{DD} = A_{VDD}+D_{VDD}$ ; modem off (transmitter and receiver are switched off)		-	0.45	0.5	mA
I <sub>DD(PVDD)</sub>	PVDD supply current	no load on digital pins, leakage current only		-	0.5	5	μA
I <sub>DD(TVDD)</sub>	TVDD supply current	CLRC66301HN, CLRC66302HN		-	100	250	mA
		CLRC66303HN		-	250	350	mA
I <sub>pd</sub>	power-down current	All OUTx pins floating					
		ambient temp = +25 °C		-	40	400	nA
		ambient temp = -40°C +85°C		-	1.5	2.1	μA
		CLRC66303: ambient temp = +105 °C		-	3.5	5.2	μA
I <sub>stby</sub>	standby current	All OUTx pins floating					
		ambient temp = 25 °C, I <sub>VDD</sub> +I <sub>TVDD</sub> + I <sub>PVDD</sub>		-	3	6	μA
		ambient temp = -40°C +105°C, I <sub>stby</sub> = I <sub>VDD</sub> +I <sub>TVDD</sub> + I <sub>PVDD</sub>		-	5.25	26	
ILPCD(sleep)	LPCD sleep current	All OUTx pins floating			L		
		LFO active, no RF field on, ambient temp = 25 °C	[1]	-	3.3	6.3	μΑ
I <sub>L</sub> PCD(avera	g <mark>d</mark> ,PCD average current	All OUTx pins floating, TxLoad = 50 ohms. LPCD_FILTER = 0; Rfon duration = 10 us, RF-off duration 300ms; $V_{TVDD}$ = 3.0V; $T_{amb}$ = 25°C; $I_{LPCD}$ = $I_{VDD}+I_{TVDD}+I_{PVDD}$			I 		J
		LPCD_TX_HIGH = 0,		-	12	-	μA
		LPCD_TX_HIGH = 1		-	23	-	
t <sub>RFON</sub>	RF-on time during LPCD	LPCD_TX_HIGH = 0; TVDD=5.0 V T=25°C;		-	10	-	μs
		LPCD_TX_HIGH = 1;			50		μs

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Symbol	Parameter	Conditions	Min	Тур	Мах	Unit
CL	external buffer capacitor	AVDD	220	470	-	nF
CL	external buffer capacitor	DVDD	220	470	-	nF
IFSEL0/C	haracteristics SIGIN/OUT7, DUT4, IFSEL1/OUT5, TCK/C DO/OUT0, IRQ, IF0, IF1, IF2	OUT3, TMS/OUT2, TDI/				
ILI	input leakage current	output disabled	0.0	50	500	nA
V <sub>IL</sub>	low-level input voltage		-0.5	-	0.3 x V <sub>DD(PVDD)</sub>	V
V <sub>IH</sub>	high-level input voltage		0.7 x V <sub>DD(PVDD)</sub>	V <sub>DD(PVD</sub>	$DY_{DD(PVDD)} + 0.5$	V
V <sub>OL</sub>	low-level output voltage		0.0	0.0	0.4	V
V <sub>OH</sub>	high-level output voltage	If pins are used as output OUTx, I <sub>OH</sub> = 4 mA driving current for each pin	V <sub>DD(PVDD)</sub> -0.4	V <sub>DD(PVD</sub>	D¥DD(PVDD)	V
Ci	input capacitance		0.0	2.5	4.5	pF
Pin chara	acteristics PDOWN					
V <sub>IL</sub>	low-level input voltage		0.0	0.0	0.4	V
V <sub>IH</sub>	high-level input voltage		0.6 x V <sub>PVDD</sub>	V <sub>DD(PVD</sub>	DYDD(PVDD)	V
Pull-up r	esistance for TCK, TMS, T	DI, IF2				
R <sub>pu</sub>	pull-up resistance		50	72	120	KΩ
Pin chara	acteristics AUX 1, AUX 2					
Vo	output voltage		0.0	-	1.8	V
CL	load capacitance		0.0	-	400	pF
Pin chara	acteristics RXP, RXN					
V <sub>pp</sub>	input voltage		0	1.65	1.8	V
Ci	input capacitance		2	3.5	5	pF
V <sub>mod(pp)</sub>	modulation voltage	$V_{mod(pp)} = V_{i(pp)(max)} - V_{i(pp)}$ (min)	-	2.5	-	mV
Pins TX1	and TX2	· · · · · ·				_
Vo	output voltage		V <sub>ss(TVSS)</sub>	-	V <sub>DD(TVDD)</sub>	V
R <sub>o</sub>	output resistance	CLRC66301, CLRC66302: T=25°C, V <sub>DD(TVDD)</sub> = 5.0V	-	1.5	-	Ω
		CLRC66303: T=25°C, V <sub>DD(TVDD)</sub> = 5.0V	-	1.2	-	Ω
Clock fre	equency Pin CLKOUT		·			
f <sub>clk</sub>	clock frequency	configured to 27.12 MHz	-	27.12	-	MHz
δ <sub>clk</sub>	clock duty cycle		-	50	-	%
Crystal c	onnection XTAL1, XTAL2	·				
V <sub>o(p-p)</sub>	peak-to-peak output voltage	pin XTAL1	-	1.0	-	V
Vi	input voltage	pin XTAL1	0.0	-	1.8	V

## **NXP Semiconductors**

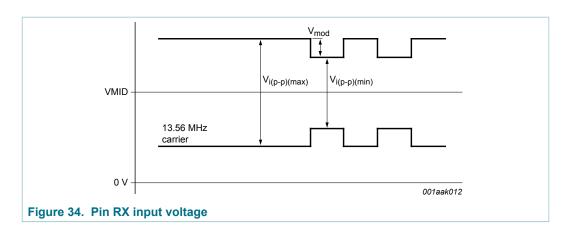
# CLRC663

## High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Symbol	Parameter	Conditions	Min	Тур	Мах	Unit
Ci	input capacitance	pin XTAL1	-	3	-	pF
Crystal r	equirements			1		
f <sub>xtal</sub>	crystal frequency	ISO/IEC14443 compliancy	27.12-14kHz	27.12	27.12+14kHz	MHz
ESR	equivalent series resistance		-	50	100	Ω
CL	load capacitance		-	10	-	pF
P <sub>xtal</sub>	crystal power dissipation		-	50	100	μW
Input cha	racteristics I/O Pin Charac	teristics IF3-SDA in I <sup>2</sup> C configu	uration			
ILI	input leakage current	output disabled	-	2	100	nA
V <sub>IL</sub>	LOW-level input voltage		-0.5	-	+0.3 V <sub>DD(PVDD)</sub>	V
VIH	HIGH-level input voltage		0.7 V <sub>DD(PVDD)</sub>	-	$V_{DD(PVDD)}$ + 0.5	V
V <sub>OL</sub>	LOW-level output voltage	I <sub>OL</sub> = 3 mA	-	-	0.3	V
I <sub>OL</sub>	LOW-level output current	V <sub>OL</sub> = 0.4 V; Standard mode, Fast mode	4	-	-	mA
		V <sub>OL</sub> = 0.6 V; Standard mode, Fast mode	6	-	-	mA
t <sub>f(0)</sub>	output fall time	Standard mode, Fast mode, C <sub>L</sub> < 400 pF	-	-	250	ns
		Fast mode +; C <sub>L</sub> < 550 pF	-	-	120	ns
t <sub>SP</sub>	pulse width of spikes that must be suppressed by the input filter		0	-	50	ns
Ci	input capacitance		-	3.5	5	pF
CL	load capacitance	Standard mode	-	-	400	pF
		Fast mode	-	-	550	pF
t <sub>EER</sub>	EEPROM data retention time	T <sub>amb</sub> = +55 °C	10	-	-	year
N <sub>EEC</sub>	EEPROM endurance (number of programming cycles)	under all operating conditions	5 x 10 <sup>5</sup>	-	-	cycle

 $\label{eq:lpd} [1] \quad I_{pd} \mbox{ is the total current for all supplies.}$ 

High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus



## **12.1 Timing characteristics**

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t <sub>SCKL</sub>	SCK LOW time		50	-	-	ns
t <sub>scкн</sub>	SCK HIGH time		50	-	-	ns
t <sub>h(SCKH-D)</sub>	SCK HIGH to data input hold time	SCK to changing MOSI	25	-	-	ns
t <sub>su(D-SCKH)</sub>	data input to SCK HIGH set- up time	changing MOSI to SCK	25	-	-	ns
t <sub>h(SCKL-Q)</sub>	SCK LOW to data output hold time	SCK to changing MISO	-	-	25	ns
t <sub>(SCKL-NSSH)</sub>	SCK LOW to NSS HIGH time		0	-	-	ns
t <sub>NSSH</sub>	NSS HIGH time	before communication	50	-	-	ns

#### Table 253. SPI timing characteristics

**Remark:** To send more bytes in one data stream, the NSS signal must be LOW during the send process. To send more than one data stream, the NSS signal must be HIGH between each data stream.

### Table 254. I<sup>2</sup>C-bus timing in fast mode and fast mode plus

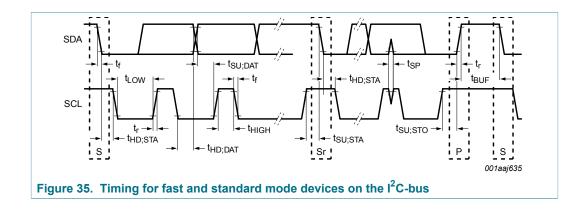
Symbol	Parameter	Conditions	Fast mode		Fast mode Plus		Unit
			Min	Max	Min	Max	
f <sub>SCL</sub>	SCL clock frequency		0	400	0	1000	kHz
t <sub>HD;STA</sub>	hold time (repeated) START condition	after this period, the first clock pulse is generated	600	-	260	-	ns
t <sub>SU;STA</sub>	set-up time for a repeated START condition		600	-	260	-	ns
t <sub>su;sто</sub>	set-up time for STOP condition		600	-	260	-	ns
t <sub>LOW</sub>	LOW period of the SCL clock		1300	-	500	-	ns
t <sub>HIGH</sub>	HIGH period of the SCL clock		600	-	260	-	ns
t <sub>HD;DAT</sub>	data hold time		0	900	-	450	ns

## **NXP Semiconductors**

# CLRC663

### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

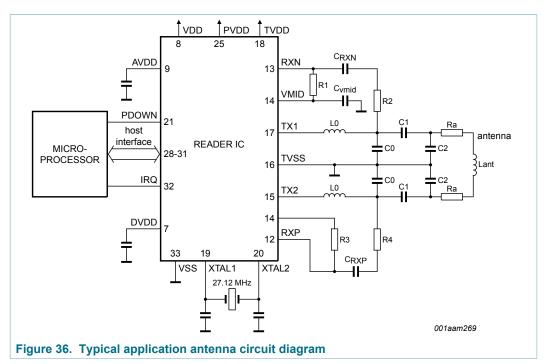
Symbol	Parameter	Conditions	Fast mode		Fast mode Plus		Unit
			Min	Max	Min	Max	
t <sub>SU;DAT</sub>	data set-up time		100	-	-	-	ns
t <sub>r</sub>	rise time	SCL signal	20	300	-	120	ns
t <sub>f</sub>	fall time	SCL signal	20	300	-	120	ns
t <sub>r</sub>	rise time	SDA and SCL signals	20	300	-	120	ns
t <sub>f</sub>	fall time	SDA and SCL signals	20	300	-	120	ns
t <sub>BUF</sub>	bus free time between a STOP and START condition		1.3	-	0.5	-	μs



## **13** Application information

A typical application diagram using a complementary antenna connection to the CLRC663 is shown in the following figure.

The antenna tuning and RF part matching is described in the application note [1] and [2].



## 13.1 Antenna design description

The matching circuit for the antenna consists of an EMC low pass filter (L0 and C0), a matching circuitry (C1 and C2), and a receiving circuits (R1 = R3, R2 = R4, C3 = C5 and C4 = C6;), and the antenna itself. The receiving circuit component values need to be designed for operation with the CLRC663. A re-use of dedicated antenna designs done for other products without adaptation of component values will result in degraded performance.

#### **13.1.1 EMC low pass filter**

The MIFARE product-based system operates at a frequency of 13.56 MHz. This frequency is derived from a quartz oscillator to clock the CLRC663 and is also the basis for driving the antenna with the 13.56 MHz energy carrier. This will not only cause emitted power at 13.56 MHz but will also emit power at higher harmonics. The international EMC regulations define the amplitude of the emitted power in a broad frequency range. Thus, an appropriate filtering of the output signal is necessary to fulfill these regulations.

**Remark:** The PCB layout has a major influence on the overall performance of the filter.

#### 13.1.2 Antenna matching

Due to the impedance transformation of the given low pass filter, the antenna coil has to be matched to a certain impedance. The matching elements C1 and C2 can be estimated and have to be fine-tuned depending on the design of the antenna coil.

The correct impedance matching is important to provide the optimum performance. The overall quality factor has to be considered to guarantee a proper ISO/IEC 14443 communication scheme. Environmental influences have to be considered as well as common EMC design rules.

For details, refer to the NXP application notes.

#### 13.1.3 Receiving circuit

The internal receiving concept of the CLRC663 makes use both side-bands of the subcarrier load modulation of the card response via a differential receiving concept (RXP, RXN). No external filtering is required.

It is recommended using the internally generated VMID potential as the input potential of pin RX. This DC voltage level of VMID has to be coupled to the Rx-pins via R2 and R4. To provide a stable DC reference voltage capacitances C4, C6 has to be connected between VMID and ground. Refer to Figure 36.

Considering the (AC) voltage limits at the Rx-pins the AC voltage divider of R1 + C3 and R2 as well as R3 + C5 and R4 has to be designed. Depending on the antenna coil design and the impedance matching, the voltage at the antenna coil varies from antenna design to antenna design. Therefore the recommended way to design the receiving circuit is to use the given values for R1(= R3), R2 (= R4), and C3 (= C5) from the above mentioned application note, and adjust the voltage at the RX-pins by varying R1(= R3) within the given limits.

Remark: R2 and R4 are AC-wise connected to ground (via C4 and C6).

#### 13.1.4 Antenna coil

The precise calculation of the antenna coils' inductance is not practicable but the inductance can be **estimated** using the following formula. We recommend designing an antenna either with a circular or rectangular shape.

$$L_1 = 2 \cdot I_1 \cdot \left( \ln \left( \frac{I_1}{D_1} \right) + - K \right) N_1^{1,8}$$

(4)

- I1 Length in cm of one turn of the conductor loop
- D1 Diameter of the wire or width of the PCB conductor respectively
- K Antenna shape factor (K = 1.07 for circular antennas and K = 1.47 for square antennas)
- L<sub>1</sub> Inductance in nH
- N<sub>1</sub> Number of turns
- Ln: Natural logarithm function

The actual values of the **antenna inductance**, **resistance**, **and capacitance at 13.56** MHz depend on various parameters such as:

- antenna construction (Type of PCB)
- thickness of conductor
- distance between the windings
- shielding layer
- metal or ferrite in the near environment

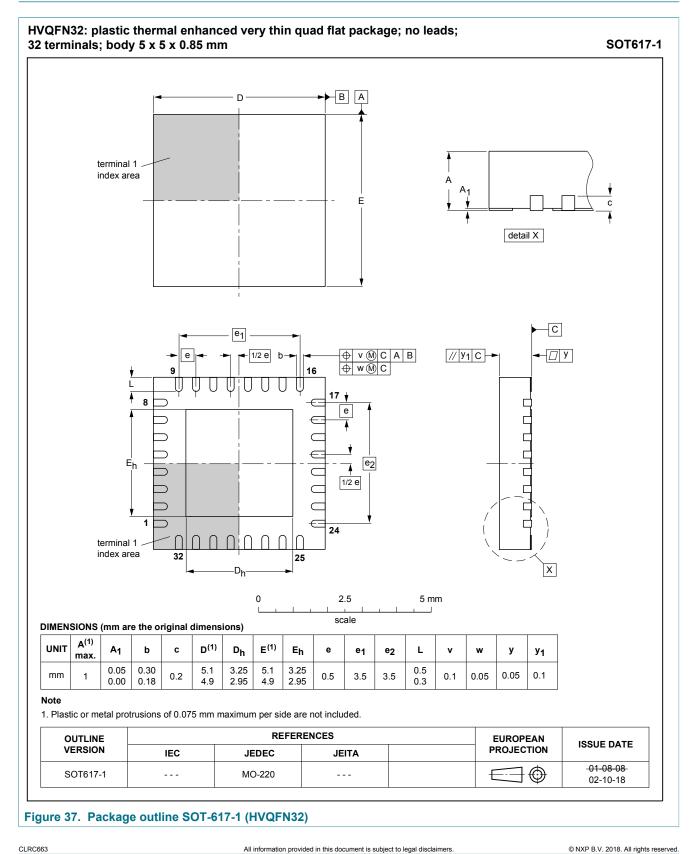
Therefore a measurement of those parameters under real life conditions, or at least a rough measurement and a tuning procedure are highly recommended to guarantee a reasonable performance. For details, refer to the above mentioned application notes.

## **NXP Semiconductors**

## **CLRC663**

High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

## 14 Package outline



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## **15 Handling information**

Moisture Sensitivity Level (MSL) evaluation has been performed according to *SNW*-FQ-225B rev.04/07/07 (*JEDEC J-STD-020C*). MSL for this package is level 2 which means 260 °C convection reflow temperature.

For MSL2:

- Dry pack is required.
- 1 year out-of-pack floor life at maximum ambient temperature 30°C/ 85 % RH.

For MSL1:

- No dry pack is required.
- No out-of-pack floor live spec. required.

## **16 Appendix**

### 16.1 LoadProtocol command register initialization

The RF configuration is loaded with the command Load Protocol. The tables below show the register configuration as performed by this command for each of the protocols. Antenna specific configurations are not covered by this register settings.

The CLRC66301 and CLRC66302 is not initialized for any antenna configuration. For this product the antenna configuration needs to be done by firmware.

The CLRC66303 antenna configuration in the user EEPROM is described in the chapter: "CLRC66303 EEprom configuration".

#### Table 255. ISO/IEC14443-A 106 / MIFARE Classic (Protocol Number 00)

Value (hex)
20
00
04
50
40
00
00
00
00
00
00
00
00
00
00
50
02
00
00
01
00
08
80
B2

Value for register	Value (hex)
TxBitMod	20
RFU	00
TxDataCon	05
TxDataMod	50
TxSymFreq	50
TxSym0H	00
TxSym0L	00
TxSym1H	00
TxSym1L	00
TxSym2	00
TxSym3	00
TxSym10Len	00
TxSym32Len	00
TxSym10BurstCtrl	00
TxSym10Mod	00
TxSym32Mod	50
RxBitMod	22
RxEofSym	00
RxSyncValH	00
RxSyncValL	00
RxSyncMod	00
RxMod	0D
RxCorr	80
FabCal	B2

Table 256. ISO/IEC14443-A 212/ MIFARE Classic (Protocol Number 01)

## Table 257. ISO/IEC14443-A 424/ MIFARE Classic (Protocol Number 02)

Value for register	Value (hex)
TxBitMod	20
RFU	00
TxDataCon	06
TxDataMod	50
TxSymFreq	60
TxSym0H	00
TxSym0L	00
TxSym1H	00
TxSym1L	00

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#### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	Value (hex)
TxSym2	00
TxSym3	00
TxSym10Len	00
TxSym32Len	00
TxSym10BurstCtrl	00
TxSym10Mod	00
TxSym32Mod	50
RxBitMod	22
RxEofSym	00
RxSyncValH	00
RxSyncValL	00
RxSyncMod	00
RxMod	0D
RxCorr	80
FabCal	B2

#### Table 258. ISO/IEC14443-A 848/ MIFARE Classic (Protocol Number 03)

Value for register	Value (hex)
TxBitMod	20
RFU	00
TxDataCon	07
TxDataMod	50
TxSymFreq	70
TxSym0H	00
TxSym0L	00
TxSym1H	00
TxSym1L	00
TxSym2	00
TxSym3	00
TxSym10Len	00
TxSym32Len	00
TxSym10BurstCtrl	00
TxSym10Mod	00
TxSym32Mod	50
RxBitMod	22
RxEofSym	00

Value for register	Value (hex)
RxSyncValH	00
RxSyncValL	00
RxSyncMod	00
RxMod	0D
RxCorr	80
FabCal	B2

#### Table 259. ISO/IEC14443-B 106 Classic (Protocol Number 04)

RFU         00           TxDataCon         04           TxDataMod         08           TxSymFreq         04           TxSym0H         00           TxSym0L         03           TxSym1H         00           TxSym1L         01           TxSym2         00           TxSym1L         01           TxSym3         00           TxSym10Len         AB           TxSym10Len         00           TxSym10Len         00           TxSym10Len         00           TxSym10Len         00           TxSym10Len         00           TxSym10Len         00           TxSym10Mod         00           TxSym10Mod         00           TxSym2Xod         00           RxEofSym         00           RxSyncValL         00           RxSyncValL         00           RxSyncMod         02           RxMod         02           RxCorr         80	Value for register	Value (hex)
TxDataCon04TxDataMod08TxDataMod04TxSymFreq04TxSym0H00TxSym0L03TxSym1H00TxSym1L01TxSym200TxSym300TxSym1OLenABTxSym10BurstCtrl00TxSym32Mod00RxBitMod04RxEofSym00RxSyncValL00RxSyncValL00RxSyncValL00RxSyncValL00RxSyncMod00RxSyncMod00RxSyncMod00RxSyncMod00RxSyncMod00RxSyncMod00RxSyncMod00RxSyncMod00RxSyncMod00RxSyncMod00RxSyncMod00RxSyncMod00RxSyncMod00RxSyncMod00RxSyncMod00RxSyncMod00RxCorr80	TxBitMod	09
TxDataMod08TxSymFreq04TxSym0H00TxSym0L03TxSym1H00TxSym1L01TxSym300TxSym300TxSym10LenABTxSym10BurstCtrl00TxSym10BurstCtrl00TxSym32Mod00RxBitMod04RxEofSym00RxSyncValL00RxSyncValL00RxSyncValL00RxSyncMod02RxMod02RxCorr80	RFU	00
TxSymFreq04TxSym0H00TxSym0L03TxSym1H00TxSym1L01TxSym200TxSym300TxSym10LenABTxSym10BurstCtrl00TxSym10BurstCtrl00TxSym2MOd00RxEofSym00RxEofSym00RxSyncValH00RxSyncValL00RxSyncValL00RxSyncValL00RxSyncValL00RxSyncValL00RxSyncValL02RxMod2LRC66301, CLRC66302: 1D CLRC66303: 0DRxCorr80	TxDataCon	04
TxSym0H         00           TxSym0L         03           TxSym1H         00           TxSym1L         01           TxSym2         00           TxSym3         00           TxSym1Len         AB           TxSym10Len         AB           TxSym10BurstCtrl         00           TxSym10Mod         08           TxSym32Mod         04           RxEofSym         00           RxSyncValH         00           RxSyncValL         00           RxCorr         80	TxDataMod	08
TxSym0L03TxSym1H00TxSym1L01TxSym200TxSym300TxSym10LenABTxSym10BurstCtrl00TxSym10Mod08TxSym32Mod00RxEofSym00RxSyncValH00RxSyncValL00RxSyncValL00RxSyncValL00RxSyncValL00RxSyncValL00RxSyncValL00RxSyncValL00RxSyncValL00RxSyncValL00RxSyncValL01RxMod2RxCorr80	TxSymFreq	04
TxSym1H00TxSym1L01TxSym200TxSym300TxSym10LenABTxSym10BurstCtrl00TxSym10BurstCtrl00TxSym32Mod00RxEofSym00RxSyncValH00RxSyncValL00RxSyncValL02RxModCLRC66301, CLRC66302: 1DRxCorr80	TxSym0H	00
TxSym1L         01           TxSym2         00           TxSym3         00           TxSym10Len         AB           TxSym32Len         00           TxSym10BurstCtrl         00           TxSym10Mod         08           TxSym32Mod         00           RxBitMod         04           RxEofSym         00           RxSyncValH         00           RxSyncValL         00           RxSyncMod         02           RxMod         2           RxMod         CLRC66301, CLRC66302: 1D           CLRC663003: 0D         RxCorr	TxSym0L	03
TxSym200TxSym300TxSym10LenABTxSym32Len00TxSym10BurstCtrl00TxSym10Mod08TxSym32Mod00RxBitMod04RxEofSym00RxSyncValH00RxSyncValL00RxMod22RxModCLRC66301, CLRC66302: 1DRxCorr80	TxSym1H	00
TxSym300TxSym10LenABTxSym32Len00TxSym10BurstCtrl00TxSym10Mod08TxSym32Mod00RxBitMod04RxEofSym00RxSyncValH00RxSyncValL00RxMod22RxModCLRC66301, CLRC66302: 1DRxCorr80	TxSym1L	01
TxSym10Len         AB           TxSym32Len         00           TxSym10BurstCtrl         00           TxSym10Mod         08           TxSym32Mod         00           RxBitMod         04           RxEofSym         00           RxSyncValH         00           RxSyncValL         00           RxSyncMod         02           RxMod         2           RxMod         CLRC66301, CLRC66302: 1D           CLRC663003: 0D         RxCorr	TxSym2	00
TxSym32Len00TxSym10BurstCtrl00TxSym10Mod08TxSym32Mod00RxBitMod04RxEofSym00RxSyncValH00RxSyncValL00RxMod02RxModCLRC66301, CLRC66302: 1DRxCorr80	TxSym3	00
TxSym10BurstCtrl00TxSym10Mod08TxSym32Mod00RxBitMod04RxEofSym00RxSyncValH00RxSyncValL00RxSyncMod2RxModCLRC66301, CLRC66302: 1DRxCorr80	TxSym10Len	AB
TxSym10Mod         08           TxSym32Mod         00           RxBitMod         04           RxEofSym         00           RxSyncValH         00           RxSyncValL         00           RxSyncMod         02           RxMod         CLRC66301, CLRC66302: 1D           CLRC663003: 0D         RxCorr	TxSym32Len	00
TxSym32Mod00RxBitMod04RxEofSym00RxSyncValH00RxSyncValL00RxSyncMod02RxModCLRC66301, CLRC66302: 1DRxCorr80	TxSym10BurstCtrl	00
RxBitMod         04           RxEofSym         00           RxSyncValH         00           RxSyncValL         00           RxSyncMod         02           RxMod         CLRC66301, CLRC66302: 1D           CLRC663003: 0D         RxCorr	TxSym10Mod	08
RxEofSym         00           RxSyncValH         00           RxSyncValL         00           RxSyncMod         02           RxMod         CLRC66301, CLRC66302: 1D           CLRC663003: 0D         RxCorr	TxSym32Mod	00
RxSyncValH         00           RxSyncValL         00           RxSyncMod         02           RxMod         CLRC66301, CLRC66302: 1D           CLRC663003: 0D         CLRC663003: 0D	RxBitMod	04
RxSyncValL         00           RxSyncMod         02           RxMod         CLRC66301, CLRC66302: 1D           CLRC663003: 0D         CLRC663003: 0D	RxEofSym	00
RxSyncMod         02           RxMod         CLRC66301, CLRC66302: 1D           CLRC663003: 0D         CLRC663003: 0D	RxSyncValH	00
RxMod         CLRC66301, CLRC66302: 1D           CLRC663003: 0D         CLRC663003: 0D	RxSyncValL	00
CLRC663003: 0D       RxCorr     80	RxSyncMod	02
RxCorr 80	RxMod	CLRC66301, CLRC66302: 1D
		CLRC663003: 0D
FabCal B2	RxCorr	80
	FabCal	B2

### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	Value (hex)
TxBitMod	09
RFU	00
TxDataCon	05
TxDataMod	08
TxSymFreq	05
TxSym0H	00
TxSym0L	03
TxSym1H	00
TxSym1L	01
TxSym2	00
TxSym3	00
TxSym10Len	AB
TxSym32Len	00
TxSym10BurstCtrl	00
TxSym10Mod	08
TxSym32Mod	00
RxBitMod	04
RxEofSym	00
RxSyncValH	00
RxSyncValL	00
RxSyncMod	02
RxMod	CLRC66301, CLRC66302: 1D
	CLRC66303: 0D
RxCorr	80
FabCal	B2

#### Table 261. ISO/IEC14443-B 424, (Protocol Number 06)

Value for register	Value (hex)
TxBitMod	09
RFU	00
TxDataCon	06
TxDataMod	08
TxSymFreq	06
TxSym0H	00
TxSym0L	03
TxSym1H	00

## High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	Value (hex)
TxSym1L	01
TxSym2	00
TxSym3	00
TxSym10Len	AB
TxSym32Len	00
TxSym10BurstCtrl	00
TxSym10Mod	08
TxSym32Mod	00
RxBitMod	04
RxEofSym	00
RxSyncValH	00
RxSyncValL	00
RxSyncMod	02
RxMod	CLRC66301, CLRC66302: 1D
	MFRC66303: 0D
RxCorr	80
FabCal	B2

#### Table 262. ISO/IEC14443-B 848, (Protocol Number 07)

Value for register	Value (hex)
TxBitMod	09
RFU	00
TxDataCon	07
TxDataMod	08
TxSymFreq	07
TxSym0H	00
TxSym0L	03
TxSym1H	00
TxSym1L	01
TxSym2	00
TxSym3	00
TxSym10Len	AB
TxSym32Len	00
TxSym10BurstCtrl	00
TxSym10Mod	08
TxSym32Mod	00

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Value for register	Value (hex)
RxBitMod	04
RxEofSym	00
RxSyncValH	00
RxSyncValL	00
RxSyncMod	02
RxMod	CLRC66301, CLRC66302: 1D
	CLRC66303: 0D
RxCorr	80
FabCal	B2

### Table 263. JIS X 6319-4 (FeliCa) 212, (Protocol Number 08)

Value for register	Value (hex)
TxBitMod	80
RFU	00
TxDataCon	05
TxDataMod	01
TxSymFreq	05
TxSym0H	B2
TxSym0L	4D
TxSym1H	00
TxSym1L	00
TxSym2	00
TxSym3	00
TxSym10Len	0F
TxSym32Len	00
TxSym10BurstCtrl	01
TxSym10Mod	01
TxSym32Mod	00
RxBitMod	18
RxEofSym	00
RxSyncValH	B2
RxSyncValL	4D
RxSyncMod	F0
RxMod	19
RxCorr	20
FabCal	В0

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### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Table 264. JIS X 6319-4 (FeliCa) 424, (Protocol Number 09)	
Value for register	Value (hex)
TxBitMod	80
RFU	00
TxDataCon	06
TxDataMod	01
TxSymFreq	06
TxSym0H	B2
TxSym0L	4D
TxSym1H	00
TxSym1L	00
TxSym2	00
TxSym3	00
TxSym10Len	0F
TxSym32Len	00
TxSym10BurstCtrl	01
TxSym10Mod	01
TxSym32Mod	00
RxBitMod	18
RxEofSym	00
RxSyncValH	B2
RxSyncValL	4D
RxSyncMod	F0
RxMod	19
RxCorr	50
FabCal	B0

## Table 264 US V 6240 4 (FaliCa) 424 (Protocol Number 00)

## Table 265. ISO/IEC15693 SLI 1/4 - SSC- 26, (Protocol Number 10)

Value for register	Value (hex)
TxBitMod	00
RFU	00
TxDataCon	83
TxDataMod	04
TxSymFreq	40
TxSym0H	00
TxSym0L	00
TxSym1H	00
TxSym1L	00

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### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	Value (hex)
TxSym2	84
TxSym3	02
TxSym10Len	00
TxSym32Len	37
TxSym10BurstCtrl	00
TxSym10Mod	00
TxSym32Mod	00
RxBitMod	00
RxEofSym	1D
RxSyncValH	00
RxSyncValL	01
RxSyncMod	00
RxMod	24
RxCorr	60
FabCal	F0

#### Table 266. ISO/IEC15693 SLI 1/4 - SSC- 53, (Protocol Number 11)

Value for register	Value (hex)
TxBitMod	00
RFU	00
TxDataCon	83
TxDataMod	04
TxSymFreq	40
TxSym0H	00
TxSym0L	00
TxSym1H	00
TxSym1L	00
TxSym2	84
TxSym3	02
TxSym10Len	00
TxSym32Len	37
TxSym10BurstCtrl	00
TxSym10Mod	00
TxSym32Mod	00
RxBitMod	00
RxEofSym	1D

Value for register	Value (hex)
RxSyncValH	00
RxSyncValL	01
RxSyncMod	00
RxMod	24
RxCorr	40
FabCal	F0

#### Table 267. SO/IEC15693 SLI 1/256 - DSC, (Protocol Number 12)

Value for register	Value (hex)
TxBitMod	00
RFU	00
TxDataCon	83
TxDataMod	04
TxSymFreq	40
TxSym0H	00
TxSym0L	00
TxSym1H	00
TxSym1L	00
TxSym2	81
TxSym3	02
TxSym10Len	00
TxSym32Len	37
TxSym10BurstCtrl	00
TxSym10Mod	00
TxSym32Mod	00
RxBitMod	00
RxEofSym	1D
RxSyncValH	00
RxSyncValL	01
RxSyncMod	00
RxMod	26
RxCorr	60
FabCal	F0

### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Table 268. EPC/UID - SSC -26, (Protocol Number 13)	
Value for register	Value (hex)
TxBitMod	80
RFU	00
TxDataCon	44
TxDataMod	00
TxSymFreq	44
TxSym0H	08
TxSym0L	22
TxSym1H	08
TxSym1L	28
TxSym2	8A
TxSym3	02
TxSym10Len	BB
TxSym32Len	37
TxSym10BurstCtrl	00
TxSym10Mod	00
TxSym32Mod	00
RxBitMod	08
RxEofSym	0B
RxSyncValH	00
RxSyncValL	00
RxSyncMod	08
RxMod	04
RxCorr	50
FabCal	F0

#### Table 268, EPC/UID - SSC -26, (Protocol Number 13)

### Table 269. EPC-V2 - 2/424 (Protocol Number 14)

Value for register	Value (hex)
TxBitMod	80
RFU	00
TxDataCon	C5
TxDataMod	00
TxSymFreq	05
TxSym0H	68
TxSym0L	41
TxSym1H	01
TxSym1L	A1

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### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	Value (hex)
TxSym2	00
TxSym3	00
TxSym10Len	8E
TxSym32Len	00
TxSym10BurstCtrl	00
TxSym10Mod	00
TxSym32Mod	00
RxBitMod	08
RxEofSym	0B
RxSyncValH	00
RxSyncValL	01
RxSyncMod	04
RxMod	0C
RxCorr	40
FabCal	F0

#### Table 270. EPC-V2 - 4/424, (Protocol Number 15)

Value for register Value (hex)	
TxBitMod	80
RFU	00
TxDataCon	C5
TxDataMod	00
TxSymFreq	05
TxSym0H	68
TxSym0L	41
TxSym1H	01
TxSym1L	A1
TxSym2	00
TxSym3	00
TxSym10Len	8E
TxSym32Len	00
TxSym10BurstCtrl	00
TxSym10Mod	00
TxSym32Mod	00
RxBitMod	08
RxEofSym	0B

### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	Value (hex)
RxSyncValH	00
RxSyncValL	01
RxSyncMod	04
RxMod	0C
RxCorr	50
FabCal	F0

#### Table 271. EPC-V2 - 2/848, (Protocol Number 16)

Value for register	Value (hex)
TxBitMod	80
RFU	00
TxDataCon	C5
TxDataMod	00
TxSymFreq	05
TxSym0H	68
TxSym0L	41
TxSym1H	01
TxSym1L	A1
TxSym2	00
TxSym3	00
TxSym10Len	8E
TxSym32Len	00
TxSym10BurstCtrl	00
TxSym10Mod	00
TxSym32Mod	00
RxBitMod	08
RxEofSym	0B
RxSyncValH	00
RxSyncValL	01
RxSyncMod	04
RxMod	0C
RxCorr	88
FabCal	F0

Table 272. EPC-V2 - 4/848, (Protocol Number 17)		
Value for register	Value (hex)	
TxBitMod	80	
RFU	00	
TxDataCon	C5	
TxDataMod	00	
TxSymFreq	05	
TxSym0H	68	
TxSym0L	41	
TxSym1H	01	
TxSym1L	A1	
TxSym2	00	
TxSym3	00	
TxSym10Len	8E	
TxSym32Len	00	
TxSym10BurstCtrl	00	
TxSym10Mod	00	
TxSym32Mod	00	
RxBitMod	08	
RxEofSym	0B	
RxSyncValH	00	
RxSyncValL	01	
RxSyncMod	04	
RxMod	0C	
RxCorr	80	
FabCal	F0	

#### Table 272. EPC-V2 - 4/848, (Protocol Number 17)

### Table 273. Jewel, (Protocol Number 18)

Value for register	Value (hex)
TxBitMod	00
RFU	00
TxDataCon	04
TxDataMod	D0
TxSymFreq	40
TxSym0H	00
TxSym0L	00
TxSym1H	00
TxSym1L	00

Value for register	Value (hex)
TxSym2	00
TxSym3	00
TxSym10Len	00
TxSym32Len	00
TxSym10BurstCtrl	00
TxSym10Mod	00
TxSym32Mod	50
RxBitMod	02
RxEofSym	00
RxSyncValH	00
RxSyncValL	01
RxSyncMod	00
RxMod	08
RxCorr	80
FabCal	F0

## 16.2 CLRC66303 EEPROM configuration

The SLRC61003 user EEPROM had been initialized with useful values for configuration of the chip using a typical 65x65mm antenna. This values stored in EEPROM can be used to configure the SLRC61003 with the command LoadReg.Typically, some of this entries will be required to be modified compared to the preset values to achieve the best RF performance for a specific antenna.

The registers 0x28...0x39 are relevant for configuration of the Antenna. For each supported protocol, a dedicated preset configuration is available. To ensure compatibility between products of the CLRC663 family, all products use the same default settings which are initialized in EEPROM, even if some of this protocols are not supported by the SLRC61003 product (e.g.ISO/IEC14443-A and B) and cannot be used.

Alternatively, the registers can be initialized by individual register write commands.

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	CO	8E
TxAmp	C1	12
DrvCon	C2	39
Txl	C3	0A
TXCrcPreset	C4	18
RXCrcPreset	C5	18
TxDataNum	C6	0F
TxModWidth	C7	21

#### Table 274. ISO/IEC14443-A 106 / MIFARE Classic

### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	EEPROM address (hex)	Value (hex)
TxSym10BurstLen	C8	00
TxWaitCtrl	C9	C0
TxWaitLo	CA	12
TxFrameCon	СВ	CF
RxSofD	CC	00
RxCtrl	CD	04
RxWait	CE	90
RxTreshold	CF	5C
Rcv	D0	12
RxAna	D1	0A

### Table 275. ISO/IEC14443-A 212/ MIFARE Classic

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	D4	8E
TxAmp	D5	D2
DrvCon	D6	11
Txl	D7	0A
TXCrcPreset	D8	18
RXCrcPreset	D9	18
TxDataNum	DA	0F
TxModWidth	DB	10
TxSym10BurstLen	DC	00
TxWaitCtrl	DD	C0
TxWaitLo	DE	12
TxFrameCon	DF	CF
RxSofD	E0	00
RxCtrl	E1	05
RxWait	E2	90
RxTreshold	E3	3C
Rcv	E4	12
RxAna	E5	0B

### Table 276. ISO/IEC14443-A 424/ MIFARE Classic

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	E8	8F
TxAmp	E9	DE

### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	EEPROM address (hex)	Value (hex)
DrvCon	EA	11
TxI	EB	0F
TXCrcPreset	EC	18
RXCrcPreset	ED	18
TxDataNum	EE	0F
TxModWidth	EF	07
TxSym10BurstLen	F0	00
TxWaitCtrl	F1	C0
TxWaitLo	F2	12
TxFrameCon	F3	CF
RxSofD	F4	00
RxCtrl	F5	06
RxWait	F6	90
RxTreshold	F7	2B
Rcv	F8	12
RxAna	F9	OB

#### Table 277. ISO/IEC14443-A 848/ MIFARE Classic

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	0100	8F
TxAmp	0101	DB
DrvCon	0102	21
Txl	0103	0F
TXCrcPreset	0104	18
RXCrcPreset	0105	18
TxDataNum	0106	0F
TxModWidth	0107	02
TxSym10BurstLen	0108	00
TxWaitCtrl	0109	C0
TxWaitLo	010A	12
TxFrameCon	010B	CF
RxSofD	010C	00
RxCtrl	010D	07
RxWait	010E	90
RxTreshold	010F	3A
Rcv	0110	12

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### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	EEPROM address (hex)	Value (hex)
RxAna	0111	0B

#### Table 278. ISO/IEC14443-B 106

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	0114	8F
TxAmp	0115	0E
DrvCon	0116	09
Txl	0117	0A
TXCrcPreset	0118	7B
RXCrcPreset	0119	7B
TxDataNum	011A	08
TxModWidth	011B	00
TxSym10BurstLen	011C	00
TxWaitCtrl	011D	01
TxWaitLo	011E	00
TxFrameCon	011F	05
RxSofD	0120	00
RxCtrl	0121	34
RxWait	0112	90
RxTreshold	0113	6F
Rcv	0114	12
RxAna	0115	03

#### Table 279. ISO/IEC14443-B 212

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	0128	8F
TxAmp	0129	0E
DrvCon	012A	09
TxI	012B	0A
TXCrcPreset	012C	7B
RXCrcPreset	012D	7B
TxDataNum	012E	08
TxModWidth	012F	00
TxSym10BurstLen	0130	00
TxWaitCtrl	0131	01
TxWaitLo	0132	00

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### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	EEPROM address (hex)	Value (hex)
TxFrameCon	0133	05
RxSofD	0134	00
RxCtrl	0135	35
RxWait	0136	90
RxTreshold	0137	3F
Rcv	0138	12
RxAna	0139	03

#### Table 280. ISO/IEC14443-B 424

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	0140	8F
TxAmp	0141	0F
DrvCon	0142	09
Txl	0143	0A
TXCrcPreset	0144	7B
RXCrcPreset	0145	7B
TxDataNum	0146	08
TxModWidth	0147	00
TxSym10BurstLen	0148	00
TxWaitCtrl	0149	01
TxWaitLo	014A	00
TxFrameCon	014B	05
RxSofD	014C	00
RxCtrl	014D	36
RxWait	014E	90
RxTreshold	014F	3F
Rcv	0150	12
RxAna	0151	03

### Table 281. ISO/IEC14443-B 848

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	0154	8F
TxAmp	0155	10
DrvCon	0156	09
Txl	0157	0A
TXCrcPreset	0158	7B

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### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	EEPROM address (hex)	Value (hex)
RXCrcPreset	0159	7B
TxDataNum	015A	08
TxModWidth	015B	00
TxSym10BurstLen	015C	00
TxWaitCtrl	015D	01
TxWaitLo	015E	00
TxFrameCon	015F	05
RxSofD	0160	00
RxCtrl	0161	37
RxWait	0162	90
RxTreshold	0163	3F
Rcv	0164	12
RxAna	0165	03

#### Table 282. JIS X 6319-4 (FeliCa) 212

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	0168	8F
TxAmp	0169	17
DrvCon	016A	01
Txl	016B	06
TXCrcPreset	016C	09
RXCrcPreset	016D	09
TxDataNum	016E	08
TxModWidth	016F	00
TxSym10BurstLen	0170	03
TxWaitCtrl	0171	80
TxWaitLo	0172	12
TxFrameCon	0173	01
RxSofD	0174	00
RxCtrl	0175	05
RxWait	0176	86
RxTreshold	0177	3F
Rcv	0178	12
RxAna	0179	02

### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	0180	8F
TxAmp	0181	17
DrvCon	0182	01
Txl	0183	06
TXCrcPreset	0184	09
RXCrcPreset	0185	09
TxDataNum	0186	08
TxModWidth	0187	00
TxSym10BurstLen	0188	03
TxWaitCtrl	0189	80
TxWaitLo	018A	12
TxFrameCon	018B	01
RxSofD	018C	00
RxCtrl	018D	06
RxWait	018E	86
RxTreshold	018F	3F
Rcv	0190	12
RxAna	0191	02

### Table 283. JIS X 6319-4 (FeliCa) 424

#### Table 284. ISO/IEC15693 SLI 1/4 - SSC- 26

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	0194	89
TxAmp	0195	10
DrvCon	0196	09
Txl	0197	0A
TXCrcPreset	0198	7B
RXCrcPreset	0199	7B
TxDataNum	019A	08
TxModWidth	019B	00
TxSym10BurstLen	019C	00
TxWaitCtrl	019D	88
TxWaitLo	019E	A9
TxFrameCon	019F	0F
RxSofD	01A0	00
RxCtrl	01A1	02
RxWait	01A2	9C

Value for register	EEPROM address (hex)	Value (hex)
RxTreshold	01A3	74
Rcv	01A4	12
RxAna	01A5	07

#### Table 285. ISO/IEC15693 SLI 1/4 - SSC-53

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	01A8	89
TxAmp	01A9	10
DrvCon	01AA	09
Txl	01AB	0A
TXCrcPreset	01AC	7B
RXCrcPreset	01AD	7B
TxDataNum	01AE	08
TxModWidth	016F	00
TxSym10BurstLen	01B0	00
TxWaitCtrl	01B1	88
TxWaitLo	01B2	A9
TxFrameCon	01B3	0F
RxSofD	01B4	00
RxCtrl	01B5	03
RxWait	01B6	9C
RxTreshold	01B7	74
Rcv	01B8	12
RxAna	01B9	03

#### Table 286. ISO/IEC15693 SLI 1/256 - DSC

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	01C0	8E
TxAmp	01C1	10
DrvCon	01C2	01
Txl	01C3	06
TXCrcPreset	01C4	7B
RXCrcPreset	01C5	7B
TxDataNum	01C6	08
TxModWidth	01C7	00
TxSym10BurstLen	01C8	00

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### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	EEPROM address (hex)	Value (hex)
TxWaitCtrl	01C9	88
TxWaitLo	01CA	A9
TxFrameCon	01CB	0F
RxSofD	01CC	00
RxCtrl	01CD	02
RxWait	01CE	10
RxTreshold	01CF	44
Rcv	01D0	12
RxAna	01D1	06

#### Table 287. EPC/UID - SSC -26

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	01D4	8F
TxAmp	01D5	10
DrvCon	01D6	01
Txl	01D7	06
TXCrcPreset	01D8	74
RXCrcPreset	01D9	7B
TxDataNum	01DA	18
TxModWidth	01DB	00
TxSym10BurstLen	01DC	00
TxWaitCtrl	01DD	50
TxWaitLo	01DE	5C
TxFrameCon	01DF	0F
RxSofD	01E0	00
RxCtrl	01E1	03
RxWait	01E2	10
RxTreshold	01E3	4E
Rcv	01E4	12
RxAna	01E5	06

#### Table 288. EPC-V2 - 2/424

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	01E8	8F
TxAmp	01E9	10
DrvCon	01EA	09

### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	EEPROM address (hex)	Value (hex)
Txl	01EB	0A
TXCrcPreset	01EC	11
RXCrcPreset	01ED	91
TxDataNum	01EE	09
TxModWidth	01EF	00
TxSym10BurstLen	01F0	00
TxWaitCtrl	01F1	80
TxWaitLo	01F2	12
TxFrameCon	01F3	01
RxSofD	01F4	00
RxCtrl	01F5	03
RxWait	01F6	A0
RxTreshold	01F7	56
Rcv	01F8	12
RxAna	01F9	OF

### Table 289. EPC-V2 - 4/424

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	0200	8F
TxAmp	0201	10
DrvCon	0202	09
Txl	0203	0A
TXCrcPreset	0204	11
RXCrcPreset	0205	91
TxDataNum	0206	09
TxModWidth	0207	00
TxSym10BurstLen	0208	00
TxWaitCtrl	0209	80
TxWaitLo	020A	12
TxFrameCon	020B	01
RxSofD	020C	00
RxCtrl	020D	03
RxWait	020E	A0
RxTreshold	020F	56
Rcv	0210	12
RxAna	0211	0F

#### Table 290. EPC-V2 - 2/848

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	0214	8F
TxAmp	0215	D0
DrvCon	0216	01
Txl	0217	0A
TXCrcPreset	0218	11
RXCrcPreset	0219	91
TxDataNum	021A	09
TxModWidth	021B	00
TxSym10BurstLen	021C	00
TxWaitCtrl	021D	80
TxWaitLo	021E	12
TxFrameCon	021F	01
RxSofD	0220	00
RxCtrl	0221	05
RxWait	0222	A0
RxTreshold	0223	26
Rcv	0224	12
RxAna	0225	0E

### Table 291. EPC-V2 - 4/848

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	0228	8F
TxAmp	0229	D0
DrvCon	022A	01
TxI	022B	0A
TXCrcPreset	022C	11
RXCrcPreset	022D	91
TxDataNum	022E	09
TxModWidth	022F	00
TxSym10BurstLen	0230	00
TxWaitCtrl	0231	80
TxWaitLo	0232	12
TxFrameCon	0233	01
RxSofD	0234	00
RxCtrl	0235	05
RxWait	0236	A0

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Value for register	EEPROM address (hex)	Value (hex)
RxTreshold	0237	26
Rcv	0238	12
RxAna	0239	0E

#### Table 292. Jewel

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	0240	8E
TxAmp	0241	15
DrvCon	0242	11
Txl	0243	06
TXCrcPreset	0244	18
RXCrcPreset	0245	18
TxDataNum	0246	0F
TxModWidth	0247	20
TxSym10BurstLen	0248	00
TxWaitCtrl	0249	40
TxWaitLo	024A	09
TxFrameCon	024B	4F
RxSofD	024C	00
RxCtrl	024D	04
RxWait	024E	8F
RxTreshold	024F	32
Rcv	0250	12
RxAna	0251	0A

#### Table 293. ISO/IEC14443 - B 106 EMVCo Optimized

Value for register	EEPROM address (hex)	Value (hex)
DrvMode	0254	8F
TxAmp	0255	0E
DrvCon	0256	09
TxI	0257	0A
TXCrcPreset	0258	7B
RXCrcPreset	0259	7B
TxDataNum	025A	08
TxModWidth	025B	00
TxSym10BurstLen	025C	00

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### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Value for register	EEPROM address (hex)	Value (hex)
TxWaitCtrl	025D	01
TxWaitLo	025E	00
TxFrameCon	025F	05
RxSofD	0260	00
RxCtrl	0261	34
RxWait	0262	90
RxTreshold	0263	9F
Rcv	0264	12
RxAna	0265	03

## **17 Abbreviations**

Table 294. Abbre	viations
Acronym	Description
ADC	Analog-to-Digital Converter
BPSK	Binary Phase Shift Keying
CRC	Cyclic Redundancy Check
CW	Continuous Wave
EGT	Extra Guard Time
EMC	Electro Magnetic Compatibility
EMD	Electro Magnetic Disturbance
EOF	End Of Frame
EPC	Electronic Product Code
ETU	Elementary Time Unit
GPIO	General Purpose Input/Output
HBM	Human Body Model
I <sup>2</sup> C	Inter-Integrated Circuit
IRQ	Interrupt Request
LFO	Low Frequency Oscillator
LPCD	Low-Power Card Detection
LSB	Least Significant Bit
MISO	Master In Slave Out
MOSI	Master Out Slave In
MSB	Most Significant Bit
NRZ	Not Return to Zero
NSS	Not Slave Select
PCD	Proximity Coupling Device
PLL	Phase-Locked Loop
RZ	Return To Zero
RX	Receiver
SAM	Secure Access Module
SOF	Start Of Frame
SPI	Serial Peripheral Interface
SW	Software
TTimer	Timing of the clk period
ТХ	Transmitter
UART	Universal Asynchronous Receiver Transmitter
UID	Unique Identification

## **NXP Semiconductors**

# CLRC663

### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Acronym	Description
VCO	Voltage Controlled Oscillator

## **18 References**

[1]

### Application note AN11019

CLRC663, MFRC630, MFRC631, SLRC610 Antenna Design Guide

[2]

Application note AN11783

CLRC663 plus Low Power Card Detection

## **19 Revision history**

Document ID	Release date	Data sheet status	Change notice	Supersedes		
CLRC663 v.4.6	20180516	Product data sheet	-	CLRC663 v.4.5		
Modifications:	<ul><li>Load protocol</li><li>Editorial updat</li></ul>	command register headlines utes	pdated, added proto	ocol number in headlines		
CLRC663 v.4.5	20171219	Product data sheet	-	CLRC663 v.4.4		
Modifications:	Deleted refere	nces to MICORE Application r	notes			
CLRC663 v.4.4	20170502	Product data sheet	-	CLRC663 v.4.3		
Modifications:	<ul> <li>Data sheet state</li> <li>V<sub>ESD</sub>(HBM) love</li> <li>Appendix 15.2</li> </ul>	<ul> <li>Section 2 "Features and benefits": updated</li> <li>Data sheet status reset to "Product" and "Company Public"</li> <li>V<sub>ESD</sub>(HBM) low value corrected.</li> <li>Appendix 15.2. RX_MOD default setting changed for Type B @ 106: 1D (66301, 66302) changed to 0D (66303).</li> </ul>				
CLRC663 v.4.3	20170220	Objective data sheet	-	CLRC663 v.4.2		
Modifications:	<ul><li>New product v</li><li>Data sheet state</li></ul>	<ul> <li><u>Section 4 "Ordering information"</u>: updated</li> <li>New product version CLRC66303 added</li> <li>Data sheet status changed to "Objective" and "Company Confidential"</li> <li>Revision only available in DocStore</li> </ul>				
CLRC663 v.4.2	20160427	Product data sheet	-	CLRC663 v.4.1		
Modifications:	Descriptive titl	e changed				
CLRC663 v.4.1	20160211	Product data sheet	-	CLRC663 v.4.0		
Modifications:	removed	<ul> <li><u>Table 1 "Quick reference data CLRC66301HN and CLRC66302HN"</u>: Table notes [3] and [4 removed</li> <li><u>Table 252 "Characteristics"</u>: TVDD supply current value updated</li> </ul>				
CLRC663 v.4.0	20151029	Product data sheet	-	CLRC663 v.3.9		
Modifications:	– I <sub>DD(TVDD)</sub> ma	aracteristics" OVDD min and max values add ax value updated to 250 mA <u>niting values"</u> : ITVDD max valu		A		
CLRC663 v.3.9	20150722	Product data sheet	-	CLRC663 v.3.8		
Modifications:	<ul> <li>Section 2 "Fea</li> <li>Figure 13 "Control</li> </ul>	neral description": updated atures and benefits": updated nnection to host with SPI": upd gister read and write access":				
CLRC663 v.3.7	20140204	Product data sheet	-	CLRC663 v.3.6		
Modifications:	<ul> <li>Typing error of</li> <li>WaterLevel ar</li> <li>WaterLevel ar</li> <li>Waterlevel Re</li> <li>FIFOLength R</li> </ul>	FIFO size corrected orrected in description for LPC nd FIFOLength updated in regised nd FIFOLength updated in regised gister updated	ster overview descri ster FIFOControl	ption		

## **NXP Semiconductors**

# CLRC663

### High performance multi-protocol NFC frontend CLRC663 and CLRC663 plus

Document ID	Release date	Data sheet status	Change notice	Supersedes
CLRC663 v.3.2	20130312	Product data sheet	-	CLRC663 v.3.5
Modifications:	Descriptive ti	PROM content tle changed inOut register (address 45h)'	: corrected	
CLRC663 v.3.1	<tbd></tbd>	Product data sheet	-	-
CLRC663 v.3.6	20130910	Product data sheet	-	CLRC663 v.3.5
Modifications:	Update of EEF	PROM content (Table 40)		
CLRC663 v.3.5	20120905	Product data sheet	-	CLRC663 v.3.4
Modifications:		ering information": updated acking information" Figures a e	dded	
CLRC663 v.3.4	20120717	Product data sheet	-	CLRC663 v.3.3
Modifications:	<u>Table 47 "Prec</u>	defined protocol overview TX	[1]": typo corrected	
CLRC663 v.3.3	20120402	Product data sheet	-	CLRC663 v.3.2
Modifications:	<ul><li>Description of</li><li>General updat</li></ul>	SAM support added e	·	
CLRC663 v.3.2	20120202	Product data sheet	-	CLRC663 v.3.1
Modifications:	<ul><li>General updat</li><li>Specification s</li></ul>	e tatus changed into "Product	data sheet"	
CLRC663 v.3.1	20110926	Preliminary data sheet	-	CLRC663 v.3.0
Modifications:	Specification s	tatus reversed from "Product	t data sheet" into "Pre	liminary data sheet"
CLRC663 v.3.0	20110919	Product data sheet	-	CLRC663 v.2.0
Modifications:	<ul> <li>Section 8 "Fun</li> <li>Section 8.7.2."</li> <li>Section 8.10.2</li> <li>Table 18 "Com "RxCtrl bits": u</li> </ul>	lering information": updated actional description". Section 1 "Product information and co "Command set overview": u amand overview", Table 57 "I pdated aracteristics": updated	onfiguration - Page 0": pdated	updated
CLRC663 v.2.0	20110615	Preliminary data sheet	-	CLRC663 v.1.0
Modifications:	General updat	e		
CLRC663 v.1.0	20110308	Objective data sheet		

## 20 Legal information

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Document status <sup>[1][2]</sup>	Product status <sup>[3]</sup>	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
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### **NXP Semiconductors**

## CLRC663

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## CLRC663

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